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Electromechanical Dynamics

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Appendix A

GLOSSARY OF COMMONLY USED SYMBOLS

Section references indicate where symbols of a given significance are introduced; grouped symbols are accompanied by their respective references. The absence of a section reference indicates that a symbol has been applied for a variety of purposes. Nomenclature used in examples is not included.

Symbol Meaning		Section	
A	cross-sectional area		
A_i	coefficient in differential equation	5.1.1	
(A_{n}^{+}, A_{n}^{-})	complex amplitudes of components of <i>n</i> th		
	mode	9.2.1	
A_w	cross-sectional area of armature conductor	6.4.1	
a	spacing of pole faces in magnetic circuit	8.5.1	
$a, (a_c, a_s)$	phase velocity of acoustic related waves	13.2.1, 11.4.1	
ab	Alfvén velocity	12.2.3	
(a, b, c)	Lagrangian coordinates	11.1	
a_i	constant coefficient in differential equation	5.1.1	
\mathbf{a}_p	instantaneous acceleration of point p fixed		
	in material	2.2.1c	
B, B_r, B_s	damping constant for linear, angular and		
	square law dampers	2.2.1b, 4.1.1, 5.2.2	
B , B _{<i>i</i>} , B_0	magnetic flux density	1.1.1a, 8.1, 6.4.2	
B _i	induced flux density	7.0	
$(B_r, B_{ra}, B_{rb}, B_{rm})$	radial components of air-gap flux		
	densities	4.1.4	
$[B_{rf}, (B_{rf})_{\rm av}]$	radial flux density due to field current	6.4.1	
b	width of pole faces in magnetic circuit	8.5	
b	half thickness of thin beam	11.4.2b	
С	contour of integration	1.1.2a	
$C, (C_a, C_b), C_o$	capacitance	2.1.2, 7.2.1a, 5.2.1	
С	coefficient in boundary condition	9.1.1	
C	the curl of the displacement	11.4	
(C^+, C^-)	designation of characteristic lines	9.1.1	

Symbol	Meaning	Section	
	specific heat capacity at constant pressure	13.1.2	
c_v	specific heat capacity at constant volume	13.1.2	
D	electric displacement	1.1.1a	
d	length		
da	elemental area	1.1.2a	
$d\mathbf{f}_n$	total elemental force on material in rigid body	2.2.1c	
dl	elemental line segment	1.1.2a	
$d\mathbf{T}_n$	torque on elemental volume of material	2.2.1c	
dV	elemental volume	1.1.2b	
E	constant of motion	5.2.1	
Ε	Young's modulus or the modulus of elasticity	9.1	
\mathbf{E}, E_{o}	electric field intensity	1.1.1a, 5.1.2d	
E_f	magnitude of armature voltage generated by field current in a synchronous		
_	machine	4.1.6a	
E_i	induced electric field intensity	7.0	
e_{11}, e_{ij}	strain tensor	9.1, 11.2	
e _{ij}	strain-rate tensor	14.1.1a	
F	magnetomotive force (mmf)	13.2.2	
F ,	force density	1.1.1a	
F_{-}	complex amplitude of $f(t)$	5.1.1	
F_0	amplitude of sinusoidal driving force	9.1.3	
f	equilibrium tension of string	9.2	
f	driving function	5.1.1	
$f, \mathbf{f}, f^e, f^s, f_j, f_i, f_1$	force	2.2.1, 2.2.1c, 3.1,	
		5.1.2a, 5.1.2b, 8.1,	
£	hitmann 1 formation	9.1	
f'	scalar function in moving coordinate	6.1	
c	system	6.1	
J C	three-dimensional surface	6.2	
J C	Integration constant	11.4.2a	
C C	a constant	5.1.2C	
C C	snear modulus of elasticity	11.2.2 6 A 1	
G	speed coefficient	2.1	
a a	air-gan length	5.2.1	
δ α α	acceleration of gravity	5120 1213	
$(\mathbf{H} H H H)$	magnetic field intensity	1 1 1a	
h	specific enthalpy	13.1.2	
$I, I, (I_r, I_s), I_f$	electrical current	10.4.3, 12.2.1a, 4.1.2.	
		6.4.1	
$(i, i_1, i_2, \dots, i_k), (i_{ar}, i_{as}, i_{br}, i_{bs}), i_a, (i_a, i_b, i_c),$	electrical current	2.1, 4.1.3, 6.4.1, 4.1.7, 6.4.1, 4.1	
$(i_{f}, i_{t}), (i_{r}, i_{s})$			

A2

Appendix	A
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Symbol	Meaning	Section	
i _n	unit vector perpendicular to area of		
	integration	6.2.1	
i _s	unit vector normal to surface of		
	integration	6.2.1	
$(i_x, i_y, i_z), (i_1, i_2, i_3)$	unit vectors in coordinate directions	2.2.1c	
J, \mathbf{J}_{f}	current density	7.0, 1.1.1a	
$J, J_r, (J_x, J_y, J_z)$	moment of inertia	5.1.2b, 4.1.1, 2.2.1c	
J_{xz}, J_{yz}	products of inertia	2.2.1c	
i	$\sqrt{-1}$	4.1.6a	
κ K	loading factor	13.2.2	
K, K,	surface current density	7.0, 1.1.1a	
ĸ	linear or torsional spring constant	2.2.1a	
K.	induced surface current density	7.0	
$k_{i}k_{a}(k_{a},k_{i})$	wavenumber	7.1.3, 10.1.3, 10.0	
k	summation index	2.1.1	
k	maximum coefficient of coupling	4.1.6b	
<i>k</i>	nth eigenvalue	9.2	
$(L, L_1, L_2), (L_2, L_4),$	inductance	2.1.1. 6.4.1. 2.1.1.	
$L_{m_1}(L_0, L_0),$		4.2.1. 4.1.1. 4.2.4	
$(L_{\alpha}, L_{\alpha}, L_{\alpha \alpha}), L_{\alpha \alpha}$		··, ··, ··-··	
L	length of incremental line segment	6.2.1	
ī	value of relative displacement for which	2.2.1a	
	spring force is zero		
I. I.m. I.	length		
M	Hartmann number	14.2.2	
M	mass of one mole of gas in kilograms	13.1.2	
M	Mach number	13.2.1	
M	mass	2.2.1c	
M	number of mechanical terminal pairs	2.1.1	
M. M.	mutual inductance	4.1.1. 4.2.4	
M	magnetization density	1.1.1a	
m	mass/unit length of string	9.2	
N	number of electrical terminal pairs	2.1.1	
N	number of turns	5.2.2	
n	number density of ions	12.3.1	
n	integer	7.1.1	
л П	unit normal vector	1.1.2	
P	polarization density	1.1.1a	
P	power	12 2 1a	
- n	number of nole pairs in a machine	418	
P p	nomer per unit area	14 2 1	
P D	pressure	5.1.2d and 12.1.4	
r n., n., n., n.,	power	4.1.6a, 4.1.6b, 4.1.2.	
revry''''''''''''''	F	4.1.6b	
Q	electric charge	7.2.1a	
q, qi, q _k	electric charge	1.1.3 and 2.1.2, 8.1, 2.1.2	
R, R _i , R _o	radius		

Symbol	Meaning	Section
$\overline{R, R_a, R_b, R_f, R_r, R_s}$	resistance	
(R, R_g)	gas constant	13.1.2
R _e	electric Reynolds number	7.0
R _m	magnetic Reynolds number	7.0
r	radial coordinate	
r	position vector of material	2.2.1c
r'	position vector in moving reference frame	6.1
r _m	center of mass of rigid body	2.2.1c
S	reciprocal modulus of elasticity	11.5.2c
S	surface of integration	1.1.2a
S	normalized frequency	7.2.4
S	membrane tension	9.2
Sz	transverse force/unit length acting on string	9.2
S	complex frequency	5.1.1
(s, s_{mT})	slip	4.1.6b
S _i	ith root of characteristic equation, a natural frequency	5.1.1
Т	period of oscillation	5.2.1
Т	temperature	13.1.2
$\mathbf{T}, T, T^e, T_{em}, T_m, T_{em}, T$	torque	2.2.1c, 5.1.2b, 3.1.1, 41.6b, 4.1.1, 6.4.1
-0, -1		64.1
т	surface force	8.4
T.,m	mechanical stress tensor	13.1.2
-1) T	the component of the stress-tensor in the	
-mn	mth-direction on a cartesian surface with	
	a normal vector in the <i>n</i> th-direction	8.1
T	constant of coulomb damping	4.1.1
T or	initial stress distribution on thin rod	911
T_0 T	longitudinal stress on a thin rod	911
T T	transverse force per unit area on	<i>y</i>
1 Z	membrane	9.2
Τ.	transverse force per unit area acting on	<i></i>
4 2	thin heam	11 4 2b
t	time	1.1.1
t	time measured in moving reference frame	6.1
1/	gravitational potential	12.1.3
	longitudinal steady velocity of string or	.2.1.0
0	membrane	10.2
11	internal energy per unit mass	13.1.1
u u	surface coordinate	11.3
$u_{a}(x-x_{a})$	unit impulse at $x = x_0$	9.2.1
ж(с	transverse deflection of wire in x -direction	10.4.3
$u_{-1}(t)$	unit step occurring at $t = 0$	5.1.2b
V.V.	velocity	7.0, 13.2.3
V	volume	1.1.2
V. V. V. V. V.	voltage	
V	potential energy	5.2.1

Appendix A

Symbol	Meaning	Section	
v, v	velocity		
(v, v_1, \ldots, v_k)	voltage	2.1.1	
$v', (v_a, v_b, v_c),$	voltage		
$v_f, v_{\rm oc}, v_t$	-		
v_n	velocity of surface in normal direction	6.2.1	
v_{o}	initial velocity distribution on thin rod	9.1.1	
vn	phase velocity	9.1.1 and 10.2	
vr	relative velocity of inertial reference frames	6.1	
v_s	$\sqrt{f/m}$ for a string under tension f and having mass/unit length m	10.1.1	
v	longitudinal material velocity on thin rod	9.1.1	
v	transverse deflection of wire in y-direction	10.4.3	
(W_e, W_m)	energy stored in electromechanical		
	coupling	3.1.1	
(W'_e, W'_m, W')	coenergy stored in electromechanical coupling	3.1.2b	
W''	hybrid energy function	5.2.1	
W	width	5.2.2	
W	energy density	11.5.2c	
w'	coenergy density	8.5	
X	equilibrium position	5.1.2a	
$(x, x_1, x_2, \ldots, x_k)$	displacement of mechanical node	2.1.1	
x	dependent variable	5.1.1	
x_p	particular solution of differential equation	5.1.1	
$(x_1, x_2, x_3), (x, y, z)$	cartesian coordinates	8.1, 6.1	
(x', y', z')	cartesian coordinates of moving frame	6.1	
(α, β)	constants along C^+ and C^- characteristics, respectively	9.1.1	
(α, β)	see (10.2.20) or (10.2.27)		
α	transverse wavenumber	11.4.3	
(α, β)	angles used to define shear strain	11.2	
(α, β)	constant angles	4.1.6b	
α	space decay parameter	7.1.4	
α	damping constant	5.1.2b	
α	equilibrium angle of torsional spring	2.2.1a	
7	ratio of specific heats	13.1.2	
γ	piezoelectric constant	11.5.2c	
$\gamma, \gamma_0, \gamma'$	angular position		
$\Delta_d(t)$	slope excitation of string	10.2.1b	
Δ_0	amplitude of sinusoidal slope excitation	10.2.1b	
Δr	distance between unstressed material points	11.2.1a	
Δs	distance between stressed positions of material points	11.2.1a	
$\delta(-)$	incremental change in ()	8.5	
$\delta, \delta_1, \delta_2$	displacement of elastic material	11.1.9.1.11.4.2a	
δ	thickness of incremental volume element	6.2.1	
δ	torque angle	4.1.6a	

Glossary of Commonly Used Symbols

Symbol	Meaning	Section	
δ_{ii}	Kronecker delta	8.1	
(δ_+, δ)	wave components traveling in the		
	$\pm x$ -directions	9.1.1	
E	linear permittivity	1.1.1b	
€0	permittivity of free space	1.1.la	
η	efficiency of an induction motor	4.1.6b	
η	second coefficient of viscosity	14.1.1c	
$\theta, \theta_i, \theta_m$	angular displacement	2.1.1, 3.1.1, 5.2.1	
θ	power factor angle; phase angle between		
	current and voltage	4.1.6a	
θ	equilibrium angle	5.2.1	
θ	angular velocity of armature	6.4.1	
θ_m	maximum angular deflection	5.2.1	
$(\lambda, \lambda_1, \lambda_2, \ldots, \lambda_k)$	magnetic flux linkage	2.1.1, 6.4.1, 4.1.7,	
λ_a		4.1.3, 4.1	
$(\lambda_a, \lambda_b, \hat{\lambda}_c)$			
$(\lambda_{ar},\lambda_{as},\lambda_{br},\lambda_{bs})$			
(λ_r, λ_s)			
2.	Lamé constant for elastic material	11.2.3	
λ	wavelength	7.1.4	
μ	linear permeability	1.1.1a	
μ , (μ_+ , μ)	mobility	12.3.1, 1.1.1b	
μ	coefficient of viscosity	14.1.1	
μ_d	coefficient of dynamic friction	2.2.16	
μ_0	permeability of free space	1.1.1a	
μ_s	coefficient of static friction	2.2.16	
v	Poisson's ratio for elastic material	11.2.2	
V VE EN	damping frequency	10.1.4	
(5 , 5)	initial deflection of string	8.2	
50 5	initial deflection of string	9.2	
S_d	amplitude of sinusoidal driving denection	9.2 0.2.1h	
$(\varsigma_n(x), \varsigma_n(x))$	amplitudes of forward and backward	9.2.10	
(5+, 5-)	traveling waves	02	
: (m)	initial valuation of string	0.2	
$\xi_0(x)$	mitial velocity of string	9.4	
ρ	free charge density	1.1.10	
Pf	surface mass density	11.1.1.4	
$\frac{\rho_8}{\Sigma}$	surface of discontinuity	62	
<u>-</u> а	conductivity	1.1.1a	
σ.	free surface charge density	1.1.1a	
σ _m	surface mass density of membrane	9.2	
- m σ ₀	surface charge density	7.2.3	
σ	surface conductivity	1.1.1a	
σ_{μ}	surface charge density	7.2.3	
τ	surface traction	8.2.1	
$ au, au_d$	diffusion time constant	7.1.1, 7.1.2a	
τ	relaxation time	7.2.1a	

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Appendix A

Symbol	Symbol Meaning	
τ _e	electrical time constant	5.2.2
$ au_m$	time for air gap to close	5.2.2
$ au_0$	time constant	5.1.3
τ_t	traversal time	7.1.2a
φ [']	electric potential	7.2
φ	magnetic flux	2.1.1
ϕ	cylindrical coordinate	2.1.1
ϕ	potential for H when $\mathbf{J}_f = 0$	8.5.2
φ	flow potential	12.2
Xe	electric susceptibility	1.1.1b
χ _m	magnetic susceptibility	1.1.1a
ψ	the divergence of the material	
	displacement	11.4
ψ	angle defined in Fig. 6.4.2	6.4.1
ψ	angular position in the air gap measured	
	from stator winding (a) magnetic axis	4.1.4
ψ	electromagnetic force potential	12.2
ψ	angular deflection of wire	10.4.3
Ω	equilibrium rotational speed	5.1.2b
Ω	rotation vector in elastic material	11.2.1a
Ω_n	real part of eigenfrequency (10.1.47)	10.1.4
ω , (ω_r , ω_s)	radian frequency of electrical excitation	4.1.6a, 4.1.2
ω	natural angular frequency (Im s)	5.1.2b
$\boldsymbol{\omega}, \boldsymbol{\omega}_m$	angular velocity	2.2.1c, 4.1.2
ω_c	cutoff frequency for evanescent waves	10.1.2
ω_d	driving frequency	9.2
ω_n	nth eigenfrequency	9.2
ω	natural angular frequency	5.1.3
(ω_r, ω_i)	real and imaginary parts of ω	10.0
∇	nabla	6.1
∇_{Σ}	surface divergence	6.2.1

REVIEW OF ELECTROMAGNETIC THEORY

B.1 BASIC LAWS AND DEFINITIONS

The laws of electricity and magnetism are empirical. Fortunately they can be traced to a few fundamental experiments and definitions, which are reviewed in the following sections. The rationalized MKS system of units is used.

B.1.1 Coulomb's Law, Electric Fields and Forces

Coulomb found that when a charge q (coulombs) is brought into the vicinity of a distribution of *charge density* $\rho_e(\mathbf{r}')$ (coulombs per cubic meter), as shown in Fig. B.1.1, a force of repulsion **f** (newtons) is given by

$$\mathbf{f} = q\mathbf{E},\tag{B.1.1}$$

where the *electric field intensity* \mathbf{E} (volts per meter) is evaluated at the position



Fig. B.1.1 The force f on the point charge q in the vicinity of charges with density $\rho_{e}(\mathbf{r}')$ is represented by the electric field intensity E times q, where E is found from (B.1.2).

 \mathbf{r} of the charge q and determined from the distribution of charge density by

$$\mathbf{E}(\mathbf{r}) = \frac{1}{4\pi\epsilon_0} \int_{V'} \rho_e(\mathbf{r}') \frac{(\mathbf{r} - \mathbf{r}')}{|\mathbf{r} - \mathbf{r}'|^3} dV'.$$
(B.1.2)

In the rationalized MKS system of units the permittivity ϵ_0 of free space is

$$\epsilon_0 = 8.854 \times 10^{-12} \approx \frac{1}{36\pi} \times 10^{-9} \, F/m.$$
 (B.1.3)

Note that the integration of (B.1.2) is carried out over all the charge distribution (excluding q), hence represents a superposition (at the location **r** of q) of the electric field intensities due to elements of charge density at the positions \mathbf{r}' .

As an example, suppose that the charge distribution $\rho_e(\mathbf{r}')$ is simply a point charge Q (coulombs) at the origin (Fig. B.1.2); that is,

$$\rho_e = Q \,\,\delta(\mathbf{r}'), \qquad (B.1.4)$$

where $\delta(\mathbf{r}')$ is the *delta function* defined by

$$\delta(\mathbf{r}') = 0, \qquad \mathbf{r}' \neq 0,$$

$$\int_{V'} \delta(\mathbf{r}') \, dV' = 1. \tag{B.1.5}$$

For the charge distribution of (B.1.4) integration of (B.1.2) gives

$$\mathbf{E}(\mathbf{r}) = \frac{Q\mathbf{r}}{4\pi\epsilon_0 |\mathbf{r}|^3}.$$
 (B.1.6)

the position r).

Hence the force on the point charge q, due to the point charge Q, is from (B.1.1)

$$\mathbf{f} = \frac{qQ\mathbf{r}}{4\pi\epsilon_0 |r|^3}.\tag{B.1.7}$$

This expression takes the familiar form of *Coulomb's law* for the force of repulsion between point charges of like sign.

We know that electric charge occurs in integral multiples of the electronic charge (1.60 \times 10⁻¹⁹ C). The charge density ρ_{e} , introduced with (B.1.2), is defined as

$$\rho_e(\mathbf{r}) = \lim_{\delta V \to 0} \frac{1}{\delta V} \sum_i q_i, \qquad (B.1.8)$$



Fig. B.1.2 Coulomb's law for point charges Q (at the origin) and q (at

where δV is a small volume enclosing the point **r** and $\sum_i q_i$ is the algebraic sum of charges within δV . The charge density is an example of a continuum model. To be valid the limit $\delta V \rightarrow 0$ must represent a volume large enough to contain a large number of charges q_i , yet small enough to appear infinitesimal when compared with the significant dimensions of the system being analyzed. This condition is met in most electromechanical systems.

For example, in copper at a temperature of 20°C the number density of free electrons available for carrying current is approximately 10^{23} electrons/ cm³. If we consider a typical device dimension to be on the order of 1 cm, a reasonable size for δV would be a cube with 1-mm sides. The number of electrons in δV would be 10^{20} , which certainly justifies the continuum model.

The force, as expressed by (B.1.1), gives the total force on a single test charge in vacuum and, as such, is not appropriate for use in a continuum model of electromechanical systems. It is necessary to use an *electric force density* F (newtons per cubic meter) that can be found by averaging (B.1.1) over a small volume.

$$\mathbf{F} = \lim_{\delta V \to 0} \frac{\sum_{i} \mathbf{f}_{i}}{\delta V} = \lim_{\delta V \to 0} \frac{\sum_{i} q_{i} \mathbf{E}_{i}}{\delta V}.$$
 (B.1.9)

Here q_i represents all of the charges in δV , \mathbf{E}_i is the electric field intensity acting on the *i*th charge, and \mathbf{f}_i is the force on the *i*th charge. As in the charge density defined by (B.1.8), the limit of (B.1.9) leads to a continuum model if the volume δV can be defined so that it is small compared with macroscopic dimensions of significance, yet large enough to contain many electronic charges. Further, there must be a sufficient amount of charge external to the volume δV that the electric field experienced by each of the test charges is essentially determined by the sources of field outside the volume. Fortunately these requirements are met in almost all physical situations that lead to useful electromechanical interactions. Because all charges in the volume δV experience essentially the same electric field \mathbf{E} , we use the definition of free charge density given by (B.1.8) to write (B.1.9) as

$$\mathbf{F} = \rho_e \mathbf{E}.\tag{B.1.10}$$

Although the static electric field intensity E can be computed from (B.1.2), it is often more convenient to state the relation between charge density and field intensity in the form of *Gauss's law*:

$$\oint_{S} \epsilon_{0} \mathbf{E} \cdot \mathbf{n} \, da = \int_{V} \rho_{e} \, dV. \tag{B.1.11}$$

In this integral law **n** is the outward-directed unit vector normal to the surface S, which encloses the volume V. It is not our purpose in this brief review to show that (B.1.11) is implied by (B.1.2). It is helpful, however, to note that



Fig. B.1.3 A hypothetical sphere of radius r encloses a charge Q at the origin. The integral of $\epsilon_0 E_r$ over the surface of the sphere is equal to the charge Q enclosed.

in the case of a point charge Q at the origin it predicts the same electric field intensity (B.1.6) as found by using (B.1.2). For this purpose the surface S is taken as the sphere of radius r centered at the origin, as shown in Fig. B.1.3. By symmetry the only component of E is radial (E_r) , and this is constant at a given radius r. Hence (B.1.11) becomes

$$4\pi r^2 E_r \epsilon_0 = Q. \tag{B.1.12}$$

Here the integration of the charge density over the volume V enclosed by S is the total charge enclosed Q but can be formally taken by using (B.1.4) with the definition provided by (B.1.5). It follows from (B.1.12) that

$$E_r = \frac{Q}{4\pi\epsilon_0 r^2},\tag{B.1.13}$$

a result that is in agreement with (B.1.6).

Because the volume and surface of integration in (B.1.11) are arbitrary, the integral equation implies a differential law. This is found by making use of the *divergence theorem**

$$\oint_{S} \mathbf{A} \cdot \mathbf{n} \, da = \int_{V} \nabla \cdot \mathbf{A} \, dV \tag{B.1.14}$$

to write (B.1.11) as

$$\int_{V} \left(\nabla \cdot \epsilon_0 \mathbf{E} - \rho_e \right) dV = 0. \tag{B.1.15}$$

* For a discussion of the divergence theorem see F. B. Hildebrand, Advanced Calculus for Engineers, Prentice-Hall, New York, 1949, p. 312.

Since the volume of integration is arbitrary, it follows that

$$\nabla \cdot \epsilon_0 \mathbf{E} = \rho_e. \tag{B.1.16}$$

From this discussion it should be apparent that this *differential* form of *Gauss's law* is implied by Coulomb's law, with the electric field intensity defined as a force per unit charge.

B.1.2 Conservation of Charge

Experimental evidence supports the postulate that electric charge is conserved. When a negative charge appears (e.g., when an electron is removed from a previously neutral atom), an equal positive charge also appears (e.g., the positive ion remaining when the electron is removed from the atom).

We can make a mathematical statement of this postulate in the following way. Consider a volume V enclosed by a surface S. If charge is conserved, the net rate of flow of electric charge out through the surface S must equal the rate at which the total charge in the volume V decreases. The current density J (coulombs per square meter-second) is defined as having the direction of flow of positive charge and a magnitude proportional to the net rate of flow of charge per unit area. Then the statement of conservation of charge is

$$\oint_{S} \mathbf{J} \cdot \mathbf{n} \, da = -\frac{d}{dt} \int_{V} \rho_{e} \, dV. \tag{B.1.17}$$

Once again it follows from the arbitrary nature of S (which is fixed in space) and the divergence theorem (B.1.14) that

$$\nabla \cdot \mathbf{J} + \frac{\partial \rho_e}{\partial t} = 0. \tag{B.1.18}$$

It is this equation that is used as a *differential* statement of *conservation of charge*.

To express conservation of charge it has been necessary to introduce a new continuum variable, the current density **J**. Further insight into the relation between this quantity and the charge density ρ_e is obtained by considering a situation in which two types of charge contribute to the current, charges q_+ with velocity \mathbf{v}_+ and charges q_- with velocity \mathbf{v}_- . The current density \mathbf{J}_+ that results from the flow of positive charge is

$$\mathbf{J}_{+} = \lim_{\delta V \to 0} \frac{1}{\delta V} \sum_{i} q_{+i} \mathbf{v}_{+i}.$$
 (B.1.19)

If we define a *charge-average velocity* \mathbf{v}_+ for the positive charges as

$$\mathbf{v}_{+} = \frac{\sum_{i} q_{+i} \mathbf{v}_{+i}}{\sum_{i} q_{+i}}$$
(B.1.20)

and the density ρ_+ of positive charges from (B.1.8) as

$$\rho_{+} = \lim_{\delta V \to 0} \frac{1}{\delta V} \sum_{i} q_{+i}, \qquad (B.1.21)$$

we can write the current density of (B.1.19) as

$$\mathbf{J}_{+} = \boldsymbol{\rho}_{+} \mathbf{v}_{+}.\tag{B.1.22}$$

Similar definitions for the charge-average velocity \mathbf{v}_{-} and charge density ρ_{-} of negative charges yields the component of current density

$$\mathbf{J}_{-} = \rho_{-} \mathbf{v}_{-}. \tag{B.1.23}$$

The total current density J is the vector sum of the two components

$$\mathbf{J} = \mathbf{J}_{+} + \mathbf{J}_{-} \tag{B.1.24}$$

Now consider the situation of a material that contains charge densities ρ_+ and ρ_- which have charge-average velocities \mathbf{v}_+ and \mathbf{v}_- with respect to the material. Assume further that the material is moving with a velocity \mathbf{v} with respect to an observer who is to measure the current. The net average velocities of positive and negative charges as seen by the observer are $\mathbf{v}_+ + \mathbf{v}$ and $\mathbf{v}_- + \mathbf{v}$, respectively. The current density measured by the observer is then from (B.1.24)

$$\mathbf{J} = (\rho_+ \mathbf{v}_+ + \rho_- \mathbf{v}_-) + \rho_e \mathbf{v}, \qquad (B.1.25)$$

where the net charge density ρ_e is given by

$$\rho_e = \rho_+ + \rho_-.$$
(B.1.26)

The first term of (B.1.25) is a net flow of charge with respect to the material and is normally called a *conduction current*. (It is often described by Ohm's law.) The last term represents the transport of net charge and is conventionally called a *convection current*. It is crucial that *net flow of charge* be distinguished from *flow of net charge*. The net charge may be zero but a current can still be accounted for by the conduction term. This is the case in metallic conductors.

B.1.3 Ampère's Law, Magnetic Fields and Forces

The magnetic flux density **B** is defined to express the force on a current element *i* dl placed in the vicinity of other currents. This element is shown in Fig. B.1.4 at the position **r**. Then, according to Ampère's experiments, the force is given by (D + D)

where

$$\mathbf{f} = i \, d\mathbf{I} \times \mathbf{B},\tag{B.1.27}$$

$$\mathbf{B} = \frac{\mu_0}{4\pi} \int_{V'} \frac{\mathbf{J} \times (\mathbf{r} - \mathbf{r}')}{|\mathbf{r} - \mathbf{r}'|^3} dV'. \tag{B.1.28}$$



Fig. B.1.4 A distribution of current density J(r') produces a force on the current element *id* which is represented in terms of the magnetic flux density **B** by (B.1.27) and (B.1.28).

Hence the flux density at the position **r** of the current element *i* dl is the superposition of fields produced by currents at the positions **r'**. In this expression the permeability of free space μ_0 is

$$\mu_0 = 4\pi \times 10^{-7} \,\mathrm{H/m.} \tag{B.1.29}$$

As an example, suppose that the distribution of current density J is composed of a current I (amperes) in the z direction and along the z-axis, as shown in Fig. B.1.5. The magnetic flux density at the position **r** can be computed



Fig. B.1.5 A current I (amperes) along the z-axis produces a magnetic field at the position r of the current element *idl*.

from (B.1.28), which for this case reduces to*

$$\mathbf{B} = \frac{\mu_0 I}{4\pi} \int_{-\infty}^{+\infty} \frac{\mathbf{i}_z \times (\mathbf{r} - z' \mathbf{i}_z)}{|\mathbf{r} - z' \mathbf{i}_z|^3} dz'.$$
(B.1.30)

Here the coordinate of the source current I is z', as shown in Fig. B.1.5, whereas the coordinate **r** that designates the position at which **B** is evaluated can be written in terms of the cylindrical coordinates (r, θ, z) . Hence (B.1.30) becomes

$$\mathbf{B} = \frac{\mu_0 I \mathbf{i}_\theta}{4\pi} \int_{-\infty}^{+\infty} \frac{\sin \psi \sqrt{(z-z')^2 + r^2}}{\left[(z-z')^2 + r^2\right]^{\frac{3}{2}}} dz', \tag{B.1.31}$$

where, from Fig. B.1.5, $\sin \psi = r/\sqrt{(z-z')^2 + r^2}$. Integration on z' gives the magnetic flux density

$$\mathbf{B} = \frac{\mu_0 I \mathbf{i}_\theta}{2\pi r} \,. \tag{B.1.32}$$

It is often more convenient to relate the magnetic flux density to the current density J by the integral of *Ampère's law* for static fields, which takes the form

$$\oint_C \mathbf{B} \cdot d\mathbf{l} = \mu_0 \int_S \mathbf{J} \cdot \mathbf{n} \, da. \tag{B.1.33}$$

Here C is a closed contour of line integration and S is a surface enclosed by C. We wish to present a review of electromagnetic theory and therefore we shall not embark on a proof that (B.1.33) is implied by (B.1.28). Our purpose is served by recognizing that (B.1.33) can also be used to predict the flux density in the situation in Fig. B.1.5. By symmetry we recognize that **B** is azimuthally directed and independent of θ and z. Then, if we select the contour C in a plane z equals constant and at a radius r, as shown in Fig. B.1.5, (B.1.33) becomes

$$2\pi r B_{\theta} = \mu_0 I. \tag{B.1.34}$$

Solution of this expression for B_{θ} gives the same result as predicted by (B.1.28). [See (B.1.32).]

The contour C and surface S in (B.1.33) are arbitrary and therefore the equation can be cast in a differential form. This is done by using Stokes' theorem^{\dagger},

$$\oint_C \mathbf{A} \cdot d\mathbf{l} = \int_S \mathbf{n} \cdot (\nabla \times \mathbf{A}) \, da, \qquad (B.1.35)$$

* Unit vectors in the coordinate directions are designated by i. Thus i_z is a unit vector in the z-direction.

[†] See F. B. Hildebrand, Advanced Calculus for Engineers, Prentice-Hall, New York, 1949, p. 318.

to write (B.1.33) as

$$\int_{S} (\nabla \times \mathbf{B} - \mu_0 \mathbf{J}) \cdot \mathbf{n} \, da = 0, \qquad (B.1.36)$$

from which the differential form of Ampère's law follows as

$$\boldsymbol{\nabla} \times \mathbf{B} = \boldsymbol{\mu}_0 \mathbf{J}. \tag{B.1.37}$$

So far the assumption has been made that the current **J** is constant in time. Maxwell's contribution consisted in recognizing that if the sources ρ_e and **J** (hence the fields **E** and **B**) are time varying the displacement current $\epsilon_0 \partial E/\partial t$ must be included on the right-hand side of (B.1.37). Thus for dynamic fields Ampère's law takes the form

$$\nabla \times \mathbf{B} = \mu_0 \mathbf{J} + \mu_0 \frac{\partial \epsilon_0 \mathbf{E}}{\partial t}.$$
 (B.1.38)

This alteration of (B.1.37) is necessary if conservation of charge expressed by (B.1.18) is to be satisfied. Because the divergence of any vector having the form $\nabla \times A$ is zero, the divergence of (B.1.38) becomes

$$\nabla \cdot \mathbf{J} + \frac{\partial (\nabla \cdot \epsilon_0 \mathbf{E})}{\partial t} = 0.$$
 (B.1.39)

Then, if we recall that ρ_e is related to E by Gauss's law (B.1.16), the conservation of charge equation (B.1.18) follows. The displacement current in (B.1.38) accounts for the rate of change of ρ_e in (B.1.18).

We shall make considerable use of Ampère's law, as expressed by (B.1.38), with Maxwell's displacement current included. From our discussion it is clear that the static form of this law results from the force law of interaction between currents. The magnetic flux density is defined in terms of the force produced on a current element. Here we are interested primarily in a continuum description of the force, hence require (B.1.27) expressed as a force density. With the same continuum restrictions implied in writing (B.1.10), we write the magnetic force density (newtons per cubic meter) as

$$\mathbf{F} = \mathbf{J} \times \mathbf{B}.\tag{B.1.40}$$

In view of our remarks it should be clear that this force density is not something that we have derived but rather arises from the definition of the flux density **B**. Further remarks on this subject are found in Section 8.1.

B.1.4 Faraday's Law of Induction and the Potential Difference

Two extensions of static field theory are required to describe dynamic fields. One of these, the introduction of the displacement current in Ampère's law, was discussed in the preceding section. Much of the significance of this generalization stems from the apparent fact that an electric field can lead to the generation of a magnetic field. As a second extension of static field theory, Faraday discovered that, conversely, time-varying magnetic fields can lead to the generation of electric fields.

Faraday's law of induction can be written in the integral form

$$\oint_C \mathbf{E} \cdot d\mathbf{l} = -\frac{d}{dt} \int_S \mathbf{B} \cdot \mathbf{n} \, da, \qquad (B.1.41)$$

where again C is a contour that encloses the surface S. The contour and surface are arbitrary; hence it follows from Stokes' theorem (B.1.35) that Faraday's law has the differential form

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t} \,. \tag{B.1.42}$$

Note that in the static case this expression reduces to $\nabla \times \mathbf{E} = 0$, which is, in addition to Gauss's law, a condition on the static electric field. That this further equation is consistent with the electric field, as given by (B.1.2), is not shown in this review. Clearly the one differential equation represented by Gauss's law could not alone determine the three components of \mathbf{E} .

In regions in which the magnetic field is either static or negligible the electric field intensity can be derived as the gradient of a scalar potential ϕ :

$$\mathbf{E} = -\nabla\phi. \tag{B.1.43}$$

This is true because the curl of the gradient is zero and (B.1.42) is satisfied. The difference in potential between two points, say a and b, is a measure of the line integral of **E**, for

$$\int_{a}^{b} \mathbf{E} \cdot d\mathbf{l} = -\int_{a}^{b} \nabla \phi \cdot d\mathbf{l} = \phi_{a} - \phi_{b}.$$
 (B.1.44)

The potential difference $\phi_a - \phi_b$ is referred to as the voltage of point *a* with respect to *b*. If there is no magnetic field **B** in the region of interest, the integral of (B.1.44) is independent of path. In the presence of a time-varying magnetic field the integral of **E** around a closed path is not in general zero, and if a potential is defined in some region by (B.1.43) the path of integration will in part determine the measured potential difference.

The physical situation shown in Fig. B.1.6 serves as an illustration of the implications of Faraday's law. A magnetic circuit is excited by a current source I(t) as shown. Because the magnetic material is highly permeable, the induced flux density B(t) is confined to the cross section A which links a circuit formed by resistances R_a and R_b in series. A cross-sectional view of the



Fig. B.1.6 (a) A magnetic circuit excited by I(t) so that flux AB(t) links the resistive loop (b) a cross-sectional view of the loop showing connection of the voltmeters.

circuit is shown in Fig. B.1.6b, in which high impedance voltmeters v_a and v_b are shown connected to the same nodes. Under the assumption that no current is drawn by the voltmeters, and given the flux density B(t), we wish to compute the voltages that would be indicated by v_a and v_b .

Three contours of integration C are defined in Fig. B.1.6b and are used with Faraday's integral law (B.1.41). The integral of E around the contour C_c is equal to the drop in potential across both of the resistances, which carry the same current *i*. Hence, since this path encloses a total flux AB(t), we have

$$i(R_a + R_b) = -\frac{d}{dt} [AB(t)].$$
 (B.1.45)

The paths of integration C_a and C_b do not enclose a magnetic flux; hence for

these paths (B.1.41) gives

$$v_a = -iR_a = \frac{R_a}{R_a + R_b} \frac{d}{dt} [AB(t)] \quad \text{for} \quad C_a, \tag{B.1.46}$$

$$v_b = iR_b = \frac{-R_b}{R_a + R_b} \frac{d}{dt} [AB(t)]$$
 for C_b , (B.1.47)

where the current i is evaluated by using (B.1.45). The most obvious attribute of this result is that although the voltmeters are connected to the same nodes they do not indicate the same values. In the presence of the magnetic induction the contour of the voltmeter leads plays a role in determining the voltage indicated.

The situation shown in Fig. B.1.6 can be thought of as a transformer with a single turn secondary. With this in mind, it is clear that Faraday's law plays an essential role in electrical technology.

The divergence of an arbitrary vector $\nabla \times \mathbf{A}$ is zero. Hence the divergence of (B.1.42) shows that the divergence of **B** is constant. This fact also follows from (B.1.28), from which it can be shown that this constant is zero. Hence an additional differential equation for **B** is

$$\nabla \cdot \mathbf{B} = 0. \tag{B.1.48}$$

Integration of this expression over an arbitrary volume V and use of the divergence theorem (B.1.14) gives

$$\oint_{S} \mathbf{B} \cdot \mathbf{n} \, da = 0. \tag{B.1.49}$$

This integral law makes more apparent the fact that there can be no net magnetic flux emanating from a given region of space.

B.2 MAXWELL'S EQUATIONS

The generality and far-reaching applications of the laws of electricity and magnetism are not immediately obvious; for example, the law of induction given by (B.1.42) was recognized by Faraday as true when applied to a conducting circuit. The fact that (B.1.42) has significance even in regions of space unoccupied by matter is a generalization that is crucial to the theory of electricity and magnetism. We can summarize the differential laws introduced in Section B.1 as

$$\nabla \cdot \epsilon_0 \mathbf{E} = \rho_e, \tag{B.2.1}$$

$$\nabla \cdot \mathbf{J} + \frac{\partial \rho_e}{\partial t} = 0, \qquad (B.2.2)$$

$$\nabla \times \mathbf{B} = \mu_0 \mathbf{J} + \mu_0 \frac{\partial \epsilon_0 \mathbf{E}}{\partial t}, \qquad (B.2.3)$$

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t},$$
 (B.2.4)

$$\nabla \cdot \mathbf{B} = \mathbf{0}.\tag{B.2.5}$$

Taken together, these laws are called *Maxwell's equations* in honor of the man who was instrumental in recognizing that they have a more general significance than any one of the experiments from which they originate. For example, we can think of a time-varying magnetic flux that induces an electric field according to (B.2.4) even in the absence of a material circuit. Similarly, (B.2.3) is taken to mean that even in regions of space in which there is no circuit, hence J = 0, a time-varying electric field leads to an induced magnetic flux density **B**.

The coupling between time-varying electric and magnetic fields, as predicted by (B.2.1 to B.2.5), accounts for the existence of electromagnetic waves, whether they be radio or light waves or even gamma rays. As we might guess from the electromechanical origins of electromagnetic theory, the propagation of electromagnetic waves is of secondary importance in the study of most electromechanical phenomena. This does not mean that electromechanical interactions are confined to frequencies that are low compared with radio frequencies. Indeed, electromechanical interactions of practical significance extend into the gigahertz range of frequencies.

To take a mature approach to the study of electromechanics it is necessary that we discriminate at the outset between essential and nonessential aspects of interactions between fields and media. This makes it possible to embark immediately on a study of nontrivial interactions. An essential purpose of this section is the motivation of approximations used in this book.

Although electromagnetic waves usually represent an unimportant consideration in electromechanics and are not discussed here in depth, they are important to an understanding of the quasi-static approximations that are introduced in Section B.2.2. Hence we begin with a brief simplified discussion of electromagnetic waves.

B.2.1 Electromagnetic Waves

Consider fields predicted by (B.2.3) and (B.2.4) in a region of free space in which J = 0. In particular, we confine our interest to situations in which the fields depend only on (x, t) (the fields are one-dimensional) and write the y-component of (B.2.3) and the z-component of (B.2.4)

$$-\frac{\partial B_z}{\partial x} = \mu_0 \epsilon_0 \frac{\partial E_y}{\partial t}, \qquad (B.2.6)$$

$$\frac{\partial E_y}{\partial x} = -\frac{\partial B_z}{\partial t}.$$
 (B.2.7)

This pair of equations, which make evident the coupling between the dynamic electric and magnetic fields, is sufficient to determine the field components B_z and E_y . In fact, if we take the time derivative of (B.2.6) and use the resulting

expression to eliminate B_z from the derivative with respect to x of (B.2.7), we obtain

$$\frac{\partial^2 E_y}{\partial x^2} = \frac{1}{c^2} \frac{\partial^2 E_y}{\partial t^2},$$
(B.2.8)

where

$$c = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = 3 \times 10^8 \text{ (m/sec)}.$$

This equation for E_y is called the *wave equation* because it has solutions in the form of

$$E_{y}(x, t) = E_{+}(x - ct) + E_{-}(x + ct).$$
(B.2.9)

That this is true may be verified by substituting (B.2.9) into (B.2.8). Hence solutions for E_v can be analyzed into components E_+ and E_- that represent waves traveling, respectively, in the +x- and -x-directions with the velocity of light c, given by (B.2.8). The prediction of electromagnetic wave propagation is a salient feature of Maxwell's equations. It results, as is evident from the derivation, because time-varying magnetic fields can induce electric fields [Faraday's law, (B.2.7)] while at the same time dynamic electric fields induce magnetic fields [Ampère's law with the displacement current included (B.2.6)]. It is also evident from the derivation that if we break this two-way coupling by leaving out the displacement current or omitting the magnetic induction term electromagnetic waves are not predicted.

Electromechanical interactions are usually not appreciably affected by the propagational character of electromagnetic fields because the velocity of propagation c is very large. Suppose that we are concerned with a system whose largest dimension is l. The time l/c required for the propagation of a wave between extremes of the system is usually short compared with characteristic dynamical times of interest; for example, in a device in which l = 0.3 m the time l/c equals 10^{-9} sec. If we were concerned with electromechanical motions with a time constant of a microsecond (which is extremely short for a device characterized by 30 cm), it would be reasonable to ignore the wave propagation. In the absence of other dynamic effects this could be done by assuming that the fields were established everywhere within the device instantaneously.

Even though it is clear that the propagation of electromagnetic waves has nothing to do with the dynamics of interest, it is not obvious how to go about simplifying Maxwell's equations to remove this feature of the dynamics. A pair of particular examples will help to clarify approximations made in the next section. These examples, which are considered simultaneously so that they can be placed in contrast, are shown in Fig. B.2.1.

B14



Fig. B.2.1 Perfectly conducting plane-parallel electrodes driven at x = -l: (a) $i(t) = i_0 \cos \omega t$; (b) $v(t) = v_0 \cos \omega t$.

A pair of perfectly conducting parallel plates has the spacing s which is much smaller than the x-z dimensions l and d. The plates are excited at x = -l by

a current source $i(t) = i_o \cos \omega t$ (amperes). (B.2.10*a*) $v(t) = v_o \cos \omega t$ (volts). (B.2.10*b*) At x = 0, the plates are terminated in

a perfectly conducting short circuit a plate.

an open circuit.

If we assume that the spacing s is small enough to warrant ignoring the effects of fringing and that the driving sources at x = -l are distributed along the z-axis, the one-dimensional fields B_z and E_y predicted by (B.2.6) and (B.2.7) represent the fields between the plates. Hence we can think of the current and voltage sources as exciting electromagnetic waves that propagate along the x-axis between the plates. The driving sources impose conditions on the fields at x = -l. They are obtained by

integrating (B.1.33) around the contour C (Fig. B.2.2*a*) which encloses the upper plate adjacent to the current source. (The surface S enclosed by C is very thin so that negligible displacement current links the loop).

integrating the electric field between (a) and (b) in Fig. B.2.2b to relate the potential difference of the voltage source to the electric field intensity $E_u(-l, t)$.

$$B_{z}(-l, t) = -\mu_{0}K = -\frac{\mu_{0}i(t)}{d} \int_{s}^{0} E_{y} dy = -sE_{y}(-l, t) = v(t).$$
(B.2.11*a*) (B.2.11*b*)



Fig. B.2.2 Boundary conditions for the systems in Fig. B.2.1

Similar conditions used at x = 0 give the boundary conditions

$$E_y(0, t) = 0$$
 (B.2.12a) $| B_z(0, t) = 0$ (B.2.12b)

It is not our purpose in this chapter to become involved with the formalism of solving the wave equation [or (B.2.6) and (B.2.7)] subject to the boundary conditions given by (B.2.11) and (B.2.12). There is ample opportunity to solve boundary value problems for electromechanical systems in the text, and the particular problem at hand forms a topic within the context of transmission lines and waveguides. For our present purposes, it suffices to guess solutions to these equations that will satisfy the appropriate boundary conditions. Then direct substitution into the differential equations will show that we have made the right choice.

$$E_{y} = -i_{o} \frac{\sin \omega t \sin (\omega x/c)}{d\epsilon_{0}c \cos (\omega l/c)}, \qquad E_{y} = -\frac{v_{o} \cos \omega t \cos (\omega x/c)}{s \cos (\omega l/c)}, \qquad (B.2.13a)$$

$$B_{z} = -\frac{\mu_{0}i_{o} \cos \omega t \cos (\omega x/c)}{d \cos (\omega l/c)}, \qquad B_{z} = -\frac{v_{o} \sin \omega t \sin (\omega x/c)}{c s \cos (\omega l/c)} \qquad (B.2.14a)$$

$$B_{z} = -\frac{v_{o} \sin \omega t \sin (\omega x/c)}{c s \cos (\omega l/c)} \qquad (B.2.14b)$$

Note that at x = -l the boundary conditions B.2.11 are satisfied, whereas at x = 0 the conditions of (B.2.12) are met. One way to show that Maxwell's equations are satisfied also (aside from direct substitution) is to use trigometric identities* to rewrite these standing wave solutions as the superposition of two traveling waves in the form of (B.2.9). Our solutions are sinusoidal, steady-state solutions, so that with the understanding that the amplitude of the field at any point along the x-axis is varying sinusoidally with time we can obtain an impression of the dynamics by plotting the instantaneous amplitudes, as shown in Fig. B.2.3. In general, the fields have the sinusoidal distribution along the x-axis of a standing wave. From (B.2.13 to B.2.14) it

* For example in (B.2.13a) sin $\omega t \sin(\omega x/c) \equiv \frac{1}{2} \{ \cos [\omega(t-x/c)] - \cos [\omega(t+x/c)] \}$.





Fig. B.2.3 Amplitude of the electric field intensity and magnetic flux density along the x-axis of the parallel-plate structures shown in Fig. B.2.1 For these plots $\omega l/c = 3\pi/4$.

is clear that as a function of time the electric field reaches its maximum amplitude when $B_z = 0$ and vice versa. Hence the amplitudes of E_y and B_z shown in Fig. B.2.3 are for different instants of time. The fields near x = 0do not in general have the same phase as those excited at x = -l. If, however, we can make the approximation that times of interest (which in this case are $1/\omega$) are much longer than the propagation time l/c,

$$\frac{l/c}{1/\omega} = \frac{\omega l}{c} \ll 1. \tag{B.2.15}$$

The sine functions can then be approximated by their arguments (which are small compared with unity) and the cosine functions are essentially equal to unity. Hence, when (B.2.15) is satisfied, the field distributions (B.2.13) and (B.2.14) become

$$E_y \simeq -\frac{i_o \sin \omega t}{d\epsilon_0 c} \left(\frac{\omega x}{c}\right), \quad (B.2.16a) \qquad E_y \simeq -\frac{v_o}{s} \cos \omega t, \qquad (B.2.16b)$$

$$B_z \simeq -\frac{\mu_0 i_o \cos \omega t}{d}$$
, (B.2.17a) $B_z \simeq -\frac{v_o}{cs} \sin \omega t \left(\frac{\omega x}{c}\right)$. (B.2.17b)

The distribution of field amplitudes in this limit is shown in Fig. B.2.4. The most significant feature of the limiting solutions is that

the magnetic field between the short-circuited plates has the same distribution as if the excitation current were static. the electric field between the opencircuited plates has the same distribution as if the excitation voltage were constant.



Fig. B.2.4 The distribution of field amplitudes between the parallel plates of Fig. B.2.1 in the limit in which $(\omega l/c) \ll 1$.

Note that the fields as they are excited at x = -l retain the same phase everywhere between the plates. This simply reflects the fact that according to the approximate equations there is no time lag between an excitation at x = -l and the field response elsewhere along the x-axis. It is in this limit that the ideas of circuit theory are applicable, for if we now compute

the voltage v(t) at x = -l

$$v(t) = -sE_y(-l, t)$$
 (B.2.18*a*)

we obtain the terminal equation for an inductance

$$v = L \frac{d}{dt} (i_a \cos \omega t), \qquad (B.2.19a)$$

where the inductance L is

 $L=\frac{sl\mu_0}{d}.$

the current
$$i(t)$$
 at $x = -l$

$$i(t) = -B_z(-l, t) \frac{a}{\mu_0}$$
 (B.2.18b)

we obtain the terminal equation for a capacitance

$$i(t) = C \frac{d}{dt} (v_o \cos \omega t), \quad (B.2.19b)$$

where the capacitance C is
$$C = \frac{\epsilon_0 dl}{\epsilon_0 dl}$$

A comparison of the examples will be useful for motivating many of the somewhat subtle ideas introduced in the main body of the book. One of the most important points that we can make here is that even though we have solved the same pair of Maxwell's equations (B.2.6) and (B.2.7) for both examples, subject to the same approximation that $\omega l/c \ll 1$ (B.2.15), we have been led to very different physical results. The difference between these

two examples arises from the boundary condition at x = 0. In the case of

a short circuit a static excitation leads to a uniform magnetic field but no electric field. The electric field is generated by Faraday's law because the magnetic field is in fact *only quasi-static* and varies slowly with time. an open circuit a static excitation results in a uniform electric field but no magnetic field. The magnetic field is induced by the displacement current in Ampère's law because the electric field is, in fact, *only quasi-static* and varies slowly with time.

B.2.2 Quasi-Static Electromagnetic Field Equations

As long as we are not interested in phenomena related to the propagation of electromagnetic waves, it is helpful to recognize that most electromechanical situations are in one of two classes, exemplified by the two cases shown in Fig. B.2.1. In the situation in which the plates are short-circuited together (Fig. B.2.1a) the limit $\omega l/c \ll 1$ means that the displacement current is of negligible importance. A characteristic of this system is that with a static excitation a large current results; hence there is a large static magnetic field. For this reason it exemplifies a magnetic field system. By contrast, in the case in which the plates are open-circuited, as shown in Fig. B.2.1b, a static excitation gives rise to a static electric field but no magnetic field. This example exemplifies an *electric field system*, in which the *magnetic induction* of Faraday's law is of negligible importance. To emphasize these points consider how we can use these approximations at the outset to obtain the approximate solutions of (B.2.19). Suppose that the excitations in Fig. B.2.1 were static. The fields between the plates are then independent of x and given by

$$E_y = 0,$$
 (B.2.20*a*) $E_y = -\frac{v}{s},$ (B.2.20*b*)

$$B_z = -\frac{\mu_0 i}{d}$$
, (B.2.21*a*) $B_z = 0$. (B.2.21*b*)

Now suppose that the fields vary slowly with time [the systems are quasistatic in the sense of a condition like (B.2.15)]. Then *i* and *v* in these equations are time-varying, hence

B_z is a function of time	2.	E_y is a function of time.	
From Faraday's law of	induction as	From Ampère's law, as	expressed
expressed by (B.2.7)		by (B.2.6)	
$\frac{\partial E_y}{\partial x} = \frac{\mu_0}{d} \frac{di}{dt} .$	(B.2.22 <i>a</i>)	$\frac{\partial B_z}{\partial x} = \frac{\mu_0 \epsilon_0}{s} \frac{dv}{dt} .$	(B.2.22 <i>b</i>)

Now the right-hand side of each of these equations is independent of x; hence they can be integrated on x. At the same time, we recognize that

$$E_y(0, t) = 0,$$
 (B.2.23*a*) $B_z(0, t) = 0,$ (B.2.23*b*)

so that integration gives

$$E_y = \frac{\mu_0 x}{d} \frac{di}{dt}.$$
 (B.2.24a)
$$B_z = \frac{\mu_0 \epsilon_0 x}{s} \frac{dv}{dt}.$$
 (B.2.24b)

Recall how the terminal voltage and current are related to these field quantities (B.2.18) and these equations become

$$v(t) = L \frac{di}{dt}$$
, (B.2.25*a*) $i(t) = C \frac{dv}{dt}$, (B.2.25*b*)

where again the inductance L and capacitance C are defined as following (B.2.19). Hence making these approximations at the outset has led to the same approximate results as those found in the preceding section by computing the exact solution and taking the limits appropriate to $\omega l/c \ll 1$.

The simple example in Fig. B.2.1 makes it plausible that Maxwell's equations can be written in two quasi-static limits appropriate to the analysis of two major classes of electromechanical interaction:

Magnetic Field Systems		Electric Field Systems	
$\mathbf{\nabla} \times \mathbf{B} = \mu_0 \mathbf{J},$	(B.2.26 <i>a</i>)	$\boldsymbol{\nabla} \times \mathbf{B} = \mu_0 \mathbf{J} + \mu_0 \epsilon_0 \frac{\partial \mathbf{J}}{\partial t}$	$\frac{E}{t}$, (B.2.26b)
$\mathbf{\nabla} \mathbf{\times} \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t},$	(B.2.27 <i>a</i>)	$\nabla \times \mathbf{E} = 0,$	(B .2.27 <i>b</i>)
$\boldsymbol{\nabla} \cdot \mathbf{B} = 0,$	(B.2.28 <i>a</i>)	$\boldsymbol{\nabla} \cdot \boldsymbol{\epsilon}_0 \mathbf{E} = \boldsymbol{\rho}_e,$	(B.2.28 <i>b</i>)
$\nabla \cdot \mathbf{J} = 0,$	(B.2.29a)	$\nabla \cdot \mathbf{J} + \frac{\partial \rho_e}{\partial t} = 0.$	(B.2.29 <i>b</i>)

Here the displacement current has been omitted from Ampère's law in the magnetic field system, whereas the magnetic induction has been dropped from Faraday's law in the electric field system. Note that if the displacement current is dropped from (B.2.26a) the charge density must be omitted from the conservation of charge equation (B.2.29a) because the latter expression is the divergence of (B.2.26a).

We have not included Guass's law for the charge density in the magnetic field system or the divergence equation for **B** in the electric field system because in the respective situations these expressions are of no interest. In fact, only the divergence of (B.2.26b) is of interest in determining the dynamics of most electric field systems and that is (B.2.29b).

B20

It must be emphasized that the examples of Fig. B.2.1 serve only to motivate the approximations introduced by (B.2.26 to B.2.29). The two systems of equations have a wide range of application. The recognition that a given physical situation can be described as a magnetic field system, as opposed to an electric field system, requires judgment based on experience. A major intent of this book is to establish that kind of experience.

In the cases of Fig. B.2.1 we could establish the accuracy of the approximate equations by calculating effects induced by the omitted terms; for example, in the magnetic field system of Fig. B.2.1*a* we ignored the displacement current to obtain the quasi-static solution of (B.2.21*a*) and (B.2.24*a*). We could now compute the correction B_z^c to the quasi-static magnetic field induced by the displacement current by using (B.2.6), with **E** given by (B.2.24*a*). This produces

$$\frac{\partial B_z^{\ c}}{\partial x} = -\frac{\mu_0^2 \epsilon_0 x}{d} \frac{d^2 i}{dt^2}.$$
 (B.2.30)

Because the right-hand side of this expression is a known function of x, it can be integrated. The constant of integration is evaluated by recognizing that the quasi-static solution satisfies the driving condition at x = -l; hence the correction field B_z^{c} must be zero there and

$$B_{z}^{c} = -\frac{\mu_{0}^{2}\epsilon_{0}(x^{2}-l^{2})}{2d}\frac{d^{2}i}{dt^{2}}.$$
 (B.2.31)

Now, to determine the error incurred in ignoring this field we take the ratio of its largest value (at x = 0) to the quasi-static field of (B.2.21a):

$$\frac{|B_z|}{|B_z|} = \frac{l^2}{2c^2} \frac{|d^2i/dt^2|}{|i|}.$$
 (B.2.32)

If this ratio is small compared with 1, the quasi-static solution is adequate. It is evident that in this case the ratio depends on the time rate of change of the excitation. In Section B.2.1, in which $i = i_o \cos \omega t$, (B.2.32) becomes

$$\frac{|B_z^c|}{|B_z|} = \frac{1}{2} \left(\frac{\omega l}{c}\right)^2 \ll 1, \tag{B.2.33}$$

which is essentially the same condition given by (B.2.15).

Once the fields have been determined by using either the magnetic field or the electric field representation it is possible to calculate the effects of the omitted terms. This procedure results in a condition characterized by (B.2.33). For this example, if the device were 30 cm long and driven at 1 MHz (this is an extremely high frequency for anything 30 cm long to respond to electromechanically) (B.2.33) becomes

$$\frac{1}{2} \left(\frac{\omega l}{c}\right)^2 = \frac{1}{2} \left(\frac{2 \cdot \pi \cdot 10^6 \cdot 0.3}{3 \times 10^8}\right)^2 = 2\pi^2 \times 10^{-6} \ll 1, \qquad (B.2.34)$$

and the quasi-static approximation is extremely good.

It is significant that the magnetic and electric field systems can be thought of in terms of their respective modes of electromagnetic energy storage. In the quasi-static systems the energy that can be attributed to the electromagnetic fields is stored either in the magnetic or electric field. This can be seen by using (B.2.26 to B.2.27) to derive Poynting's theorem for the conservation of electromagnetic energy. If the equations in (B.2.27) are multiplied by \mathbf{B}/μ_0 and subtracted from the equations in (B.2.26) multiplied by \mathbf{E}/μ_0 , it follows that

Then, because of a vector identity,* these equations take the form

$$-\nabla \cdot \left(\mathbf{E} \times \frac{\mathbf{B}}{\mu_{0}}\right) = \mathbf{E} \cdot \mathbf{J} + \frac{\partial}{\partial t} \left(\frac{1}{2} \frac{\mathbf{B} \cdot \mathbf{B}}{\mu_{0}}\right). \quad (\mathbf{B}.2.36a) + \frac{\partial}{\partial t} \left(\frac{1}{2} \epsilon_{0} \mathbf{E} \cdot \mathbf{E}\right). \quad (\mathbf{B}.2.36b)$$

Now, if we integrate these equations over a volume V enclosed by a surface S, the divergence theorem (B.1.14) gives

$$-\oint_{S} \frac{\mathbf{E} \times \mathbf{B}}{\mu_{0}} \cdot \mathbf{n} \, da = \int_{V} \mathbf{E} \cdot \mathbf{J} \, dV + \frac{\partial}{\partial t} \int_{V} w \, dV, \qquad (B.2.37)$$

where

$$w = \frac{1}{2} \frac{\mathbf{B} \cdot \mathbf{B}}{\mu_0}. \qquad (\mathbf{B}.2.38a) \quad | \quad w = \frac{1}{2} \epsilon_0 \mathbf{E} \cdot \mathbf{E}. \qquad (\mathbf{B}.2.38b)$$

The term on the left in (B.2.37) (including the minus sign) can be interpreted as the flux of energy into the volume V through the surface S. This energy is either dissipated within the volume V, as expressed by the first term on the right, or stored in the volume V, as expressed by the second term. Hence

* $\nabla \cdot (\mathbf{A} \times \mathbf{C}) = \mathbf{C} \cdot \nabla \times \mathbf{A} - \mathbf{A} \cdot \nabla \times \mathbf{C}.$

(w) can be interpreted as an electromagnetic energy density. The electromagnetic energy of the magnetic field system is stored in the magnetic field alone. Similarly, an electric field system is one in which the electromagnetic energy is stored in the electric field.

The familiar elements of electrical circuit theory illustrate the division of interactions into those defined as magnetic field systems and those defined as electric field systems. From the discussion in this and the preceding section it is evident that the short-circuited plates in Fig. B.2.1 constitute an inductor, whereas the open-circuited plates can be represented as a capacitor. This fact is the basis for the development of electromechanical interactions undertaken in Chapter 2. From this specific example it is evident that the magnetic field system includes interactions in which we can define lumped-parameter variables like the inductance, but it is not so evident that this model also describes the magnetic field and the magnetoelastic interactions of solids in a magnetic field, even including electromechanical aspects of microwave magnetics.

Similarly, the electric field system includes not only the electromechanics of systems that can be modeled in terms of circuit concepts like the capacitance but ferroelectric interactions between solids and electric fields, the electrohydrodynamics of a variety of liquids and slightly ionized gases in an electric field, and even the most important oscillations of an electron beam. Of course, if we are interested in the propagation of an electromagnetic wave through an ionospheric plasma or through the slightly ionized wake of a space vehicle, the full set of Maxwell's equations must be used.

There are situations in which the propagational aspects of the electromagnetic fields are not of interest, yet neither of the quasi-static systems is appropriate. This is illustrated by short-circuiting the parallel plates of Fig. B.2.1 at x = 0 by a resistive sheet. A static current or voltage applied to the plates at x = -l then leads to both electric and magnetic fields between the plates. If the resistance of the sheet is small, the electric field between the plates is also small, and use of the exact field equations would show that we are still justified in ignoring the displacement current. In this case the inductance of Fig. B.2.1*a* is in series with a resistance. In the opposite extreme, if the resistance of the resistive sheet were very high, we would still be justified in ignoring the magnetic induction of Faraday's law. The situation shown in Fig. B.2.1*b* would then be modeled by a capacitance shunted by a resistance. The obvious questions are, when do we make a transition from the first case to the second and why is not this intermediate case of more interest in electromechanics?

The purpose of practical electromechanical systems is either the conversion of an electromagnetic excitation into a force that can perform work on a mechanical system or the reciprocal generation of electromagnetic energy from a force of mechanical origin. From (B.1.10) and (B.1.40) there are two fundamental types of electromagnetic force. Suppose that we are interested in producing a force of electrical origin on the upper of the two plates in Fig. B.2.1. We have the option of imposing a large current to interact with its induced magnetic field or of using a large potential to create an electric field that would interact with induced charges on the upper plate. Clearly, we are not going to impose a large potential on the plates if they are terminated in a small resistance or attempt to drive a large current through the plates with an essentially open circuit at x = 0. The electrical dissipation in both cases would be prohibitively large. More likely, if we intended to use the force $\mathbf{J} \times \mathbf{B}$, we would make the resistance as small as possible to minimize the dissipation of electric power and approach the case of Fig. B.2.1a. The essentially open circuit shown in Fig. B.2.1b would make it possible to use a large potential to create a significant force of the type $\rho_e E$ without undue power dissipation. In the intermediate case the terminating resistance could be adjusted to make the electric and magnetic forces about equal. As a practical matter, however, the resulting device would probably melt before it served any useful electromechanical function. The power dissipated in the termination resistance would be a significant fraction of any electric power converted to mechanical form.*

The energy densities of (B.2.38) provide one means of determining when the problem shown in Fig. B.2.1 (but with a resistive sheet terminating the plates at x = 0) is intermediate between a magnetic and an electric field system. In the intermediate case the energy densities are equal

$$\frac{1}{2}\epsilon_0 \mathbf{E} \cdot \mathbf{E} = \frac{1}{2} \frac{\mathbf{B} \cdot \mathbf{B}}{\mu_0}.$$
 (B.2.39)

Now, if the resistive sheet has a total resistance of R, then from (B.2.18a) applied at x = 0 (B.2.40)

$$E_{y}s = -iR. \tag{B.2.40}$$

The current can be evaluated in terms of the magnetic field at x = 0 by using (B.2.18b):

$$E_y s = B_z \frac{dR}{\mu_0} \,. \tag{B.2.41}$$

Substitution of the electric field, as found from this expression into (B.2.39), gives (B, b, c) = (B, b, c)

$$\frac{\epsilon_0}{2} B_z^2 \left(\frac{Rd}{s\mu_0}\right)^2 = \frac{1}{2} \frac{B_z^2}{\mu_0}.$$
 (B.2.42)

* It is interesting that for this particular intermediate case the electric force tends to pull the plates together, whereas the magnetic force tends to push them apart. Hence, because the two forces are equal in magnitude, they just cancel. Hence, if the energy densities are equal, we obtain the following relation among the physical parameters of the system:

$$\frac{dR}{s} = \left(\frac{\mu_0}{\epsilon_0}\right)^{\frac{1}{2}}.$$
 (B.2.43)

It would be a digression to pursue this point here, but (B.2.43) is the condition that must be satisfied if an electromagnetic wave launched between the plates at x = -l is to be absorbed, without reflection, by the resistive sheet*; that is, the intermediate case is one in which all the power fed into the system, regardless of the frequency or time constant, is dissipated by the resistive sheet.

B.3 MACROSCOPIC MODELS AND CONSTITUENT RELATIONS

When solids, liquids, and gases are placed in electromagnetic fields, they influence the field distribution. This is another way of saying that the force of interaction between charges or between currents is influenced by the presence of media. The effect is not surprising because the materials are comprised of charged particles.

Problems of physical significance can usually be decomposed into parts with widely differing scales. At the molecular or submolecular level we may be concerned with the dynamics of individual charges or of the atoms or molecules to which they are attached. These systems tend to have extremely small dimensions when compared with the size of a physical device. On the macroscopic scale we are not interested in the detailed behavior of the microscopic constituents of a material but rather only a knowledge of the average behavior of variables, since only these averages are observable on a macroscopic scale. The charge and current densities introduced in Section B.1 are examples of such variables, hence it is a macroscopic picture of fields and media that we require here.

There are three major ways in which media influence macroscopic electromagnetic fields. Hence the following sections undertake a review of magnetization, polarization, and conduction in common materials.

B.3.1 Magnetization

The macroscopic motions of electrons, even though associated with individual atoms or molecules, account for aggregates of charge and current

^{*} The propagation of an electromagnetic wave on structures of this type is discussed in texts concerned with transmission lines or TEM wave guide modes. For a discussion of this matching problem see R. B. Adler, L. J. Chu, and R. M. Fano, *Electromagnetic Energy Transmission and Radiation*, Wiley, New York, 1960, p. 111, or S. Ramo, J. R. Whinnery, and T. Van Duzer, *Fields and Waves in Communication Electronics*, Wiley, New York, p. 27.

Review of Electromagnetic Theory

(when viewed at the macroscopic level) that induce electric and magnetic fields. These field sources are not directly accessible; for example, the equivalent currents within the material cannot be circulated through an external circuit. The most obvious sources of magnetic field that are inaccessible in this sense are those responsible for the field of a permanent magnet. The earliest observations on magnetic fields involved the lodestone, a primitive form of the permanent magnet. Early investigators such as Oersted found that magnetic fields produced by a permanent magnet are equivalent to those induced by a circulating current. In the formulation of electromagnetic theory we must distinguish between fields due to sources within the material and those from applied currents simply because it is only the latter sources that can be controlled directly. Hence we divide the source currents into *free currents* (with the density J_{r}) and *magnetization currents* (with the density J_m). Ampère's law then takes the form

$$\nabla \times \left(\frac{\mathbf{B}}{\mu_0}\right) = \mathbf{J}_m + \mathbf{J}_f.$$
 (B.3.1)

By convention it is also helpful to attribute a fraction of the field induced by these currents to the magnetization currents in the material. Hence (B.3.1) is written as

$$\nabla \times \left(\frac{\mathbf{B}}{\mu_0} - \mathbf{M}\right) = \mathbf{J}_f,$$
 (B.3.2)

where the magnetization density M is defined by

$$\nabla \times \mathbf{M} = \mathbf{J}_m. \tag{B.3.3}$$

Up to this point in this chapter it has been necessary to introduce only two field quantities to account for interactions between charges and between currents. To account for the macroscopic properties of media we have now introduced a new field quantity, the magnetization density **M**, and in the next section similar considerations concerning electric polarization of media lead to the introduction of the polarization density **P**. It is therefore apparent that macroscopic field theory is formulated in terms of four field variables. In our discussion these variables have been **E**, **B**, **M**, and **P**. An alternative representation of the fields introduces the *magnetic field intensity* **H**, in our development *defined* as

$$\mathbf{H} = \left(\frac{\mathbf{B}}{\mu_0} - \mathbf{M}\right). \tag{B.3.4}$$

From our definition it is clear that we could just as well deal with B and H as the macroscopic magnetic field vectors rather than with B and M. This is

particularly appealing, for then (B.3.2) takes the simple form

$$\nabla \times \mathbf{H} = \mathbf{J}_{f}.\tag{B.3.5}$$

When the source quantities J_f and M are specified independently, the magnetic field intensity H (or magnetic flux density B) can be found from the quasi-static magnetic field equations. A given constant magnetization density corresponds to the case of the permanent magnet. In most cases, however, the source quantities are functions of the field vectors, and these functional relations, called *constituent relations*, must be known before the problems can be solved. The constituent relations represent the constraints placed on the fields by the internal physics of the media being considered. Hence it is these relations that make it possible to separate the microscopic problem from the macroscopic one of interest here.

The simplest form of constituent relation for a magnetic material arises when it can be considered *electrically linear* and *isotropic*. Then the *permeability* μ is constant in the relation

$$\mathbf{B} = \mu \mathbf{H}.\tag{B.3.6}$$

The material is isotropic because **B** is collinear with **H** and a particular constant (μ) times **H**, regardless of the direction of **H**. A material that is *homogeneous* and isotropic will in addition have a permeability μ that does not vary with position in the material. Another way of expressing (B.3.6) is to define a magnetic susceptibility χ_m (dimensionless) such that

$$\mathbf{M} = \boldsymbol{\chi}_m \mathbf{H}, \tag{B.3.7}$$

where

$$\mu = \mu_0 (1 + \chi_m). \tag{B.3.8}$$

Magnetic materials are commonly found with B not a linear function of H and the constitutive law takes the general form

$$\mathbf{B} = \mathbf{B}(\mathbf{H}). \tag{B.3.9}$$

We deal with some problems involving materials of this type, but with few exceptions confine our examples to situations in which **B** is a single-valued function of **H**. In certain magnetic materials in some applications the **B**-H curve must include hysteresis and (B.3.9) is not single-valued.*

The differential equations for a magnetic field system in the presence of moving magnetized media are summarized in Table 1.2.

B.3.2 Polarization

The force between a charge distribution and a test charge is observed to change if a dielectric material is brought near the region occupied by the test

* G. R. Slemon, Magnetoelectric Devices, Wiley, New York, 1966, p. 115.
Review of Electromagnetic Theory

charge. Like the test charge, the charged particles which compose the dielectric material experience forces due to the applied field. Although these charges remain identified with the molecules of the material, their positions can be distorted incrementally by the electric force and thus lead to a polarization of the molecules.

The basic sources of the electric field are charges. Hence it is natural to define a *polarization charge density* ρ_p as a source of a fraction of the electric field which can be attributed to the inaccessible sources within the media. Thus Gauss's law (B.1.16) is written

$$\nabla \cdot \epsilon_0 \mathbf{E} = \rho_f + \rho_p, \tag{B.3.10}$$

where the *free charge density* ρ_f resides on conducting electrodes and other parts of the system capable of supporting conduction currents. The free charges do not remain attached to individual molecules but rather can be conducted from one point to another in the system.

In view of the form taken by Gauss's law, it is convenient to identify a field induced by the polarization charges by writing (B.3.10) as

$$\nabla \cdot (\epsilon_0 \mathbf{E} + \mathbf{P}) = \rho_f, \tag{B.3.11}$$

where the *polarization density* \mathbf{P} is related to the polarization charge density by

$$\rho_p = -\nabla \cdot \mathbf{P}. \tag{B.3.12}$$

As in Section B.3.1, it is convenient to define a new vector field that serves as an alternative to \mathbf{P} in formulating the electrodynamics of polarized media. This is the *electric displacement* \mathbf{D} , defined as

$$\mathbf{D} = \epsilon_0 \mathbf{E} + \mathbf{P} \tag{B.3.13}$$

In terms of this field, Gauss's law for electric fields (B.3.11) becomes

$$\nabla \cdot \mathbf{D} = \rho_f. \tag{B.3.14}$$

The simple form of this expression makes it desirable to use **D** rather than **P** in the formulation of problems.

If a polarization charge model is to be used to account for the effects of polarizable media on electric fields, we must recognize that the motion of these charges can lead to a current. In fact, now that two classes of charge density have been identified we must distinguish between two classes of current density. The free current density J_f accounts for the conservation of free charge so that (B.1.18) can be written as

$$\nabla \cdot \mathbf{J}_f + \frac{\partial \rho_f}{\partial t} = 0. \tag{B.3.15}$$

Appendix **B**

In view of (B.3.11), this expression becomes

$$\nabla \cdot \mathbf{J}_f + \frac{\partial}{\partial t} \nabla \cdot (\epsilon_0 \mathbf{E} + \mathbf{P}) = 0.$$
 (B.3.16)

Now, if we write Ampère's law (B.2.26b) as

$$\nabla \times \left(\frac{\mathbf{B}}{\mu_0}\right) = \mathbf{J}_f + \mathbf{J}_p + \frac{\partial}{\partial t} \epsilon_0 \mathbf{E},$$
 (B.3.17)

where J_p is a current density due to the motion of polarization charges, the divergence of (B.3.17) must give (B.3.16). Therefore

$$\nabla \cdot \mathbf{J}_{p} + \frac{\partial}{\partial t} \left(-\nabla \cdot \mathbf{P} \right) = 0. \tag{B.3.18}$$

which from (B.3.12) is an expression for the conservation of polarization charge. This expression does not fully determine the polarization current density J_{p} , because in general we could write

$$\mathbf{J}_{p} = \frac{\partial \mathbf{P}}{\partial t} + \boldsymbol{\nabla} \times \mathbf{A}, \qquad (B.3.19)$$

where A is an arbitrary vector, and still satisfy (B.3.18). At this point we could derive the quantity A (which would turn out to be $P \times v$, where v is the velocity of the polarized medium). It is important, however, to recognize that this represents an unnecessary digression. In the electric field system the magnetic field appears in only one of the equations of motion—Ampère's law. It does not appear in (B.2.27b) to (B.2.29b), nor will it appear in any constitutive law used in this book. For this reason the magnetic field serves simply as a quantity to be calculated once the electromechanical problem has been solved. We might just as well lump the quantity A with the magnetic field in writing Ampère's law. In fact, if we are consistent, the magnetic field intensity H can be defined as given by

$$\nabla \times \mathbf{H} = \mathbf{J}_f + \frac{\partial \mathbf{D}}{\partial t},$$
 (B.3.20)

with no loss of physical significance. In an electric field system the magnetic field is an alternative representation of the current density J_f . A review of the quasi-static solutions for the system in Fig. B.2.1*b* illustrates this point.

In some materials (ferroelectrics) the polarization density P is constant. In most common dielectrics, however, the polarization density is a function of E. The simplest constituent relation for a dielectric is that of linear and isotropic material,

$$\mathbf{P} = \epsilon_0 \boldsymbol{\lambda}_e \mathbf{E}, \tag{B.3.21}$$

where χ_e is the *dielectric susceptibility* (dimensionless) that may be a function of space but not of E. For such a material we define the *permittivity* ϵ as

$$\epsilon = \epsilon_0 (1 + \chi_e). \tag{B.3.22}$$

and then write the relation between **D** and **E** as [see (B.3.13)]

$$\mathbf{D} = \epsilon \mathbf{E}. \tag{B.3.23}$$

This mathematical model of polarizable material is used extensively in this book.

The differential equations for the electric field system, in the presence of moving polarized media, are summarized in Table 1.2.

B.3.3 Electrical Conduction

In both magnetic and electric field systems the conduction process accounts for the free current density J_f in a fixed conductor. The most common model for this process is appropriate in the case of an isotropic, linear, conducting medium which, when stationary, has the constituent relation (often called *Ohm's law*)

$$\mathbf{J}_f = \sigma \mathbf{E}.\tag{B.3.24}$$

Although (B.3.24) is the most widely used mathematical model of the conduction process, there are important electromechanical systems for which it is not adequate. This becomes apparent if we attempt to derive (B.3.24), an exercise that will contribute to our physical understanding of Ohm's law.

In many materials the conduction process involves two types of charge carrier (say, ions and electrons). As discussed in Section B.1.2, a macroscopic model for this case would recognize the existence of free charge densities ρ_+ and ρ_- with charge average velocities \mathbf{v}_+ and \mathbf{v}_- , respectively. Then

$$\mathbf{J}_f = \rho_+ \mathbf{v}_+ + \rho_- \mathbf{v}_-. \tag{B.3.25}$$

The problem of relating the free current density to the electric field intensity is thus a problem in electromechanics in which the velocities of the particles carrying the free charge must be related to the electric fields that apply forces to the charges.

The charge carriers have finite mass and thus accelerate when subjected to a force. In this case there are forces on the positive and negative carriers, respectively, given by (B.1.10) (here we assume that effects from a magnetic field are ignorable):

$$\mathbf{F}_{+} = \rho_{+}\mathbf{E}, \qquad (B.3.26)$$

$$\mathbf{F}_{-} = \rho_{-}\mathbf{E}.\tag{B.3.27}$$

Appendix B

As the charge carriers move, their motion is retarded by collisions with other particles. On a macroscopic basis the retarding force of collisions can be thought of as a viscous damping force that is proportional to velocity. Hence we can picture the conduction process in two extremes. With no collisions between particles the electric force densities of (B.3.26 and B.3.27) continually accelerate the charges, for the only retarding forces are due to acceleration expressed by Newton's law. In the opposite extreme a charge carrier suffers collisions with other particles so frequently that its average velocity quickly reaches a limiting value, which in view of (B.3.26 and B.3.27) is proportional to the applied electric field. It is in this second limiting case that Ohm's law assumes physical significance. By convention *mobilities* μ_+ and μ_- which relate these limiting velocities to the field **E** are defined

$$\mathbf{v}_{+} = \boldsymbol{\mu}_{+} \mathbf{E}, \tag{B.3.28}$$

$$\mathbf{v}_{-} = \mu_{-} \mathbf{E}.$$
 (B.3.29)

In terms of these quantities, (B.3.25) becomes

$$\mathbf{J}_{f} = (\rho_{+}\mu_{+} + \rho_{-}\mu_{-})\mathbf{E}. \tag{B.3.30}$$

It is important to recognize that it is only when the collisions between carriers and other particles dominate the accelerating effect of the electric field that the conduction current takes on a form in which it is dependent on the instantaneous value of E. Fortunately, (B.3.30) is valid in a wide range of physical situations. In fact, in a metallic conductor the number of charge carriers is extremely high and very nearly independent of the applied electric field. The current carriers in most metals are the electrons, which are detached from atoms held in the lattice structure of the solid. Therefore the negatively charged electrons move in a background field of positive charge and, to a good approximation, $\rho_{+} = -\rho_{-}$. Then (B.3.30) becomes

$$\mathbf{J} = \sigma \mathbf{E}, \tag{B.3.31}$$

where the conductivity is defined as

$$\rho_+(\mu_+ - \mu_-).$$
 (B.3.32)

The usefulness of the conductivity as a parameter stems from the fact that both the number of charges available for conduction and the net mobility (essentially that of the electrons) are constant. This makes the conductivity essentially independent of the electric field, as assumed in (B.3.24).*

^{*} We assume here that the temperature remains constant. A worthwhile qualitative description of conduction processes in solids is given in J. M. Ham and G. R. Slemon, *Scientific Basis of Electrical Engineering*, Wiley, New York, 1961, p. 453.

In some types of material (notably slightly ionized gases) which behave like insulators, the conduction process cannot be described simply by Ohm's law. In such materials the densities of charge carriers and even the mobilities may vary strongly with electric field intensity.

B.4 INTEGRAL LAWS

The extensive use of circuit theory bears testimony to the usefulness of the laws of electricity and magnetism in integral form. Physical situations that would be far too involved to describe profitably in terms of field theory have a lucid and convenient representation in terms of circuits. Conventional circuit elements are deduced from the integral forms of the field equations. The description of lumped-parameter electromechanical systems, as undertaken in Chapter 2, requires that we generalize the integral laws to include time-varying surfaces and contours of integration. Hence it is natural that we conclude this appendix with a discussion of the integral laws.

B.4.1 Magnetic Field Systems

Faraday's law of induction, as given by (B.1.42), has the differential form

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}.$$
 (B.4.1)

This expression can be integrated over a surface S enclosed by the contour C. Then, according to Stokes's theorem,

$$\oint_{C} \mathbf{E} \cdot d\mathbf{l} = -\int_{S} \frac{\partial \mathbf{B}}{\partial t} \cdot \mathbf{n} \, da. \tag{B.4.2}$$

Now, if S and C are fixed in space, the time derivative on the right can be taken either before or after the surface integral of $\mathbf{B} \cdot \mathbf{n}$ is evaluated. Note that $\int_{S} \mathbf{B} \cdot \mathbf{n} \, da$ is only a function of time. For this reason (B.1.41) could be written with the total derivative outside the surface integral. It is implied in the integral equation (B.1.41) that S is fixed in space.

Figure B.4.1 shows an example in which it is desirable to be able to use (B.4.2), with S and C varying in position as a function of time. The contour C is a rectangular loop that encloses a surface S which makes an angle $\theta(t)$ with the horizontal. Although the induction law is not limited to this case, the loop could comprise a one-turn coil, in which case it is desirable to be able to use (B.4.2) with C fixed to the coil. The integral law of induction would be much more useful if it could be written in the form

$$\oint_C \mathbf{E}' \cdot d\mathbf{l} = -\frac{d}{dt} \int_S \mathbf{B} \cdot \mathbf{n} \, da. \tag{B.4.3}$$



Fig. B.4.1 Contour C enclosing a surface S which varies as a function of time. The rectangular loop links no magnetic flux when $\theta = 0, \pi, \ldots$

In this form the quantity on the right is the negative time rate of change of the flux linked by the contour C, whereas E' is interpreted as the electric field measured in the moving frame of the loop. An integral law of induction in the form of (B.4.3) is essential to the lumped-parameter description of magnetic field systems. At this point we have two choices. We can accept (B.4.3) as an empirical result of Faraday's original investigations or we can show mathematically that (B.4.2) takes the form of (B.4.3) if

$$\mathbf{E}' = \mathbf{E} + \mathbf{v} \times \mathbf{B},\tag{B.4.4}$$

where \mathbf{v} is the velocity of $d\mathbf{l}$ at the point of integration. In any case this topic is pursued in Chapter 6 to clarify the significance of electric and magnetic fields measured in different frames of reference.

The mathematical connection between (B.4.2) and (B.4.3) is made by using the integral theorem

$$\frac{d}{dt} \int_{S} \mathbf{A} \cdot \mathbf{n} \, da = \int_{S} \left[\frac{\partial \mathbf{A}}{\partial t} + (\nabla \cdot \mathbf{A}) \mathbf{v} \right] \cdot \mathbf{n} \, da + \oint_{C} (\mathbf{A} \times \mathbf{v}) \cdot d\mathbf{I}, \quad (B.4.5)$$

where v is the velocity of S and C and in the case of (B.4.3), $A \rightarrow B$. Before we embark on a proof of this theorem, an example will clarify its significance.

Example B.4.1. The coil shown in Fig. B.4.1 rotates with the angular deflection $\theta(t)$ in a uniform magnetic flux density $\mathbf{B}(t)$, directed as shown. We wish to compute the rate of change of the flux linked by the coil in two ways: first by computing $\int_{S} \mathbf{B} \cdot \mathbf{n} \, da$ and taking

Review of Electromagnetic Theory

its derivative [the left-hand side of (B.4.5)], then by using the surface and contour integrations indicated on the right-hand side of (B.4.5). This illustrates how the identity allows us to carry out the surface integration before rather than after the time derivative is taken. From Fig. B.4.1 we observe that

$$\int_{S} \mathbf{B} \cdot \mathbf{n} \, da = -B_0(t) 2ad \sin \theta, \qquad (a)$$

so that the first calculation gives

$$\frac{d}{dt} \int_{S} \mathbf{B} \cdot \mathbf{n} \, da = -2ad \sin \theta \, \frac{dB_0}{dt} - B_0 2ad \cos \theta \frac{d\theta}{dt} \,. \tag{b}$$

To evaluate the right-hand side of (B.4.5) observe that $\nabla \cdot \mathbf{B} = 0$ and [from (a)]

$$\int_{S} \frac{\partial \mathbf{B}}{\partial t} \cdot \mathbf{n} \, da = -2ad \sin \theta \, \frac{dB_0}{dt}.$$
 (c)

The quantity $\mathbf{B} \times \mathbf{v}$ is collinear with the axis of rotation in Fig. B.4.1; hence there is no contribution to the line integral along the pivoted ends of the loop. Because both the velocity $\mathbf{v} = \mathbf{i}_{\theta} a (d\theta/dt)$ and line elements $d\mathbf{l}$ are reversed in going from the upper to the lower horizontal contours, the line integral reduces to twice the value from the upper contour.

$$\oint_C \mathbf{B} \times \mathbf{v} \cdot d\mathbf{I} = -2B_0 a d \cos \theta \frac{d\theta}{dt}$$
(d)

From (c) and (d) it follows that the right-hand side of (B.4.5) also gives (b). Thus, at least for this example, (B.4.5) provides alternative ways of evaluating the time rate of change of the flux linked by the contour C.

The integral theorem of (B.4.5) can be derived by considering the deforming surface S shown at two instants of time in Fig. B.4.2. In the incremental time interval Δt the surface S moves from S_1 to S_2 , and therefore by



Fig. B.4.2 When t = t, the surface S enclosed by the contour C is as indicated by S_1 and C_1 . By the time $t = t + \Delta t$ this surface has moved to S_2 , where it is enclosed by the contour C_2 .

definition

$$\frac{d}{dt}\int_{S} \mathbf{A} \cdot \mathbf{n} \, da = \lim_{\Delta t \to 0} \frac{1}{\Delta t} \left(\int_{S_2} \mathbf{A} \left| \int_{t+\Delta t} \cdot \mathbf{n} \, da - \int_{S_1} \mathbf{A} \left| \int_t \cdot \mathbf{n} \, da \right). \quad (B.4.6)$$

Here we have been careful to show that when the integral on S_2 is evaluated $t = t + \Delta t$, in contrast to the integration on S_1 , which is carried out when t = t.

The expression on the right in (B.4.6) can be evaluated at a given instant in time by using the divergence theorem (B.1.14) to write

$$\int_{V} \nabla \cdot \mathbf{A} \, dV \simeq \int_{S^2} \mathbf{A} \Big|_t \cdot \mathbf{n} \, da - \int_{S_1} \mathbf{A} \Big|_t \cdot \mathbf{n} \, da - \Delta t \oint_{C_1} \mathbf{A} \cdot \mathbf{v} \times d\mathbf{l} \quad (B.4.7)$$

for the volume V traced out by the surface S in the time Δt . Here we have used the fact that $-\mathbf{v} \Delta t \times d\mathbf{l}$ is equivalent to a surface element **n** da on the surface traced out by the contour C in going from C_1 to C_2 in Fig. B.4.2. To use (B.4.7) we make three observations. First, as $\Delta t \rightarrow 0$,

$$\int_{S_2} \mathbf{A} \bigg|_{t+\Delta t} \cdot \mathbf{n} \, da \simeq \int_{S_2} \mathbf{A} \bigg|_t \cdot \mathbf{n} \, da + \int_{S_1} \frac{\partial \mathbf{A}}{\partial t} \bigg|_t \Delta t \cdot \mathbf{n} \, da + \cdots \quad (B.4.8)$$

Second, it is a vector identity that

$$\mathbf{A} \cdot \mathbf{v} \times d\mathbf{l} = \mathbf{A} \times \mathbf{v} \cdot d\mathbf{l}. \tag{B.4.9}$$

Third, an incremental volume dV swept out by the surface da is essentially the base times the perpendicular height or

$$dV = \Delta t \mathbf{v} \cdot \mathbf{n} \, da. \tag{B.4.10}$$

From these observations (B.4.7) becomes

$$\Delta t \int_{S_1} (\nabla \cdot \mathbf{A}) \mathbf{v} \cdot \mathbf{n} \, da \simeq \int_{S_2} \mathbf{A} \bigg|_{t+\Delta t} \cdot \mathbf{n} \, da - \int_{S_1} \Delta t \, \frac{\partial \mathbf{A}}{\partial t} \bigg|_t \cdot \mathbf{n} \, da$$
$$- \int_{S_1} \mathbf{A} \bigg|_t \cdot \mathbf{n} \, da - \Delta t \oint_{C_1} \mathbf{A} \times \mathbf{v} \cdot d\mathbf{l}. \quad (B.4.11)$$

This expression can be solved for the quantity on the right in (B.4.6) to give

$$\frac{d}{dt} \int_{S} \mathbf{A} \cdot \mathbf{n} \, da = \lim_{\mathbf{A}t \to 0} \left\{ \int_{S_1} \left[(\nabla \cdot \mathbf{A}) \mathbf{v} + \frac{\partial \mathbf{A}}{\partial t} \right] \cdot \mathbf{n} \, da + \oint_{C_1} \mathbf{A} \times \mathbf{v} \cdot d\mathbf{l} \right\}.$$
(B.4.12)

The limit of this expression produces the required relation (B.4.5).

Use of (B.4.5) to express the right-hand side of (B.4.2) results in

$$\int_{S} \frac{\partial \mathbf{B}}{\partial t} \cdot \mathbf{n} \, da = \frac{d}{dt} \int_{S} \mathbf{B} \cdot \mathbf{n} \, da - \int_{S} (\nabla \cdot \mathbf{B}) \mathbf{v} \cdot \mathbf{n} \, da - \oint_{C} (\mathbf{B} \times \mathbf{v}) \cdot d\mathbf{l}.$$
(B.4.13)

Because $\nabla \cdot \mathbf{B} = 0$, (B.4.2) then reduces to (B.4.3), with E' given by (B.4.4).

The integral laws for the magnetic field system are summarized in Table 1.2 at the end of Chapter 1. In these equations surfaces and contours of integration can, in general, be time-varying.

B.4.2 Electric Field System

Although the integral form of Faraday's law can be taken as an empirical fact, we require (B.4.5) to write Ampère's law in integral form for an electric field system. If we integrate (B.3.20) over a surface S enclosed by a contour C, by Stokes's theorem it becomes

$$\oint_C \mathbf{H} \cdot d\mathbf{l} = \int_S \mathbf{J}_f \cdot \mathbf{n} \, da \, + \int_S \frac{\partial \mathbf{D}}{\partial t} \cdot \mathbf{n} \, da. \tag{B.4.14}$$

As with the induction law for the magnetic field system, this expression can be generalized to include the possibility of a deforming surface S by using (B.4.13) with $\mathbf{B} \rightarrow \mathbf{D}$ to rewrite the last term. If, in addition, we use (B.3.14) to replace $\nabla \cdot \mathbf{D}$ with ρ_{t} , (B.4.14) becomes

$$\oint_C \mathbf{H}' \cdot d\mathbf{l} = \int_S \mathbf{J}'_f \cdot \mathbf{n} \, da + \frac{d}{dt} \int_S \mathbf{D} \cdot \mathbf{n} \, da, \qquad (B.4.15)$$

where

$$\mathbf{H}' = \mathbf{H} - \mathbf{v} \times \mathbf{D}, \tag{B.4.16}$$

$$\mathbf{J}_f' = \mathbf{J}_f - \rho_f \mathbf{v}. \tag{B.4.17}$$

The fields \mathbf{H}' and \mathbf{J}'_{f} can be interpreted as the magnetic field intensity and free current density measured in the moving frame of the deforming contour. The significance of these field transformations is discussed in Chapter 6. Certainly the relationship between \mathbf{J}'_{f} (the current density in a frame moving with a velocity \mathbf{v}) and the current density \mathbf{J}_{f} (measured in a fixed frame), as given by (B.4.17), is physically reasonable. The free charge density appears as a current in the negative \mathbf{v} -direction when viewed from a frame moving at the velocity \mathbf{v} . If was reasoning of this kind that led to (B.1.25).

As we have emphasized, it is the divergence of Ampère's differential law that assumes the greatest importance in electric field systems, for it accounts for conservation of charge. The integral form of the conservation of charge



Fig. B.4.3 The sum of two surfaces S_1 and S_2 "spliced" together at the contour to enclose the volume V.

equation, including the possibility of a deforming surface of integration, is obtained by using (B.4.15). For this purpose integrations are considered over two deforming surfaces, S_1 and S_2 , as shown in Fig. B.4.3. These surfaces are chosen so that they are enclosed by the same contour C. Hence, taken together, S_1 and S_2 enclose a volume V.

Integration of (B.4.15) over each surface gives

$$\oint_C \mathbf{H}' \cdot d\mathbf{l}_1 = \int_{S_1} \mathbf{J}'_f \cdot \mathbf{n} \, da + \frac{d}{dt} \int_{S_1} \mathbf{D} \cdot \mathbf{n} \, da. \tag{B.4.18}$$

$$\oint_{C} \mathbf{H}' \cdot d\mathbf{l}_{2} = \int_{S_{2}} \mathbf{J}'_{f} \cdot \mathbf{n} \, da + \frac{d}{dt} \int_{S_{2}} \mathbf{D} \cdot \mathbf{n} \, da. \tag{B.4.19}$$

Now, if **n** is defined so that it is directed out of the volume V on each surface, the line integral enclosing S_1 will be the negative of that enclosing S_2 . Then the sum of (B.4.18 and B.4.19) gives the desired integral form of the conservation of charge equation:

$$\oint_{S} \mathbf{J}'_{f} \cdot \mathbf{n} \, da \, + \frac{d}{dt} \int_{V} \rho_{f} \, dV = 0. \tag{B.4.20}$$

In writing this expression we have used Gauss's theorem and (B.3.14) to show the explicit dependence of the current density through the deforming surface on the enclosed charge density.

The integral laws for electric field systems are summarized in Table 1.2 at the end of Chapter 1.

B.5 RECOMMENDED READING

The following texts treat the subject of electrodynamics and provide a comprehensive development of the fundamental laws of electricity and magnetism.

R. M. Fano, L. J. Chu, and R. B. Adler, *ElectromagneticFields, Energy, and Forces*, Wiley, New York, 1960; J. D. Jackson, *Classical Electrodynamics*, Wiley, New York, 1962: S. Ramo, J. R. Whinnery, and T. Van Duzer, *Fields and Waves in Communication Electronics*, Wiley, New York, 1965; W. K. H. Panofsky and M. Phillips, *Classical Electricity and Magnetism*, Addison-Wesley, Reading, Mass., 1956; J. A. Stratton, *Electromagnetic Theory*, McGraw-Hill, New York, 1941.

Many questions arise in the study of the effects of moving media on electric and magnetic fields concerning the macroscopic representation of polarized and magnetized media; for example, in this appendix we introduced the fields **E** and **B** as the quantities defined by the force law. Then **P** and **M** (or **D** and **H**) were introduced to account for the effects of polarization and magnetization. Hence the effect of the medium was accounted for by equivalent polarization charges ρ_{p} and magnetization currents J_{m} . Other representations can be used in which a different pair of fundamental vectors is taken, as defined by the force law (say, **E** and **H**), and in which the effects of media are accounted for by an equivalent magnetic charge instead of an equivalent current. If we are consistent in using the alternative formulations of the field equations, they predict the same physical results, including the force on magnetized and polarized media. For a complete discussion of these matters see **P**. Penfield, and **H**. Haus, *Electrodynamics of Moving Media*, M.I.T. Press, Cambridge, Mass., 1967.

Appendix C

SUMMARY OF PARTS I AND II AND USEFUL THEOREMS

IDENTITIES $\mathbf{A} \times \mathbf{B} \cdot \mathbf{C} = \mathbf{A} \cdot \mathbf{B} \times \mathbf{C},$ $\mathbf{A} \times (\mathbf{B} \times \mathbf{C}) = \mathbf{B}(\mathbf{A} \cdot \mathbf{C}) - \mathbf{C}(\mathbf{A} \cdot \mathbf{B})$ $\nabla(\phi + \psi) = \nabla\phi + \nabla\psi,$ $\nabla \cdot (\mathbf{A} + \mathbf{B}) = \nabla \cdot \mathbf{A} + \nabla \cdot \mathbf{B},$ $\nabla \times (\mathbf{A} + \mathbf{B}) = \nabla \times \mathbf{A} + \nabla \times \mathbf{B},$ $\nabla(\phi\psi) = \phi \,\nabla\psi + \psi \,\nabla\phi,$ $\nabla \cdot (\psi \mathbf{A}) = \mathbf{A} \cdot \nabla \psi + \psi \nabla \cdot \mathbf{A},$ $\nabla \cdot (\mathbf{A} \times \mathbf{B}) = \mathbf{B} \cdot \nabla \times \mathbf{A} - \mathbf{A} \cdot \nabla \times \mathbf{B},$ $\nabla \cdot \nabla \phi = \nabla^2 \phi,$ $\nabla \cdot \nabla \times \mathbf{A} = 0.$ $\nabla \times \nabla \phi = 0$, $\nabla \times (\nabla \times A) = \nabla (\nabla \cdot A) - \nabla^2 A,$ $(\nabla \times \mathbf{A}) \times \mathbf{A} = (\mathbf{A} \cdot \nabla)\mathbf{A} - \frac{1}{2}\nabla(\mathbf{A} \cdot \mathbf{A}),$ $\nabla (\mathbf{A} \cdot \mathbf{B}) = (\mathbf{A} \cdot \nabla)\mathbf{B} + (\mathbf{B} \cdot \nabla)\mathbf{A} + \mathbf{A} \times (\nabla \times \mathbf{B}) + \mathbf{B} \times (\nabla \times \mathbf{A})$ $\nabla \times (\phi \mathbf{A}) = \nabla \phi \times \mathbf{A} + \phi \nabla \times \mathbf{A},$ $\nabla \times (\mathbf{A} \times \mathbf{B}) = \mathbf{A}(\nabla \cdot \mathbf{B}) - \mathbf{B}(\nabla \cdot \mathbf{A}) + (\mathbf{B} \cdot \nabla)\mathbf{A} - (\mathbf{A} \cdot \nabla)\mathbf{B}.$

THEOREMS



C2

Appendix \mathbf{D}

GLOSSARY OF COMMONLY USED SYMBOLS

Section references indicate where symbols of a given significance are introduced; grouped symbols are accompanied by their respective references. The absence of a section reference indicates that a symbol has been applied for a variety of purposes. Nomenclature used in examples is not included.

Symbol	Symbol Meaning	
A	cross-sectional area	
A_i	coefficient in differential equation	5.1.1
(A_{n}^{+}, A_{n}^{-})	complex amplitudes of components of <i>n</i> th	
	mode	9.2.1
A_w	cross-sectional area of armature conductor	6.4.1
a	spacing of pole faces in magnetic circuit	8.5.1
$a, (a_c, a_s)$	phase velocity of acoustic related waves	13.2.1, 11.4.1
a _b	Alfvén velocity	12.2.3
(a, b, c)	Lagrangian coordinates	11.1
a_i	constant coefficient in differential equation	5.1.1
\mathbf{a}_p	instantaneous acceleration of point p fixed	2210
B, B_r, B_s	damping constant for linear, angular and	2.2.10
	square law dampers	2.2.1b, 4.1.1, 5.2.2
$\mathbf{B}, \mathbf{B}_i, B_0$	magnetic flux density	1.1.1a, 8.1, 6.4.2
B_i	induced flux density	7.0
$(B_r, B_{ra}, B_{rb}, B_{rm})$	radial components of air-gap flux	
	densities	4.1.4
$[B_{rf}, (B_{rf})_{\rm av}]$	radial flux density due to field current	6.4.1
b	width of pole faces in magnetic circuit	8.5
b	half thickness of thin beam	11.4.2b
С	contour of integration	1.1.2a
$C, (C_a, C_b), C_o$	capacitance	2.1.2, 7.2.1a, 5.2.1
С	coefficient in boundary condition	9.1.1
C	the curl of the displacement	11.4
(<i>C</i> ⁺ , <i>C</i> ⁻)	designation of characteristic lines	9.1.1

Symbol	Meaning	Section	
	specific heat capacity at constant pressure	13.1.2	
c_v	specific heat capacity at constant volume	13.1.2	
D	electric displacement	1.1.1a	
d	length		
da	elemental area	1.1.2a	
$d\mathbf{f}_n$	total elemental force on material in rigid body	2.2.1c	
dl	elemental line segment	1.1.2a	
$d\mathbf{T}_n$	torque on elemental volume of material	2.2.1c	
dV	elemental volume	1.1.2b	
Ε	constant of motion	5.2,1	
Ε	Young's modulus or the modulus of elasticity	9.1	
\mathbf{E}, E_{α}	electric field intensity	1.1.1a, 5.1.2d	
E_f	magnitude of armature voltage generated by field current in a synchronous		
r -	machine	4.1.6a	
E_i	induced electric field intensity	7.0	
e_{11}, e_{ij}	strain tensor	9.1, 11.2	
e_{ij}	stram-rate tensor	14.1.1a	
r F	force density	13.2.2	
r r	force density	1.1.1a	
F	complex amplitude of $f(t)$	5.1.1	
F_0	amplitude of sinusoidal driving force	9.1.3	
f	equilibrium tension of string	9.2	
f c c ce ce c c c	driving function	5.1.1	
$f, \mathbf{i}, f^{e}, f^{s}, f_{j}, f_{i}, f_{1}$	Iorce	2.2.1, 2.2.1c, 3.1, 5.1.2a, 3.1.2b, 8.1, 9.1	
f	arbitrary scalar function	6.1	
f'	scalar function in moving coordinate system	6.1	
f	three-dimensional surface	6.2	
f	integration constant	11.4.2a	
Ġ	a constant	5.1.2c	
G	shear modulus of elasticity	11.2.2	
G	speed coefficient	6.4.1	
G	conductance	3.1	
8	air-gap length	5.2.1	
g, g	acceleration of gravity	5.1.2c, 12.1.3	
$(\mathbf{H}, H_x, H_y, H_z)$	magnetic field intensity	1.1.1a	
h	specific enthalpy	13.1.2	
$\mathbf{I}, I, (I_r, I_s), I_f$	electrical current	10.4.3, 12.2.1a, 4.1.2, 6.4.1	
$(i, i_1, i_2, \dots, i_k), (i_{ar}, i_{as}, i_{br}, i_{bs}), i_a, (i_a, i_b, i_c), (i_f, i_t), (i_r, i_s)$	electrical current	2.1, 4.1.3, 6.4.1, 4.1.7, 6.4.1, 4.1	

Appendix 1	D
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Symbol	Section	
i _n	unit vector perpendicular to area of	
	integration	6.2.1
i _s	unit vector normal to surface of	
	integration	6.2.1
$(i_x, i_y, i_z), (i_1, i_2, i_3)$	unit vectors in coordinate directions	2.2.1c
J, J_f	current density	7.0, 1.1.1a
$J, J_r, (J_x, J_y, J_z)$	moment of inertia	5.1.2b, 4.1.1, 2.2.1c
J_{xz}, J_{yz}	products of inertia	2.2.1c
1	$\sqrt{-1}$	4.1.6a
, K	loading factor	13.2.2
 K. K.	surface current density	7.0, 1.1.1a
K	linear or torsional spring constant	2.2.1a
 K.	induced surface current density	7.0
$k_{1}k_{2}(k_{2},k_{3})$	wavenumber	7.1.3. 10.1.3. 10.0
k	summation index	211
k	maximum coefficient of coupling	4 1 6b
k.	nth eigenvalue	9.2
(L, L, L) (L, L)	inductance	211 641 211
$(\underline{L}_{1}, \underline{L}_{2}), (\underline{L}_{a}, \underline{L}_{j}), (\underline{L}_{a}, \underline{L}_{j}), (\underline{L}_{a}, \underline{L}_{a}), (\underline{L}_{a}), (\underline{L}_{a}, \underline{L}_{a}), (\underline{L}_{a}, \underline{L}_{a}), (\underline{L}_{a}, \underline{L}_{a}), (\underline{L}_{a}, \underline{L}_{a}), (\underline{L}_{a}, \underline{L}_{a})), (\underline{L}_{a}, \underline{L}), (\underline{L}_{a}, \underline{L}), (\underline{L}_{a}, \underline{L})), (\underline{L}_{a}, \underline{L}), (\underline{L}, \underline{L})), (\underline{L}, \underline{L}), (\underline{L}, \underline{L})), ($		4.2.1, 4.1.1, 4.2.4
$(L_{r}, L_{s}, L_{sr}), L_{ss}$		() 1
	length of incremental line segment	6.2.1
	spring force is zero	2.2.1 a
l, l_w, l_y	length	
М	Hartmann number	14.2.2
M	mass of one mole of gas in kilograms	13.1.2
M	Mach number	13.2.1
М	mass	2.2.1c
М	number of mechanical terminal pairs	2.1.1
M, M ₈	mutual inductance	4.1.1, 4.2.4
М	magnetization density	1 .1.1a
m	mass/unit length of string	9.2
Ν	number of electrical terminal pairs	2.1.1
N	number of turns	5.2.2
n	number density of ions	12.3.1
n	integer	7.1.1
n	unit normal vector	1.1.2
P	polarization density	1.1.1a
Р	power	12.2.1a
Р	number of pole pairs in a machine	4.1.8
Р	power per unit area	14.2.1
Р	pressure	5.1.2d and 12.1.4
$P_{e}, P_{g}, P_{m}, P_{r}$	power	4.1.6a, 4.1.6b, 4.1.2,
		4.1.6b
Q	electric charge	7.2.1a
9, 9i, 9k	electric charge	1.1.3 and 2.1.2, 8.1, 2.1.2
R, R_i, R_o	radius	

Symbol	Meaning	Section	
$\overline{R_{t}, R_{a}, R_{b}, R_{f}, R_{r}, R_{s}}$	resistance		
(R, R_g)	gas constant	13.1.2	
R _e	electric Reynolds number	7.0	
R _m	magnetic Reynolds number	7.0	
r	radial coordinate		
r	position vector of material	2.2.1c	
r'	position vector in moving reference frame	6.1	
Г _т	center of mass of rigid body	2.2.1c	
S	reciprocal modulus of elasticity	11.5.2c	
S	surface of integration	1.1.2a	
S	normalized frequency	7.2.4	
S	membrane tension	9.2	
Sz	transverse force/unit length acting on string	9.2	
s	complex frequency	5.1.1	
(s, s_{mT})	slip	4.1.6b	
Si	ith root of characteristic equation, a	5.1.1	
~	natural frequency		
$\frac{T}{\pi}$	period of oscillation	5.2.1	
T T T T T T	temperature	13.1.2	
$T_{0}, T_{1}, T_{em}, T_{em}, T_{m}, T_{0}, T_{1}$	torque	2.2.1c, 5.1.2b, 3.1.1, 4.1.6b, 4.1.1, 6.4.1,	
_		6.4.1	
T	surface force	8.4	
T_{ij}^{m}	mechanical stress tensor	13.1.2	
T _{mn}	the component of the stress-tensor in the mth-direction on a cartesian surface with		
	a normal vector in the <i>n</i> th-direction	8.1	
T.,	constant of coulomb damping	4.1.1	
- 07 T.	initial stress distribution on thin rod	9.1.1	
T^{o}	longitudinal stress on a thin rod	911	
- T	transverse force per unit area on	2	
- 2	membrane	9.2	
Ta	transverse force per unit area acting on		
4	thin beam	11.4.2b	
t	time	1.1.1	
t'	time measured in moving reference frame	6.1	
U	gravitational potential	12.1.3	
U	longitudinal steady velocity of string or	10.0	
	membrane	10.2	
u	internal energy per unit mass	13.1.1	
u	surface coordinate	11.3	
$u_0(x-x_0)$	unit impulse at $x = x_0$	9.2.1	
u 	transverse deflection of wire in x-direction	10.4.3	
$u_{-1}(t)$	unit step occurring at $t = 0$	3.1.2D	
v, v _m	velocity	1.0, 13.2.3	
V V V V V V	volume	1,1,2	
r, _{Va} , _{Vf} , _{Vo} , _{Vs}	voltage	501	
V	potential energy	5.2.1	

Appendix D

Symbol	Meaning	Section	
<i>v</i> , v	velocity		
(v, v_1, \ldots, v_k)	voltage	2.1.1	
$v', (v_a, v_b, v_c),$	voltage		
$v_f, v_{\rm oc}, v_i$			
v _n	velocity of surface in normal direction	6.2.1	
v_o	initial velocity distribution on thin rod	9.1.1	
vp	phase velocity	9.1.1 and 10.2	
v ^r	relative velocity of inertial reference frames	6.1	
v _s	$\sqrt{f/m}$ for a string under tension f and having mass/unit length m	10.1.1	
v	longitudinal material velocity on thin rod	9.1.1	
v	transverse deflection of wire in y-direction	10.4.3	
(W_{e}, W_{m})	energy stored in electromechanical	211	
(W' W' W')	coepergy stored in electromechanical	3.1.1 3.1.2h	
("e, "m, ")	coupling	5.1.20	
W"	hybrid energy function	5.2.1	
W	width	5.2.2	
w.	energy density	11.5.2c	
w	coenergy density	8.5	
X	equilibrium position	5.1.2a	
$(x, x_1, x_2, \ldots, x_k)$	displacement of mechanical node	2.1.1	
x	dependent variable	5.1.1	
x_p	particular solution of differential equation	5.1.1	
$(x_1, x_2, x_3), (x, y, z)$	cartesian coordinates	8.1, 6.1	
(x', y', z')	cartesian coordinates of moving frame	6.1	
(α, β)	constants along C^{-} and C^{-} characteristics, respectively	9.1.1	
(α, β)	see (10.2.20) or (10.2.27)		
α	transverse wavenumber	11.4.3	
(α, β)	angles used to define shear strain	11.2	
(α, β)	constant angles	4.1.6b	
α	space decay parameter	7.1.4	
α	damping constant	5.1.2b	
α	equilibrium angle of torsional spring	2.2.1a	
γ	ratio of specific heats	13.1.2	
γ.	piezoelectric constant	11.5.2c	
Y, Y0, Y	angular position		
$\Delta_d(t)$	slope excitation of string	10.2.1b	
Δ_0	amplitude of sinusoidal slope excitation	10.2.1b	
Δr	distance between unstressed material points	11.2.1a	
Δs	distance between stressed positions of		
	material points	11.2.1a	
δ()	incremental change in ()	8.5	
$\boldsymbol{\delta}, \delta_1, \delta_0$	displacement of elastic material	11.1, 9.1, 11.4.2a	
δ	thickness of incremental volume element	6.2.1	
ð	torque angle	4.1.6a	

Symbol	Meaning	Section
δ_{ii}	Kronecker delta	8.1
$(\delta_{\perp}, \delta_{\perp})$	wave components traveling in the	
	$\pm x$ -directions	9.1.1
£	linear permittivity	1.1.1b
En	permittivity of free space	1.1.1a
n -0	efficiency of an induction motor	4.1 6b
n	second coefficient of viscosity	14 1 1c
θ. θ. θ	angular displacement	211 311 521
θ	nower factor angle: phase angle between	2.1, 0.1.1, 0.2.1
•	current and voltage	416a
A	equilibrium angle	5 2 1
à		5.2.1
0	angular velocity of armature	6.4.l
	maximum angular deflection	5.2.1
$(\Lambda, \Lambda_1, \Lambda_2, \ldots, \Lambda_k)$	magnetic flux linkage	2.1.1, 6.4.1, 4.1.7,
λ_a		4.1.3, 4.1
$(\Lambda_a, \Lambda_b, \Lambda_c)$		
$(\Lambda_{ar}, \Lambda_{as}, \Lambda_{br}, \Lambda_{bs})$		
(Λ_r, Λ_s)	To make a mode and Compatibulity and the international	11 0 2
λ 2	Lame constant for elastic material	11.2.3
٨	wavelength	7.1.4
μ	linear permeability	1.1.1a
μ , (μ_+ , μ)	mobility	12.3.1, 1.1.16
μ	coefficient of viscosity	14.1.1
μ_d	coefficient of dynamic friction	2.2.1b
μ_0	permeability of free space	1.1.1a
μ_s	coefficient of static friction	2.2.1b
v	Poisson's ratio for elastic material	11.2.2
v	damping frequency	10.1.4
(ξ, ξ)	continuum displacement	8.5
ξ ₀	initial deflection of string	9.2
ξ _d	amplitude of sinusoidal driving deflection	9.2
$(\xi_n(x), \hat{\xi}_n(x))$	nth eigenfunctions	9.2.1b
(ξ ₊ , ξ ₋)	amplitudes of forward and backward	
	traveling waves	9.2
$\dot{\xi}_{0}(x)$	initial velocity of string	9.2
D SUCCESSION	mass density	2.2.1c
ρ _f	free charge density	1.1.1a
ρ,	surface mass density	11.3
Σ	surface of discontinuity	6.2
σ	conductivity	1.1.1a
σ,	free surface charge density	1.1.1a
σ _m	surface mass density of membrane	9.2
σ	surface charge density	7.2.3
σ	surface conductivity	1.1.1a
σ,	surface charge density	7.2.3
τ	surface traction	8.2.1
τ, τ_d	diffusion time constant	7.1.1, 7.1.2a
T	relaxation time	7.2.1a

Appendix D

Symbol	Symbol Meaning				
Te	electrical time constant	5.2.2			
τ_m	time for air gap to close	5.2.2			
τ _o	time constant	5.1.3			
τ_t	traversal time	7.1.2 a			
ø	electric potential	7.2			
ø	magnetic flux	2.1.1			
φ	cylindrical coordinate	2.1.1			
ø	potential for H when $J_f = 0$	8.5.2			
φ	flow potential	12.2			
Xe	electric susceptibility	1.1 .1b			
Xm	magnetic susceptibility	1.1.1a			
ψ	the divergence of the material				
	displacement	11.4			
ψ	angle defined in Fig. 6.4.2	6.4.1			
Ψ	angular position in the air gap measured from stator winding (a) magnetic axis	414			
W	electromagnetic force potential	12.2			
v v	angular deflection of wire	10.4.3			
Ω	equilibrium rotational speed	5.1.2b			
Ω	rotation vector in elastic material	11.2.1a			
Ω"	real part of eigenfrequency (10.1.47)	10.1.4			
$\omega, (\omega_r, \omega_s)$	radian frequency of electrical excitation	4.1.6a, 4.1.2			
ω	natural angular frequency (Im s)	5.1.2b			
ω, ω,	angular velocity	2.2.1c, 4.1.2			
ω	cutoff frequency for evanescent waves	10.1.2			
ω _d	driving frequency	9.2			
ω _n	nth eigenfrequency	9.2			
ω	natural angular frequency	5.1.3			
(ω_r, ω_i)	real and imaginary parts of ω	10.0			
V	nabla	6.1			
∇_{Σ}	surface divergence	6.2.1			

Appendix E

SUMMARY OF PARTS I AND II AND USEFUL THEOREMS

IDENTITIES $\mathbf{A} \times \mathbf{B} \cdot \mathbf{C} = \mathbf{A} \cdot \mathbf{B} \times \mathbf{C},$ $\mathbf{A} \times (\mathbf{B} \times \mathbf{C}) = \mathbf{B}(\mathbf{A} \cdot \mathbf{C}) - \mathbf{C}(\mathbf{A} \cdot \mathbf{B})$ $\nabla(\phi + \psi) = \nabla\phi + \nabla\psi,$ $\nabla \cdot (\mathbf{A} + \mathbf{B}) = \nabla \cdot \mathbf{A} + \nabla \cdot \mathbf{B},$ $\nabla \times (\mathbf{A} + \mathbf{B}) = \nabla \times \mathbf{A} + \nabla \times \mathbf{B},$ $\nabla(\phi\psi) = \phi \,\nabla\psi + \psi \,\nabla\phi,$ $\nabla \cdot (\psi \mathbf{A}) = \mathbf{A} \cdot \nabla \psi + \psi \nabla \cdot \mathbf{A},$ $\nabla \cdot (\mathbf{A} \times \mathbf{B}) = \mathbf{B} \cdot \nabla \times \mathbf{A} - \mathbf{A} \cdot \nabla \times \mathbf{B},$ $\nabla \cdot \nabla \phi = \nabla^2 \phi,$ $\nabla \cdot \nabla \times \mathbf{A} = 0.$ $\nabla \times \nabla \phi = 0$, $\nabla \times (\nabla \times \mathbf{A}) = \nabla (\nabla \cdot \mathbf{A}) - \nabla^2 \mathbf{A},$ $(\nabla \times \mathbf{A}) \times \mathbf{A} = (\mathbf{A} \cdot \nabla)\mathbf{A} - \frac{1}{2}\nabla(\mathbf{A} \cdot \mathbf{A}),$ $\nabla (\mathbf{A} \cdot \mathbf{B}) = (\mathbf{A} \cdot \nabla)\mathbf{B} + (\mathbf{B} \cdot \nabla)\mathbf{A} + \mathbf{A} \times (\nabla \times \mathbf{B}) + \mathbf{B} \times (\nabla \times \mathbf{A})$ $\nabla \times (\phi \mathbf{A}) = \nabla \phi \times \mathbf{A} + \phi \nabla \times \mathbf{A},$ $\nabla \times (\mathbf{A} \times \mathbf{B}) = \mathbf{A}(\nabla \cdot \mathbf{B}) - \mathbf{B}(\nabla \cdot \mathbf{A}) + (\mathbf{B} \cdot \nabla)\mathbf{A} - (\mathbf{A} \cdot \nabla)\mathbf{B}.$

THEOREMS



E2

	Differential Equa	ations	Integral Equations	
Magnetic field system	$\nabla \times \mathbf{H} = \mathbf{J}_{f}$	(1.1.1)	$\oint_C \mathbf{H} \cdot d\mathbf{l} = \int_S \mathbf{J}_f \cdot \mathbf{n} da$	(1.1.20)
	$\boldsymbol{\nabla} \cdot \mathbf{B} = 0$	(1.1.2)	$\oint_S \mathbf{B} \cdot \mathbf{n} da = 0$	(1.1.21)
	$\boldsymbol{\nabla}\cdot\mathbf{J}_f=0$	(1.1.3)	$\oint_S \mathbf{J}_f \cdot \mathbf{n} da = 0$	(1,1.22)
	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	(1.1.5)	$\oint_C \mathbf{E}' \cdot d\mathbf{l} = -\frac{d}{dt} \int_S \mathbf{B} \cdot \mathbf{n} da$	(1.1.23)
			where $\mathbf{E}' = \mathbf{E} + \mathbf{v} \times \mathbf{B}$	
Electric field system	$\nabla \times \mathbf{E} = 0$	(1.1.11)	$\oint_C \mathbf{E} \cdot d\mathbf{l} = 0$	(1.1 .24)
	$\nabla \cdot \mathbf{D} = \rho_f$	(1.1.12)	$\oint_S \mathbf{D} \cdot \mathbf{n} da = \int_V \rho_f dV$	(1.1.25)
	$\nabla \cdot \mathbf{J}_f = -\frac{\partial \rho_f}{\partial t}$	(1.1.14)	$\oint_{S} \mathbf{J}'_{r} \cdot \mathbf{n} da = - \frac{d}{dt} \int_{V} \rho_{r} dV$	(1.1.26)
	$\nabla \times \mathbf{H} = \mathbf{J}_f + \frac{\partial \mathbf{D}}{\partial t}$	(1.1.15)	$\oint_C \mathbf{H}' \cdot d\mathbf{l} = \int_S \mathbf{J}'_f \cdot \mathbf{n} da + \frac{d}{dt} \int_S \mathbf{D} \cdot \mathbf{n} da$	(1.1.27)
			where $\mathbf{J}_{f}' = \mathbf{J}_{f} - \rho_{f} \mathbf{v}$	
			$\mathbf{H}' = \mathbf{H} - \mathbf{v} \times \mathbf{D}$	

Table 1.2 Summary of Ouast-Static Diectionagnetic Equa	Table 1.2	Summary of	of	Ouasi-Static	Electromagnetic	Equation
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Definition of Terminal Variables

Flux

Charge

Voltage

 $q_k = \int_{V_k} \rho_f \, dV$

 $\lambda_k = \int_{S_k} \mathbf{B} \cdot \mathbf{n} \, da$

Current

$$i_k = \int_{S_k} \mathbf{J}_f \cdot \mathbf{n}' \, da \qquad \qquad v_k = \int_a^b \mathbf{E} \cdot d\mathbf{I}$$

Terminal Conditions

$$v_{k} = \frac{d\lambda_{k}}{dt} \qquad i_{k} = \frac{dq_{k}}{dt}$$

$$\lambda_{k} = \lambda_{k}(i_{1}\cdots i_{N}; \text{ geometry}) \qquad q_{k} = q_{k}(v_{1}\cdots v_{N}; \text{ geometry})$$

$$i_{k} = i_{k}(\lambda_{1}\cdots \lambda_{N}; \text{ geometry}) \qquad v_{k} = v_{k}(q_{1}\cdots q_{N}; \text{ geometry})$$

Magnetic Field Systems		Electric Field Systems	
	Conservation	n of Energy	
$dW_m = \sum_{j=1}^N i_j d\lambda_j - \sum_{j=1}^M f_j^e dx_j$	(a)	$dW_e = \sum_{j=1}^N v_j dq_j - \sum_{j=1}^M f_j^e dx_j$	(b)
$dW'_m = \sum_{j=1}^N \lambda_j di_j + \sum_{j=1}^M f_j^e dx_j$	(c)	$dW'_{e} = \sum_{j=1}^{N} q_{j} dv_{j} + \sum_{j=1}^{M} f_{j}^{e} dx_{j}$	(d)
For	ces of Electric O	rigin, $j = 1, \ldots, M$	
$f_{j}^{e} = - \frac{\partial W_{m}(\lambda_{1}, \dots, \lambda_{N}; x_{1}, \dots, x_{M})}{\partial x_{j}}$	(e)	$f_j^e = -\frac{\partial W_e(q_1, \dots, q_N; x_1, \dots, x_M)}{\partial x_j}$	(f)
$f_j^e = \frac{\partial W'_m(i_1, \dots, i_N; x_1, \dots, x_M)}{\partial x_j}$	(g)	$f_j^e = \frac{\partial W'_e(v_1, \dots, v_N; x_1, \dots, x_M)}{\partial x_j}$	(h)
	Relation of Ener	rgy to Coenergy	
$W_m + W'_m = \sum_{j=1}^N \lambda_j i_j$	(i)	$W_e + W'_e = \sum_{j=1}^N v_j q_j$	(j)
Energy and	Coenergy from I	Electrical Terminal Relations	

and M Mechanical Terminal Pairs*

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$$W_{m} = \sum_{j=1}^{N} \int_{0}^{\lambda} i_{j}(\lambda_{1}, \dots, \lambda_{j-1}, \lambda'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) d\lambda'_{j} \quad (k) \qquad W_{e} = \sum_{j=1}^{N} \int_{0}^{q_{j}} v_{j}(q_{1}, \dots, q_{j-1}, q'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) dq'_{j} \quad (l) \\ W'_{m} = \sum_{j=1}^{N} \int_{0}^{i_{j}} \lambda_{j}(i_{1}, \dots, i_{j-1}, i'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) di'_{j} \quad (m) \qquad W'_{e} = \sum_{j=1}^{N} \int_{0}^{v_{j}} q_{j}(v_{1}, \dots, v_{j-1}, v'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) dv'_{j} \quad (n)$$

* The mechanical variables f_j and x_j can be regarded as the *j*th force and displacement or the *j*th torque T_j and angular displacement θ_j .

	Differential Equ	ations	Transforma	tions	Boundary Conditions	
	$\nabla \times \mathbf{H} = \mathbf{J}_f$	(1.1.1)	$\mathbf{H}' = \mathbf{H}$	(6.1.35)	$\mathbf{n} \times (\mathbf{H}^a - \mathbf{H}^b) = \mathbf{K}_f$	(6.2.14)
	$\nabla \cdot \mathbf{B} = 0$	(1.1.2)	$\mathbf{B}' = \mathbf{B}$	(6.1.37)	$\mathbf{n} \cdot (\mathbf{B}^a - \mathbf{B}^b) = 0$	(6.2.7)
Magnetic field	$\nabla \cdot \mathbf{J}_f = 0$	(1.1.3)	$\mathbf{J}_f' = \mathbf{J}_f$	(6.1.36)	$\mathbf{n} \cdot (\mathbf{J}_f^a - \mathbf{J}_f^b) + \boldsymbol{\nabla}_{\boldsymbol{\Sigma}} \cdot \mathbf{K}_f = 0$	(6.2.9)
systems	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	(1.1.5)	$\mathbf{E}' = \mathbf{E} + \mathbf{v}' \times \mathbf{B}$	(6.1.38)	$\mathbf{n} \times (\mathbf{E}^a - \mathbf{E}^b) = v_n (\mathbf{B}^a - \mathbf{B}^b)$	(6.2.22)
	$\mathbf{B} = \mu_0 (\mathbf{H} + \mathbf{M})$	(1.1.4)	$\mathbf{M}' = \mathbf{M}$	(6.1.39)		
	$\nabla \times \mathbf{E} = 0$	(1.1.11)	$\mathbf{E}' = \mathbf{E}$	(6.1.54)	$\mathbf{n} \times (\mathbf{E}^a - \mathbf{E}^b) = 0$	(6.2.31)
	$\nabla \cdot \mathbf{D} = \rho_f$	(1.1.12)	$\mathbf{D}' = \mathbf{D}$	(6.1.55)	$\mathbf{n} \cdot (\mathbf{D}^a - \mathbf{D}^b) = \sigma_f$	(6.2.33)
Electric field systems	$\nabla \cdot \mathbf{J}_f = -\frac{\partial \rho_f}{\partial t}$	(1.1.14)	$\begin{aligned} \rho_f' &= \rho_f \\ \mathbf{J}_f' &= \mathbf{J}_f - \rho_f \mathbf{v}^r \end{aligned}$	(6.1.56) (6.1.58)	$\mathbf{n} \cdot (\mathbf{J}_f^{\ a} - \mathbf{J}_f^{\ b}) + \nabla_{\Sigma} \cdot \mathbf{K}_f = v_n(\rho_f^{\ a} - \rho_f^{\ b}) - \frac{\partial \sigma_f}{\partial t}$	(6.2.36)
	$\nabla \times \mathbf{H} = \mathbf{J}_f + \frac{\partial \mathbf{D}}{\partial t}$	(1.1.15)	$\mathbf{H}' = \mathbf{H} - \mathbf{v}^r \times \mathbf{D}$	(6.1.57)	$\mathbf{n} \times (\mathbf{H}^a - \mathbf{H}^b) = \mathbf{K}_f + v_n \mathbf{n} \times [\mathbf{n} \times (\mathbf{D}^a - \mathbf{D}^b)]$	(6.2.38)
	$\mathbf{D} = \epsilon_0 \mathbf{E} + \mathbf{P}$	(1.1.13)	$\mathbf{P'} = \mathbf{P}$	(6.1.59)		

 Table 6.1 Differential Equations, Transformations, and Boundary Conditions for Quasi-static Electromagnetic Systems with Moving Media

Appendix F

GLOSSARY OF COMMONLY USED SYMBOLS

Section references indicate where symbols of a given significance are introduced; grouped symbols are accompanied by their respective references. The absence of a section reference indicates that a symbol has been applied for a variety of purposes. Nomenclature used in examples is not included.

Symbol	Meaning	Section
A	cross-sectional area	
A,	coefficient in differential equation	5.1.1
(A_{n}^{+}, A_{n}^{-})	complex amplitudes of components of <i>n</i> th	
	mode	9.2.1
An	cross-sectional area of armature conductor	6.4.1
a	spacing of pole faces in magnetic circuit	8.5.1
$a_1(a_c,a_s)$	phase velocity of acoustic related waves	13.2.1, 11.4.1
ab	Alfvén velocity	12.2.3
(a, b, c)	Lagrangian coordinates	11.1
ai	constant coefficient in differential equation	5.1.1
an	instantaneous acceleration of point p fixed	
u	in material	2.2.1c
B, B_r, B_s	damping constant for linear, angular and	
	square law dampers	2.2.1b, 4.1.1, 5.2.2
$\mathbf{B}, \mathbf{B}_i, B_0$	magnetic flux density	1.1.1a, 8.1, 6.4.2
B ₄	induced flux density	7.0
$(B_r, B_{ra}, B_{rb}, B_{rm})$	radial components of air-gap flux	
	densities	4.1.4
$[B_{rf}, (B_{rf})_{av}]$	radial flux density due to field current	6.4.1
b	width of pole faces in magnetic circuit	8.5
Ь	half thickness of thin beam	11 .4.2 b
С	contour of integration	1.1.2a
$C, (C_a, C_b), C_o$	capacitance	2.1.2, 7.2.1a, 5.2.1
С	coefficient in boundary condition	9.1.1
С	the curl of the displacement	11.4
(<i>C</i> ⁺ , <i>C</i> [−])	designation of characteristic lines	9.1.1

Symbol	Meaning	Section	
с _р	specific heat capacity at constant pressure	13.1.2	
C _v	specific heat capacity at constant volume	13.1.2	
D	electric displacement	1.1.1a	
d	length		
da	elemental area	1.1.2a	
df _n	total elemental force on material in rigid body	2.2.1c	
dl	elemental line segment	1.1.2a	
dT.	torque on elemental volume of material	2.2.1c	
dV	elemental volume	1.1.2b	
Ε	constant of motion	5.2.1	
Ε	Young's modulus or the modulus of elasticity	9.1	
E. <i>E</i> .	electric field intensity	1 1 1a 5 1 2d	
E, 20 E _f	magnitude of armature voltage generated by field current in a synchronous	A 1 6a	
F.	induced electric field intensity	7.0	
	strain tensor	01 11 2	
e11, eij	strain rate tensor	7.1, 11.2 14 1 10	
c _{ij} F	magnetomotive force (mmf)	12.7.7	
r F	force density	11.10	
r r		1.1.1a	
r r	complex amplitude of $f(t)$	5.1.1	
r ₀	amplitude of sinusoidal driving force	9.1.3	
Ĵ	equilibrium tension of string	9.2	
f A a m m a a a a	driving function	5.1.1	
<i>j</i> , f, <i>f°, f</i> [*] , <i>f_i, f_i, f_i</i> , <i>f</i> ₁	force	2.2.1, 2.2.1c, 3.1, 5.1.2a, 3.1.2b, 8.1, 9 1	
f	arbitrary scalar function	61	
f'	scalar function in moving coordinate	6.1	
£	three dimensional surface	6.2	
J f	integration appatent	0.2	
) C	Integration constant	11,4,2 a	
C	a constant	3.1.20	
C C	shear modulus of elasticity	11.2.2	
G	speed coefficient	0.4.1	
0 ~	conductance oir con longth	5.1	
8 a a	an-gap length	5,2.1	
χ, ε (Η Η Η Η Η)	magnetic field intensity	J.1.20, 12.1.5	
$(11, 11_{x}, 11_{y}, 11_{z})$	magnetic netalny	1317	
$\mathbf{\tilde{I}} I (I I) I$	electrical current	1043 12210 117	
L, L, (17, 18), 17		6.4.1	
$(i, i_1, i_2, \dots, i_k),$ $(i_{ar}, i_{as}, i_{br}, i_{bs}),$ $i_{ar}, (i_{ar}, i_{as}, i_{br}),$	electrical current	2.1, 4.1.3, 6.4.1, 4.1.7, 6.4.1, 4.1	
-a, \-a, -0, -c/)			

Appendix F

Symbol	Meaning	Section	
i _n	unit vector perpendicular to area of		
	integration	6.2.1	
i _s	unit vector normal to surface of		
	integration	6.2.1	
$(i_x, i_y, i_z), (i_1, i_2, i_3)$	unit vectors in coordinate directions	2.2.1c	
J, J_f	current density	7.0, 1.1.1a	
$J, J_{\tau}, (J_x, J_y, J_z)$	moment of inertia	5.1.2b, 4.1.1, 2.2.1c	
$J_{\alpha z}, J_{\gamma z}$	products of inertia	2.2.1c	
1	$\sqrt{-1}$	4.1.6a	
, K	loading factor	13.2.2	
K. K.	surface current density	7.0, 1.1.1a	
ĸ	linear or torsional spring constant	2.2.1a	
K.	induced surface current density	7.0	
$k_{1} k_{2} (k_{2} k_{3})$	wavenumber	7.1.3. 10.1.3. 10.0	
k	summation index	2.1.1	
k	maximum coefficient of coupling	4.1.6b	
<i>k</i>	<i>n</i> th eigenvalue	9.2	
$(L, L_{2}, L_{3}), (L_{2}, L_{4}),$	inductance	2.1.1. 6.4.1. 2.1.1.	
$L_{m_1}(L_0, L_0)$		4.2.1. 4.1.1. 4.2.4	
$(L_n, L_n, L_n), L_n$,,	
$L = \frac{1}{2}, \frac{1}{2$	length of incremental line segment	6.2.1	
ī	value of relative displacement for which	2.2.1a	
	spring force is zero		
I. I. L.	length		
M	Hartmann number	14.2.2	
M	mass of one mole of pas in kilograms	13.1.2	
M	Mach number	13.2.1	
M	mass	2.2.1c	
M	number of mechanical terminal pairs	2.1.1	
M.M.	mutual inductance	4.1.1.4.2.4	
M	magnetization density	1.1.1a	
m	mass/unit length of string	9.2	
N	number of electrical terminal pairs	2.1.1	
N	number of turns	5.2.2	
n	number density of jons	12.3.1	
n	integer	7.1.1	
n	unit normal vector	1.1.2	
P	polarization density	1.1.1a	
- P	power	12.2.1a	
~ n	number of nole pairs in a machine	4.1.8	
Р л	power per upit area	14.2.1	
r n	pressure	5.1.2d and 12.1.4	
r Dai Dai Dani, Da	power	4.1.6a, 4.1.6b, 4.1.2	
r vir gir mift	F	4.1.6b	
Q	electric charge	7.2.1a	
9, 9i, 9k	electric charge	1.1.3 and 2.1.2, 8.1, 2.1.2	
R, R_i, R_o	radius		

Symbol	Meaning	Section	
$\overline{R, R_a, R_b, R_f, R_r, R_s}$	resistance		
(R, R_g)	gas constant	13.1.2	
R,	electric Reynolds number	7.0	
R _m	magnetic Reynolds number	7.0	
r	radial coordinate		
r	position vector of material	2.2.1c	
r'	position vector in moving reference frame	6.1	
r _m	center of mass of rigid body	2.2.1c	
S	reciprocal modulus of elasticity	11.5.2c	
S	surface of integration	1.1.2a	
S	normalized frequency	7.2.4	
S	membrane tension	9.2	
S_z	transverse force/unit length acting on string	9.2	
S	complex frequency	5.1.1	
(s, s_{mT})	slip	4.1.6b	
Si	ith root of characteristic equation, a natural frequency	5.1.1	
Т	neriod of oscillation	521	
T T	temperature	13.1.2	
$\mathbf{\hat{T}}, T, T^e, T_{em}, T_m,$	torque	2.2.1c, 5.1.2b, 3.1.1,	
<i>I</i> ₀ , <i>I</i> ₁		4.1.60, 4.1.1, 6.4.1, 6.4.1	
Т	surface force	8.4	
T_{ii}^{m}	mechanical stress tensor	13,1,2	
T_{mn}	the component of the stress-tensor in the		
11611	mth-direction on a cartesian surface with		
	a normal vector in the <i>n</i> th-direction	8.1	
Ter	constant of coulomb damping	4.1.1	
- от Т.	initial stress distribution on thin rod	9.1.1	
T T	longitudinal stress on a thin rod	9.1.1	
T.	transverse force per unit area on		
- 2	membrane	9.2	
$T_{\rm b}$	transverse force per unit area acting on		
- 2	thin beam	11.4.2b	
t	time	1.1.1	
ť	time measured in moving reference frame	6.1	
U	gravitational potential	12.1.3	
U	longitudinal steady velocity of string or		
-	membrane	10.2	
11	internal energy per unit mass	13.1.1	
u u	surface coordinate	11.3	
$u_0(x-x_0)$	unit impulse at $x = x_0$	9.2.1	
u v	transverse deflection of wire in x-direction	10.4.3	
$u_{-1}(t)$	unit step occurring at $t = 0$	5.1.2b	
V, V_	velocity	7.0, 13.2.3	
V	volume	1.1.2	
V, V_a, V_t, V_a, V_s	voltage		
V	potential energy	5.2.1	

Appendix F

Symbol	Meaning	Section	
0, ₹	velocity		
(v, v_1, \ldots, v_k)	voltage	2.1.1	
$v', (v_a, v_b, v_c),$	voltage		
v_f, v_{0c}, v_t			
v _n	velocity of surface in normal direction	6.2.1	
vo	initial velocity distribution on thin rod	9.1,1	
v_p	phase velocity	9.1.1 and 10.2	
vr	relative velocity of inertial reference frames	6.1	
v_s	$\sqrt{f/m}$ for a string under tension f and having mass/unit length m	10.1.1	
v	longitudinal material velocity on thin rod	9.1.1	
v	transverse deflection of wire in y-direction	10.4.3	
(W_{e}, W_{m})	energy stored in electromechanical	211	
(W' W' W')	coupling	3.1.1 3.1.2h	
(w_e, w_m, w)	coupling	3.1.20	
W''	hybrid energy function	5.2.1	
w	width	5.2.2	
w	energy density	11.5.2c	
w'	coenergy density	8.5	
X	equilibrium position	5.1.2a	
$(x, x_1, x_2, \ldots, x_k)$	displacement of mechanical node	2.1.1	
2	dependent variable	5.1.1	
x_{n}	particular solution of differential equation	5.1.1	
$(x_1, x_2, x_3), (x, y, z)$	cartesian coordinates	8.1, 6.1	
(x', y', z')	cartesian coordinates of moving frame	6.1	
(α, β)	constants along C^+ and C^- characteristics,		
	respectively	9.1.1	
(α, β)	see (10.2.20) or (10.2.27)		
α	transverse wavenumber	11.4.3	
(α, β)	angles used to define shear strain	11.2	
(α, β)	constant angles	4.1.6b	
α	space decay parameter	7.1.4	
α	damping constant	5.1.26	
α	equilibrium angle of torsional spring	2.2.1a	
γ	ratio of specific heats	13.1.2	
γ,	piezoelectric constant	11.5.2c	
Y, Y0, Y	angular position	40 a 41	
$\Delta_d(t)$	slope excitation of string	10.2.15	
Δ_0	amplitude of sinusoidal slope excitation	10.2.16	
Δr	distance between unstressed material points	11.2.1a	
Δs	distance between stressed positions of	11 2 10	
\$()	incremental change in (11.2.1a 9.5	
0() \$ \$ \$	displacement of elastic material	0.5 11 1 0 1 11 4 7 m	
υ, 0 ₁ , 0 ₀	thickness of incremental volume element	6 2 1	
8	torque angle	416a	
U	iorque angre	1,1,04	

Symbol	Meaning	Section	
δ,,	Kronecker delta	8.1	
(δ_+, δ)	wave components traveling in the		
	$\pm x$ -directions	9.1.1	
£	linear permittivity	1.1.1b	
€0	permittivity of free space	1.1.1 a	
ή	efficiency of an induction motor	4.1.6b	
η	second coefficient of viscosity	14.1.1c	
$\theta, \theta_i, \theta_m$	angular displacement	2.1.1, 3.1.1, 5.2.1	
θ	power factor angle; phase angle between		
	current and voltage	4.1.6a	
θ	equilibrium angle	5.2.1	
θ	angular velocity of armature	6.4.1	
θ_m	maximum angular deflection	5.2.1	
$(\hat{\lambda}, \lambda_1, \lambda_2, \ldots, \lambda_k)$	magnetic flux linkage	2.1.1, 6.4.1, 4.1.7,	
λ_a		4.1.3, 4.1	
$(\lambda_a, \lambda_b, \lambda_c)$			
$(\lambda_{ar},\lambda_{as},\lambda_{br},\lambda_{bs})$			
(λ_r, λ_s)			
λ	Lamé constant for elastic material	11.2.3	
2.	wavelength	7.1.4	
μ	linear permeability	1.1.1a	
μ, (μ ₊ , μ_)	mobility	12.3.1, 1.1.1b	
μ	coefficient of viscosity	14.1.1	
μ_d	coefficient of dynamic friction	2.2.16	
μ_0	permeability of free space	1.1.1a	
μ_s	coefficient of static triction	2.2.16	
v	Poisson's ratio for elastic material	11.2.2	
ν (Ε. Ε.)	damping frequency	10.1.4	
(5 , <i>5</i>)	initial deflection of string	8.5	
50 5	initial deflection of string	9.2	
s_d (ε (m) $\hat{\varepsilon}$ (m))	with eigenfunctions	9.2 0.2.1h	
$(\varsigma_n(x), \varsigma_n(x))$	amplitudes of forward and backward	9.2.10	
(5+, 5-)	traveling waves	9.2	
; ; ;	initial unlocity of string	0.2	
$s_0(x)$	man donaity	9.2	
ρ	free charge density	1 1 19	
ρ _f	surface mass density	11.1.0	
Σ^{P_s}	surface of discontinuity	67	
 σ	conductivity	1.1.1a	
- σ4	free surface charge density	1.1.1a	
σ,,	surface mass density of membrane	9.2	
σ_0	surface charge density	7.2.3	
σ_s	surface conductivity	1.1.1a	
σ_u	surface charge density	7.2.3	
τ	surface traction	8.2.1	
τ, τ_d	diffusion time constant	7.1.1, 7.1.2a	
τ	relaxation time	7.2.1a	

F6

Appendix F

Symbol	Meaning	Section	
	electrical time constant	5.2.2	
τ_m	time for air gap to close	5.2.2	
το	time constant	5.1.3	
τ_t	traversal time	7.1.2a	
φ ^ˆ	electric potential	7.2	
ø	magnetic flux	2.1.1	
φ	cylindrical coordinate	2.1.1	
φ	potential for H when $J_f = 0$	8.5.2	
φ	flow potential	12.2	
Xe	electric susceptibility	1.1.1b	
Xm	magnetic susceptibility	1.1.1a	
Ψ	the divergence of the material		
	displacement	11.4	
ψ	angle defined in Fig. 6.4.2	6.4.1	
ψ	angular position in the air gap measured		
	from stator winding (a) magnetic axis	4.1.4	
Ψ	electromagnetic force potential	12.2	
ψ	angular deflection of wire	10.4.3	
Ω	equilibrium rotational speed	5.1.2b	
Ω	rotation vector in elastic material	11.2.1a	
Ω_n	real part of eigenfrequency (10.1.47)	10.1.4	
$\omega, (\omega_r, \omega_s)$	radian frequency of electrical excitation	4.1.6a, 4.1.2	
ω	natural angular frequency (Im s)	5.1.2b	
ω , ω _m	angular velocity	2.2.1c, 4.1.2	
ω	cutoff frequency for evanescent waves	10.1.2	
ω _d	driving frequency	9.2	
ω _n	nth eigenfrequency	9.2	
ω	natural angular frequency	5.1.3	
(ω_r, ω_i)	real and imaginary parts of ω	10.0	
V	nabla	6.1	
∇_{Σ}	surface divergence	6.2.1	

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F7

Appendix G

SUMMARY OF PARTS I AND II AND USEFUL THEOREMS

IDENTITIES $\mathbf{A} \times \mathbf{B} \cdot \mathbf{C} = \mathbf{A} \cdot \mathbf{B} \times \mathbf{C},$ $\mathbf{A} \times (\mathbf{B} \times \mathbf{C}) = \mathbf{B}(\mathbf{A} \cdot \mathbf{C}) - \mathbf{C}(\mathbf{A} \cdot \mathbf{B})$ $\nabla(\phi + \psi) = \nabla\phi + \nabla\psi,$ $\nabla \cdot (\mathbf{A} + \mathbf{B}) = \nabla \cdot \mathbf{A} + \nabla \cdot \mathbf{B},$ $\nabla \times (\mathbf{A} + \mathbf{B}) = \nabla \times \mathbf{A} + \nabla \times \mathbf{B},$ $\nabla(\phi\psi) = \phi \,\nabla\psi + \psi \,\nabla\phi,$ $\nabla \cdot (\psi \mathbf{A}) = \mathbf{A} \cdot \nabla \psi + \psi \nabla \cdot \mathbf{A},$ $\nabla \cdot (\mathbf{A} \times \mathbf{B}) = \mathbf{B} \cdot \nabla \times \mathbf{A} - \mathbf{A} \cdot \nabla \times \mathbf{B},$ $\nabla \cdot \nabla \phi = \nabla^2 \phi.$ $\nabla \cdot \nabla \times \mathbf{A} = \mathbf{0},$ $\nabla \times \nabla \phi = 0$, $\nabla \times (\nabla \times \mathbf{A}) = \nabla (\nabla \cdot \mathbf{A}) - \nabla^2 \mathbf{A},$ $(\nabla \times \mathbf{A}) \times \mathbf{A} = (\mathbf{A} \cdot \nabla)\mathbf{A} - \frac{1}{2}\nabla(\mathbf{A} \cdot \mathbf{A}),$ $\nabla (\mathbf{A} \cdot \mathbf{B}) = (\mathbf{A} \cdot \nabla)\mathbf{B} + (\mathbf{B} \cdot \nabla)\mathbf{A} + \mathbf{A} \times (\nabla \times \mathbf{B}) + \mathbf{B} \times (\nabla \times \mathbf{A})$ $\nabla \times (\phi \mathbf{A}) = \nabla \phi \times \mathbf{A} + \phi \nabla \times \mathbf{A},$ $\nabla \times (\mathbf{A} \times \mathbf{B}) = \mathbf{A}(\nabla \cdot \mathbf{B}) - \mathbf{B}(\nabla \cdot \mathbf{A}) + (\mathbf{B} \cdot \nabla)\mathbf{A} - (\mathbf{A} \cdot \nabla)\mathbf{B}.$ G1

THEOREMS



	Differential Equa	ations	Integral Equations	
Magnetic field system	$\nabla \times \mathbf{H} = \mathbf{J}_f$	(1.1.1)	$\oint_C \mathbf{H} \cdot d\mathbf{l} = \int_S \mathbf{J}_f \cdot \mathbf{n} da$	(1.1.20)
	$\nabla \cdot \mathbf{B} = 0$	(1.1.2)	$\oint_S \mathbf{B} \cdot \mathbf{n} da = 0$	(1.1.21)
	$\boldsymbol{\nabla}\cdot\mathbf{J}_f=0$	(1.1.3)	$\oint_S \mathbf{J}_f \cdot \mathbf{n} da = 0$	(1.1.22)
	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	(1.1.5)	$\oint_C \mathbf{E}' \cdot d\mathbf{l} = -\frac{d}{dt} \int_S \mathbf{B} \cdot \mathbf{n} da$	(1.1.23)
			where $\mathbf{E}' = \mathbf{E} + \mathbf{v} \times \mathbf{B}$	
Electric field system	$\nabla \times \mathbf{E} = 0$	(1.1.11)	$\oint_C \mathbf{E} \cdot d\mathbf{l} = 0$	(1.1 .24)
	$\nabla \cdot \mathbf{D} = \rho_f$	(1.1.12)	$\oint_S \mathbf{D} \cdot \mathbf{n} da = \int_V \rho_f dV$	(1.1.25)
	$\nabla \cdot \mathbf{J}_f = -\frac{\partial \rho_f}{\partial t}$	(1.1.14)	$\oint_{S} \mathbf{J}'_{r} \cdot \mathbf{n} da = -\frac{d}{dt} \int_{V} \rho_{f} dV$	(1.1.26)
	$\nabla \times \mathbf{H} = \mathbf{J}_f + \frac{\partial \mathbf{D}}{\partial t}$	(1.1.15)	$\oint_C \mathbf{H}' \cdot d\mathbf{l} = \int_S \mathbf{J}'_r \cdot \mathbf{n} da + \frac{d}{dt} \int_S \mathbf{D} \cdot \mathbf{n} da$	(1.1.27)
			where $\mathbf{J}_{f}' = \mathbf{J}_{f} - \rho_{f} \mathbf{v}$	
			$\mathbf{H}' = \mathbf{H} - \mathbf{v} \times \mathbf{D}$	

Table 1.2 Summary of Ouasi-Static Electromagnetic Equation	Table 1	1.2	Summary	of	Ouasi-Static	Electromagnetic	Equation
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Table 2.1 Summary of Terminal Variables and Terminal Relations



Definition of Terminal Variables

Flux

$$\lambda_k = \int_{S_k} \mathbf{B} \cdot \mathbf{n} \, da$$

 $q_k = \int_{V_k} \rho_f \, dV$ Voltage

Charge

Current

$$i_k = \int_{S_k'} \mathbf{J}_f \cdot \mathbf{n}' \, da$$

...

$$v_k = \int_a^b \mathbf{E} \cdot d\mathbf{l}$$

Terminal Conditions

$$v_{k} = \frac{d\lambda_{k}}{dt} \qquad i_{k} = \frac{dq_{k}}{dt}$$

$$\lambda_{k} = \lambda_{k}(i_{1}\cdots i_{N}; \text{ geometry}) \qquad q_{k} = q_{k}(v_{1}\cdots v_{N}; \text{ geometry})$$

$$i_{k} = i_{k}(\lambda_{1}\cdots \lambda_{N}; \text{ geometry}) \qquad v_{k} = v_{k}(q_{1}\cdots q_{N}; \text{ geometry})$$

Magnetic Field Systems		Electric Field Systems		
Con	Conservation of Energy			
$dW_m = \sum_{j=1}^N i_j d\lambda_j - \sum_{j=1}^M f_j^e dx_j$	(a)	$dW_{e} = \sum_{j=1}^{N} v_{j} dq_{j} - \sum_{j=1}^{M} f_{j}^{e} dx_{j}$	(b)	
$dW'_m = \sum_{j=1}^N \lambda_j di_j + \sum_{j=1}^M f_j^e dx_j$	(c)	$dW'_{e} = \sum_{j=1}^{N} q_{j} dv_{j} + \sum_{j=1}^{M} f_{j}^{e} dx_{j}$	(d)	
Forces of Electric Origin, $j = 1,, M$				
$f_j^e = - \frac{\partial W_m(\lambda_1, \dots, \lambda_N; x_1, \dots, x_M)}{\partial x_j}$	(e)	$f_j^e = -\frac{\partial W_e(q_1, \dots, q_N; x_1, \dots, x_M)}{\partial x_j}$	(f)	
$f_j^e = \frac{\partial W'_m(i_1, \dots, i_N; x_1, \dots, x_M)}{\partial x_j}$	(g)	$f_j^e = \frac{\partial W'_e(v_1, \dots, v_N; x_1, \dots, x_M)}{\partial x_j}$	(h)	
Relation of Energy to Coenergy				
$W_m + W'_m = \sum_{j=1}^N \lambda_j i_j$	(i)	$W_e + W'_e = \sum_{j=1}^N v_j q_j$	(j)	
Energy and Coenergy from Electrical Terminal Relations				

Table 3.1 Energy Relations for an Electromechanical Coupling Network with N Electrical and M Mechanical Terminal Pairs*

$$W_{m} = \sum_{j=1}^{N} \int_{0}^{\lambda_{j}} i_{j}(\lambda_{1}, \dots, \lambda_{j-1}, \lambda'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) d\lambda'_{j} \quad (k) \qquad W_{e} = \sum_{j=1}^{N} \int_{0}^{q_{j}} v_{j}(q_{1}, \dots, q_{j-1}, q'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) dq'_{j} \quad (l) \\ W'_{m} = \sum_{j=1}^{N} \int_{0}^{i_{j}} \lambda_{j}(i_{1}, \dots, i_{j-1}, i'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) di'_{j} \quad (m) \qquad W'_{e} = \sum_{j=1}^{N} \int_{0}^{v_{j}} q_{j}(v_{1}, \dots, v_{j-1}, v'_{j}, 0, \dots, 0; x_{1}, \dots, x_{M}) dv'_{j} \quad (n)$$

* The mechanical variables f_j and x_j can be regarded as the *j*th force and displacement or the *j*th torque T_j and angular displacement θ_j .

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	Differential Equ	ations	Transforma	tions	Boundary Conditions	
	$\nabla \times \mathbf{H} = \mathbf{J}_f$	(1.1.1)	$\mathbf{H}' = \mathbf{H}$	(6.1.35)	$\mathbf{n} \times (\mathbf{H}^a - \mathbf{H}^b) = \mathbf{K}_f$	(6.2.14)
	$\nabla \cdot \mathbf{B} = 0$	(1.1.2)	$\mathbf{B}' = \mathbf{B}$	(6.1.37)	$\mathbf{n} \cdot (\mathbf{B}^a - \mathbf{B}^b) = 0$	(6.2.7)
Magnetic field	$\nabla \cdot \mathbf{J}_f = 0$	(1.1.3)	$\mathbf{J}_f' = \mathbf{J}_f$	(6.1.36)	$\mathbf{n} \cdot (\mathbf{J}_f^a - \mathbf{J}_f^b) + \boldsymbol{\nabla}_{\boldsymbol{\Sigma}} \cdot \mathbf{K}_f = 0$	(6.2.9)
systems	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	(1.1.5)	$\mathbf{E}' = \mathbf{E} + \mathbf{v}' \times \mathbf{B}$	(6.1.38)	$\mathbf{n} \times (\mathbf{E}^a - \mathbf{E}^b) = v_n (\mathbf{B}^a - \mathbf{B}^b)$	(6.2.22)
	$\mathbf{B} = \mu_0 (\mathbf{H} + \mathbf{M})$	(1.1.4)	$\mathbf{M}' = \mathbf{M}$	(6.1.39)		
	$\nabla \times \mathbf{E} = 0$	(1.1.11)	$\mathbf{E}' = \mathbf{E}$	(6.1.54)	$\mathbf{n} \times (\mathbf{E}^a - \mathbf{E}^b) = 0$	(6.2.31)
	$\nabla \cdot \mathbf{D} = \rho_f$	(1.1.12)	$\mathbf{D}' = \mathbf{D}$	(6.1.55)	$\mathbf{n} \cdot (\mathbf{D}^a - \mathbf{D}^b) = \sigma_f$	(6.2.33)
			$\rho_f' = \rho_f$	(6.1.56)		
Electric field	$\nabla \cdot \mathbf{J}_f = -\frac{d\rho_f}{\partial t}$	(1.1.14)	$\mathbf{J}_f' = \mathbf{J}_f - \rho_f \mathbf{v}^r$	(6.1.58)	$\mathbf{n} \cdot (\mathbf{J}_f^a - \mathbf{J}_f^b) + \nabla_{\Sigma} \cdot \mathbf{K}_f = v_n (\rho_f^a - \rho_f^b) - \frac{\partial \sigma_f}{\partial t}$	(6.2.36)
systems	$\nabla \times \mathbf{H} = \mathbf{J}_f + \frac{\partial \mathbf{D}}{\partial t}$	(1.1.15)	$\mathbf{H}' = \mathbf{H} - \mathbf{v}' \times \mathbf{D}$	(6.1.57)	$\mathbf{n} \times (\mathbf{H}^a - \mathbf{H}^b) = \mathbf{K}_f + v_n \mathbf{n} \times [\mathbf{n} \times (\mathbf{D}^a - \mathbf{D}^b)]$	(6.2.38)
	$\mathbf{D} = \epsilon_0 \mathbf{E} + \mathbf{P}$	(1.1.13)	$\mathbf{P}' = \mathbf{P}$	(6.1.59)		

 Table 6.1 Differential Equations, Transformations, and Boundary Conditions for Quasi-static Electromagnetic Systems with Moving Media

From Chapter 8; The Stress Tensor and Related Tensor Concepts

In what follows we assume a right-hand cartesian coordinate system x_1, x_2, x_3 . The component of a vector in the direction of an axis carries the subscript of that axis. When we write F_m we mean the *m*th component of the vector **F**, where *m* can be 1, 2, or 3. When the index is repeated in a single term, it implies summation over the three values of the index

dH.

dH.

ЭН

AH.

and

$$\frac{\partial \mathbf{u}_n}{\partial x_n} = \frac{\partial \mathbf{u}_1}{\partial x_1} + \frac{\partial \mathbf{u}_2}{\partial x_2} + \frac{\partial \mathbf{u}_3}{\partial x_3} = \mathbf{\nabla} \cdot \mathbf{H}$$
$$H_n \frac{\partial}{\partial x_n} = H_1 \frac{\partial}{\partial x_1} + H_2 \frac{\partial}{\partial x_2} + H_3 \frac{\partial}{\partial x_3} = \mathbf{H} \cdot \mathbf{\nabla}.$$

This illustrates the summation convention. On the other hand, $\partial H_m/\partial x_n$ represents any one of the nine possible derivatives of components of H with respect to coordinates. We define the Kronecker delta δ_{mn} which has the values

$$\delta_{mn} = \begin{cases} 1, \text{ when } m = n, \\ 0, \text{ when } m \neq n. \end{cases}$$
(8.1.7)

The component T_{mn} of the stress tensor can be physically interpreted as the mth component of the traction (force per unit area) applied to a surface with a normal vector in the n-direction.



Fig. 8.2.2 Rectangular volume with center at (x_1, x_2, x_3) showing the surfaces and directions of the stresses T_{mn} .

The x_1 -component of the total force applied to the material within the volume of Fig. 8.2.2 is

$$f_{1} = T_{11}\left(x_{1} + \frac{\Delta x_{1}}{2}, x_{2}, x_{3}\right) \Delta x_{2} \Delta x_{3} - T_{11}\left(x_{1} - \frac{\Delta x_{1}}{2}, x_{2}, x_{3}\right) \Delta x_{2} \Delta x_{3} + T_{12}\left(x_{1}, x_{2} + \frac{\Delta x_{2}}{2}, x_{3}\right) \Delta x_{1} \Delta x_{3} - T_{12}\left(x_{1}, x_{2} - \frac{\Delta x_{2}}{2}, x_{3}\right) \Delta x_{1} \Delta x_{3} + T_{13}\left(x_{1}, x_{2}, x_{3} + \frac{\Delta x_{3}}{2}\right) \Delta x_{1} \Delta x_{2} - T_{13}\left(x_{1}, x_{2}, x_{3} - \frac{\Delta x_{3}}{2}\right) \Delta x_{1} \Delta x_{2}.$$

$$(8.2.3)$$

Here we have evaluated the components of the stress tensor at the centers of the surfaces on which they act; for example, the stress component T_{11} acting on the top surface is evaluated at a point having the same x_{2} and x_{3} -coordinates as the center of the volume but an x_{1} coordinate $\Delta x_{1}/2$ above the center.

The dimensions of the volume have already been specified as quite small. In fact, we are interested in the limit as the dimensions go to zero. Consequently, each component of the stress tensor is expanded in a Taylor series about the value at the volume center with only linear terms in each series retained to write (8.2.3) as

$$f_{1} = \left(T_{11} + \frac{\Delta x_{1}}{2} \frac{\partial T_{11}}{\partial x_{1}} - T_{11} + \frac{\Delta x_{1}}{2} \frac{\partial T_{11}}{\partial x_{1}}\right) \Delta x_{2} \Delta x_{3}$$
$$+ \left(T_{12} + \frac{\Delta x_{2}}{2} \frac{\partial T_{12}}{\partial x_{2}} - T_{12} + \frac{\Delta x_{2}}{2} \frac{\partial T_{12}}{\partial x_{2}}\right) \Delta x_{1} \Delta x_{3}$$
$$+ \left(T_{13} + \frac{\Delta x_{3}}{2} \frac{\partial T_{13}}{\partial x_{3}} - T_{13} + \frac{\Delta x_{3}}{2} \frac{\partial T_{13}}{\partial x_{3}}\right) \Delta x_{1} \Delta x_{2}$$

or

$$f_1 = \left(\frac{\partial T_{11}}{\partial x_1} + \frac{\partial T_{12}}{\partial x_2} + \frac{\partial T_{13}}{\partial x_3}\right) \Delta x_1 \Delta x_2 \Delta x_3.$$
(8.2.4)

All terms in this expression are to be evaluated at the center of the volume (x_1, x_2, x_3) . We have thus verified our physical intuition that space-varying stress tensor components are necessary to obtain a net force.

From (8.2.4) we can obtain the x_1 -component of the force density **F** at the point (x_1, x_2, x_3) by writing

$$F_1 = \lim_{\Delta x_1, \Delta x_2, \Delta x_3 \to 0} \frac{f_1}{\Delta x_1 \Delta x_2 \Delta x_3} = \frac{\partial T_{11}}{\partial x_1} + \frac{\partial T_{12}}{\partial x_2} + \frac{\partial T_{13}}{\partial x_3}.$$
 (8.2.5)

The limiting process makes the expansion of (8.2.4) exact. The summation convention is used to write (8.2.5) as

$$F_1 = \frac{\partial T_{1n}}{\partial x_n}.$$
(8.2.6)

A similar process for the other two components of the force and force density yields the general result that the *m*th component of the force density at a point is 2π

$$F_m = \frac{\partial T_{mn}}{\partial x_n} \,. \tag{8.2.7}$$

Now suppose we wish to find the *m*th component of the total force f on material contained within the volume V. We can find it by performing the volume integration:

$$f_m = \int_V F_m \, dV = \int_V \frac{\partial T_{mn}}{\partial x_n} \, dV. \tag{8.1.13}$$

When we define the components of a vector A as

$$A_1 = T_{m1}, \quad A_2 = T_{m2}, \quad A_3 = T_{m3}, \quad (8.1.14)$$

we can write (8.1.13) as

$$f_m = \int_V \frac{\partial A_n}{\partial x_n} \, dV = \int_V (\nabla \cdot \mathbf{A}) \, dV. \tag{8.1.15}$$

We now use the divergence theorem to change the volume integral to a surface integral,

$$f_m = \oint_S \mathbf{A} \cdot \mathbf{n} \, da = \oint_S A_n n_n \, da, \qquad (8.1.16)$$

where n_n is the *n*th component of the outward-directed unit vector **n** normal to the surface S and the surface S encloses the volume V. Substitution from (8.1.14) back into this expression yields

$$f_m = \oint_S T_{mn} n_n \, da. \tag{8.1.17}$$

where $T_{mn}n_n$ is the *m*th component of the surface traction τ .

The traction τ is a vector. The components of this vector depend on the coordinate system in which τ is expressed; for example, the vector might be directed in one of the coordinate directions (x_1, x_2, x_3) , in which case there would be only one nonzero component of τ . In a second coordinate system (x'_1, x'_2, x'_3) , this same vector might have components in all of the coordinate directions. Analyzing a vector into orthogonal components along the coordinate axes is a familiar process. The components in a cartesian coordinate system (x'_1, x'_2, x'_3) are related to those in the cartesian coordinate system (x_1, x_2, x_3) by the three equations

$$\tau_p' = a_{p\tau}\tau_r, \tag{8.2.10}$$

where a_{pr} is the cosine of the angle between the x'_{p} -axis and the x_{r} -axis.

Similarly, the components of the stress tensor transform according to the equation

$$T'_{pq} = a_{pr}a_{qs}T_{rs}.$$
 (8.2.17)

This relation provides the rule for finding the components of the stress in the primed coordinates, given the components in the unprimed coordinates. It serves the same purpose in dealing with tensors that (8.2.10) serves in dealing with vectors.

Equation 8.2.10 is the transformation of a vector τ from an unprimed to a primed coordinate system. There is, in general, nothing to distinguish the two coordinate systems. We could just as well define a transformation from the primed to the unprimed coordinates by

$$\tau_s = b_{sp} \tau'_p, \tag{8.2.18}$$

where b_{sp} is the cosine of the angle between the x_s -axis and the x'_p -axis. But b_{sp} , from the definition following (8.2.10), is then also

$$b_{sp} \equiv a_{ps}; \tag{8.2.19}$$

that is, the transformation which reverses the transformation (8.2.10) is

$$\tau_s = a_{ps} \tau'_p. \tag{8.2.20}$$

Now we can establish an important property of the direction cosines a_{ps} by transforming the vector τ to an arbitrary primed coordinate system and then transforming the components τ'_m back to the unprimed system in which they must be the same as those we started with. Equation 8.2.10 provides the first transformation, whereas (8.2.20) provides the second; that is, we substitute (8.2.10) into (8.2.20) to obtain

$$\tau_s = a_{ps} a_{pr} \tau_r. \tag{8.2.21}$$

Remember that we are required to sum on both p and r; for example, consider the case in which s = 1:

$$\tau_{1} = (a_{11}a_{11} + a_{21}a_{21} + a_{31}a_{31})\tau_{1} + (a_{11}a_{12} + a_{21}a_{22} + a_{31}a_{32})\tau_{2} + (a_{11}a_{13} + a_{21}a_{23} + a_{31}a_{33})\tau_{3}.$$
(8.2.22)

This relation must hold in general. We have not specified either a_{ps} or τ_m . Hence the second two bracketed quantities must vanish and the first must be unity. We can express this fact much more concisely by stating that in general

$$a_{ps}a_{pr} = \delta_{sr} \tag{8.2.23}$$

G10

Description	Force Density F	Stress Tensor T_{mn} $F_m = \frac{\partial T_{mn}}{\partial x_n} (8.1.10)$	Surface Force Density* $T_m = [T_{mn}]n_n$ (8.4.2)
Force on media carrying free current density J _f ,	$J_f \times B$	$T_{mn} = \mu H_m H_n - \delta_{mn} \frac{1}{2} \mu H_k H_k$	$T = K_f \times \mu \langle H \rangle$ $K_f = n \times [H]$
μ constant	(8.1.3)	(8.1.11)	(8.4.3)
Force on media supporting free charge density ρ_f , ϵ constant	ρ _f E	$T_{mn} = \epsilon E_m E_n - \delta_{mn^{\frac{1}{2}}} \epsilon E_k E_k$	$\mathbf{T} = \sigma_f \langle \mathbf{E} \rangle$ $\sigma_f = \mathbf{n} \cdot [\epsilon \mathbf{E}]$
	(8.3.3)	(8.3.10)	(8.4.8)
Force on free current plus magnetization force in which $\mathbf{B} = \mu \mathbf{H}$ both before and after media are deformed	$\mathbf{J}_f \times \mathbf{B} - \tfrac{1}{2} \mathbf{H} \cdot \mathbf{H} \nabla \boldsymbol{\mu}$	$T_{mn} = \mu H_m H_n$	
	$+ \frac{1}{2} \nabla \left(\mathbf{H} \cdot \mathbf{H} \rho \frac{\partial \mu}{\partial \rho} \right)$	$-\frac{1}{2}\delta_{mn}\left(\mu-\rho\frac{\partial\mu}{\partial\rho} ight)H_{k}H_{k}$	
	(8.5.38)	(8.5.41)	
Force on free charge plus polarization force in which $D = \epsilon E$ both before and after media are deformed	$ ho_f \mathbf{E} - \frac{1}{2} \mathbf{E} \cdot \mathbf{E} \boldsymbol{\nabla} \boldsymbol{\epsilon}$	$T_{mn} = \epsilon E_m E_n$	
	$+ \frac{1}{2} \nabla \left(\mathbf{E} \cdot \mathbf{E} \rho \frac{\partial \epsilon}{\partial \rho} \right)$	$-\frac{1}{2}\delta_{mn}\left(\epsilon- horac{\partial\epsilon}{\partial ho} ight)^{E_{k}E_{k}}$	
	(8.5.45)	(8.5.46)	

 Table 8.1
 Electromagnetic Force Densities, Stress Tensors, and Surface Force Densities for Quasi-static

 Magnetic and Electric Field Systems*

Material	<i>E</i> -units of 10 ¹¹ N/m ²	ρ-units of 10 ³ kg/m ³	v _p -units† of m/sec
Aluminum (pure and alloy)	0.68-0.79	2.66-2.89	5100
Brass (60-70% Cu, 40-30% Zn)	1.0-1.1	8.368.51	3500
Copper	1.17-1.24	8.95-8.98	3700
Iron, cast (2.7-3.6% C)	0.89-1.45	6.96-7.35	4000
Steel (carbon and low alloy)	1.93-2.20	7.737.87	5100
Stainless steel (18% Cr, 8% Ni)	1.93-2.06	7.65-7.93	5100
Titanium (pure and alloy)	1.06-1.14	4.52	4900
Glass	0.49-0.79	2.383.88	4500
Methyl methacrylate	0.024-0.034	1.16	1600
Polyethylene	$1.38 - 3.8 \times 10^{-3}$	0.915	530
Rubber	0.79-4.1 × 10 ⁻⁵	0.99-1.245	46

Table 9.1 Modulus of Elasticity E and Density ρ for Representative Materials*

* See S. H. Crandall, and N. C. Dahl, An Introduction to the Mechanics of Solids, McGraw-Hill, New York, 1959, for a list of references for these constants and a list of these constants in English units.

† Computed from average values of E and ρ .

Table 9.2 Summary of One-Dimensional Mechanical Continua Introduced in Chapter 9

Thin Elastic Rod		
	$\rho \frac{\partial^2 \delta}{\partial t^2} = E \frac{\partial^2 \delta}{\partial x^2} + F_x$ $T = E \frac{\partial \delta}{\partial x}$ $\delta \text{longitudinal (x) displacement}$ T normal stress $\rho \text{mass density}$ E modulus of elasticity $F_x \text{longitudinal body force density}$	
Wire or "S	tring"	
	$m \frac{\partial^2 \xi}{\partial t^2} = f \frac{\partial^2 \xi}{\partial x^2} + S_z$ ξ —transverse displacement m—mass/unit length	
	f—tension (constant force) S_z —transverse force/unit length	
Membra	ane	
	$\sigma_m \frac{\partial^2 \xi}{\partial t^2} = S\left(\frac{\partial^2 \xi}{\partial x^2} + \frac{\partial^2 \xi}{\partial y^2}\right) + T_z$	
	ξ —transverse displacement σ_m —surface mass density S—tension in y- and z-directions (constant force per unit length) T_z —z-directed force per unit area	

G13

INDEX

Numbers preceded by letters are Appendix references. Appendices A, B, and C are in Part One; Appendices D and E, Part Two; and Appendices F and G, Part Three.

Acceleration, centrifugal fluid, 729 centripetal, 59 Coriolis, 59 Eulerian variable, 727 fluid, 727 instantaneous, 45 Accelerator, electric field, 776 MHD, 825 particle, 608 Acoustic delay lines, 480 Acoustic waves, compressional in solid. 673 dilatational in solid, 673 elastic media, 671 fluid, 544 gases, 845 guided, 679, 683, 693 magnetic fields and, 846 membrane, 509 shear elastic, 675 string, 509 thin beam, 683 thin rod, 487, 681 Acyclic machine, 286 Air-gap magnetic fields, 114 Alfvén velocity, 763 Alfvén waves, 759 compressible fluids and, 841 cylindrical geometry, 767 effect of conductivity on, 772 mechanical analogue to, 766 nature of, 764 numerical example of, 771 resonances and, 771 standing, 771 torsional, 765 Amortisseur winding, 164 Ampère, 1 Ampère's law, B6, C3, E3, G3 dynamic, B9 electromechanical, 304 example of, B7 integral form of, B36, C3, E3, G3 magnetization and, B26 Amplifying wave, coupled system and, 608 electric field induced, 605 evanescent wave and, 607 space-time behavior of, 604, 606 Angular frequency, 513 Angular momentum, 248 Angular velocity, 47 Applications of electromechanics, 2 Approximations, electromechanical, 206 Armature, ac machine, 120 dc machine, 141, 293 Armature reaction, 297

Astrophysics and MHD, 552 Attenuation, microwave, 561 Average power converted, salient pole machine, 155 smooth-air-gap machine, 124 Beats in space, 595 Bernoulli's equation, 738 example of, 752 **Bessel functions**, 408 roots of, 409 Bias, linear transducer operation and, 201 piezoelectricity and, 711 Bode plot, 206 Boundary, analytic description of, 269, 668 examples of moving, 269, 276, 279, 280, 364, 392, 397, 451, 460, 563, 574, 605, 627, 704, 783 moving, 267 well defined, 267 Boundary condition, Alfvén waves, 769 causality and, 491, 592, 607 conservation of charge, 279, 374, 376, 394, 399 convection and, 267, 587, 598 dispersion and, 618 elastic media, 671, 676 electric displacement, 278 electric field intensity, 275, 278 electric field systems, 277, E6, G6 electromagnetic field, 267 electromechanical, 668 field transformations and, 275 geometric effect of, 280 initial condition and, 513 inviscid fluid, 752 inviscid fluid slip, 740 longitudinal and transverse, 680 magnetic field intensity, 273, 280 magnetic field systems, 270, E6, G6 magnetic field system current, 272 magnetic fluid, 774 magnetic flux density, 271 MHD, 769 motion and, 267, 491, 587, 592, 598, 607 string and membrane, 522 summary of electromagnetic, 268, E6, G6 thin rod, 493 viscous fluid, 873 Boundary layer dynamics, 602 Brake, induction, 134 MHĎ, 744 Breakdown, electrical, 576, 782 Breakdown strength of air, 576

Brush, dc machine, 292 liquid-metal, 316, 878 metal-graphite, 883 **Bullard's equation**, 336 Cables, charge relaxation in high voltage, 380 nonuniform conductivity in, 380 Capability curve of synchronous generator, 170 Capacitance, electrical linearity and, 30 example for calculation of, 32, 33 generalized, 28 quasi-static limit and, B18 Causality, boundary conditions and, 592, 607 condition of, 491, 592, 607 Center of mass, 46 Channel, variable-area MHD, 751 Characteristic dynamical times, excitation and, 332 material motion and, 332 Characteristic equation, 181 Characteristics, wave fronts and, 618 wave propagation and, 488, 490 waves with convection and, 586 Charge, B1 conservation of, B5 net flow and flow of net, B6 test, 12 total, 29 Charge-average velocity, B5 Charge carriers, effect of motion on, 290 Charge conservation, differential form of, B5 integral form of, B5 Charge density, B1 effect of motion on, 290, 334, 382, 387, 388, 392, 397, 401 free, 7, B28 magnetic field system and, 288 Charge distribution, effect of motion on, 334, 382, 387, 388, 392, 397, 401 Charge relaxation, 330, 370 electrical transient, 372 examples of, 372, 375 excitation frequency and, 378, 400 frequency in frame of material and, 399 general equation for, 371 lumped-parameter models for, 331, 375 magnetic diffusion and, 401 motion sinusoidal excitation with, 392 moving frame and, 381 nonuniform properties and, 378 sources of charge and, 372 spatially and temporally periodic fields and, 397 steady motion and, 380 thunder storms, and, 388 traveling wave in a moving material and, 397 uniform properties and, 372 Choking, constant area flow, 824

Circuit breaker, transducer for a, 22 Circuit theory, 16 Coefficient, of sliding friction, 42 of static friction, 42 Coefficients of viscosity, independence of, 870 Coenergy, 73, E5, G5 electrical linearity and, 76 potential well motions and, 217 Coenergy density, electric field system, 464, 714 magnetic field system, 456 Collector rings, 120 Commutation in dc machines, 296 Commutator, 140 of dc machines, 292 Commutator bars, 142 Commutator machines, 140 ac generator, 329 brake operation of, 306 compound wound, 310 electrical power input of, 303 equation for armature of, 300 equation for field of, 297 equation of motion for, 297 generator operation of, 306 linear amplifier, 304 mechanical power output of, 303 motor operation of, 306 operation with alternating currents and, 312 properties of, 303 separately excited, 306 series excitation of, 309 shunt excitation of, 309 speed curves of, shunt excited, 310 speed regulation of, 307 summary of equations for, 303 torque-current curves of series excited, 311 torque-speed curves of shunt excited, 310 transient performance of, 306 Compensating networks, 198 Compensation in feedback loops, 198 Compressibility constant, 845 Compressibility of fluid, 725 Compressible fluids, 813 electromechanical coupling to, 820 Conduction, electrical, 7, B30 in electric field system, effect of motion on, 371 heat, 815 motion and electrical, 284, 289 Conduction current, B6 absence of net free charge and, 374 Conduction machine, MHD, 740 variable area, MHD, 753 see also Commutator machine; DC machines Conductivity, air and water, 388 electrical, 7 electrical surface, 7 mechanical strength and, 698 nonuniform, 380 numerical values of, 345, 377

Conductor, electric field perfect, 29, 213,

390, 400, 401 magnetic field perfect, 18, 211, 223, 354, 401, 563 Confinement, electromechanical, 4, 407 Conservation, of charge, B5 displacement current and, B9 integral form of, B37 of energy, 63, 66 continuum, 456, 464 continuum coupling and, 455 equation, steady state, 820 fluid, 814 incompressible fluid, 757 integral form of, 819 of flux, lumped-parameter, 211, 220 perfectly conducting fluid and, 761 of mass, differential law of, 731 example of, 730 fluid, 729, 814 integral form of, 730 of momentum, fluid, 731, 814 integral form of, 733, 734 interfacial, 671 stress and, 733 Conservative systems, 213 Constant charge dynamics, 205, 213 Constant-current constant-flux dynamics, 220Constant-current constraint, continuum, 628 Constant-current dynamics, 220 Constant flux, lumped and continuum, 212 Constant flux dynamics, fluid, 761 lumped-parameter, 211, 220 Constant of the motion, fluid, 738 Constant voltage dynamics, 204, 212, 226 Constituent relations, electromagnetic, 283, B25 fluid, 815 fluid mechanical, 735 materials in motion and, 283 moving media in electric field systems and, 289 moving media in magnetic field systems and, 284 Constitutive law, mobile ion, 778 piezoelectric slab, 712 Contact resistance, MHD, 750 Contacts, sliding, 42 Continuity of space, 35 Continuum and discrete dynamics, 553 Continuum descriptions, 727 Continuum electromechanical systems, 251 Contour, deforming, 11, B32 Control, dc machines and, 291 Controlled thermonuclear reactions, 354 Convection, dynamical effect of, 584 and instability, 593 Convection current, B6 Convective derivative, 259, 584 charge relaxation and, 381 example of, 729 magnetic diffusion and, 357 see also Substantial derivative

Convective second derivative, 585 Coordinate system, inertial, 254 Corona discharge, 776, 782 Corona wind, demonstration of, 782 Couette flow, plane, 876 Coulomb's law, B1 point charge, B2 Coupling, electromechanical, 15, 60 Coupling to continuous media at terminal pairs, 498 Coupling network, lossless and conservative, 63 Creep, failure in solids by, 704 Critical condition for instability, 568 Crystals, electromechanics of, 651 piezoelectric materials as, 711 Current, balanced two-phase, 113 conduction, B6 convection, B6 displacement, B9 electric field system, 29 free, B25 magnetization, B25 polarization, B29 Current density, B5 diffusion of, 343 distribution of, 332 free, 7 Current law, Kirchhoff's, 16 Currents as functions of flux linkages, 26 Current transformation, examples of, 226 Cutoff condition, 559 Cutoff frequency, 559 elastic shear waves, 695 membrane, 623 Cutoff waves, 556 electromagnetic plasma, 638 membrane, 623 power flow and, 637 thin beam, 684 see also Evanescent wave Cyclic energy conversion processes, 79 Cylindrical coordinates, stress components in. 437 Cylindrical modes, 648 Damped waves, driven response of, 577 Damper, linear ideal, 40 lumped element, 36, 40 square-law, 43, 229 Damper winding in rotating machine, 164 Damping, magnetic fluid, 750 negative, 198 spatial decay and, 560 wave dynamics with, 576 Damping constant, 41 Damping frequency, 577 DC generator, magnetic saturation in, 310 self-excited, 310 DC machines, 140; see also Commutator machines DC motor, self-excited, 308 series excited, 311

Index

starting torque of, 310 torque-speed curves for, 306 Definitions, electromagnetic, 7, B1 Deforming contours of integration, 10, 18, 262, B32, 761 Degree of freedom, 49 Delay line, acoustic, 480 acoustic magnetostrictive, 708 fidelity in, 501 mechanical, 499 shear waves and, 696 Delta function, B2 Kronecker, 421 Derivative, convective, 259, 584, 726 individual, 728 particle, 728 Stokes, 728 substantial, 259, 584, 728 total, 728 Dielectrophoresis, 783 Difference equation, 620 Differential equation, order of, 180 linear, 180 Differential operators, moving coordinates and, 257 Diffusion, magnetic, 576 magnetic analogous to mechanical, 580 of magnetic field and current, 335 Diffusion equation, 337 Diffusion time constant, 341 numerical values of, 344 Diffusion wave, magnetic, 358 picture of, 581 space-time behavior of, 359 Dilatational motion of fluid, 866 Direction cosines, definition of, 435 relation among, 439 Discrete systems, electromechanics of, 60 Discrete variables, mechanical, 36 summary of electrical, 35 Dispersion equation, absolutely unstable wire, 567 Alfvén wave, 769 amplifying wave, 602 convective instability, 602 damped waves, 577 elastic guided shear waves, 695 electron oscillations, 601 evanescent wave, 557 kink instability, 629 magnetic diffusion with motion, 357 membrane, 623 moving wire destabilized by magnetic field, 602 moving wire stabilized by magnetic field, 596 ordinary waves, 513 with convection, 594 on wire, 555 on wires and membranes, 513 resistive wall interactions, 609 sinusoidal steady-state, and 514 wire with convection and damping, 609

Displacement, elastic materials, 486 elastic media, 652 lumped parameter systems, 36 one-dimensional, 483 relative, 657 and rotation, 657 and strain, 658 transformation of, 659 translational, 657 Displacement current, B9 Displacement current negligible, B19 Distributed circuits, electromechanical, 651 Divergence, surface, 272 tensor, 422, G9 theorem, B4, C2, E2, G2, G9 Driven and transient response, unstable system, 569 Driven response, one-dimensional continuum, 511 unstable wire, 568 Driving function, differential equation, 180 sinusoidal, 181 Dynamics, constant charge, 205, 213 constant current, 220 constant flux, 211, 220 constant voltage, 204, 212, 226 lumped-parameter, 179 reactance dominated, 138, 211, 220, 242, 336, 354, 368, 563 resistance dominated, 138, 209, 233, 242, 336, 354, 368, 503, 583, 611 two-dimensional, 621 Dynamics of continua, x-t plane, 488, 586 omega-k plane, 511, 554 Dynamo, electrohydrodynamic, 388 Eddy currents, 342, 628 Efficiency of induction machine, 134 EHD, 3, 552, 776 EHD pump, demonstration of, 783 Eigenfrequencies, 518 electromechanical filter, 707 magnetic field, shift of, 562 not harmonic, 563, 684 wire stiffened by magnetic field, 562 Eigenfunction, 518 Eigenmode, 517 complex boundary conditions and, 533 orthogonality of, 341, 519, 520 Eigenvalues, 518 dispersion and, 562 graphic solution for, 526 kink instability, 630 Elastic beam, resonant circuit element, 688 Elastic constants, independence of, 664 numerical values of, 486 Elastic continua, 479 Elastic failure, example of electromechanical, 701 Elastic force density, 667 Elastic guiding structures, 693 Elasticity, summary of equations of, 666, 668 Elasticity equations, steps in derivation of,

651

Elastic material, ideal, 485 linear, 485 Elastic media, 651 electromechanical design and, 697 electromechanics of, 696 equations of motion for, 653 quasi-statics of, 503 Elastic model, membrane, 509 thin rod, 480 wire, 509 Elastic waves, lumped mechanical elements and, 507 shear, 543 thin rod, 543 see also Acoustic waves Electrical circuits, 16 Electric displacement, 7, B28 Electric field, effect of motion on, 334, 382, 387, 388, 392, 397, 401 Electric field coupling to fluids, 776 Electric field equations, periodic solution to, 281 Electric field intensity, 7, B1 Electric field system, B19 differential equations for, 8, E3, G3 integral equations for, 11, E3, G3 Electric field transformation, example of, 262 Faraday's law and, 262 Electric force, field description of, 440 fluids and, 776 stress tensor for, 441 Electric force density, 418, 463 Electric Reynolds number, 335, 370, 381, 383, 395, 399, 401, 575, 780 mobility model and, 780 Electric shear, induced surface charge and, 400 Electric surface force, 447 Electrification, frictional, 552 Electroelasticity, 553 Electrogasdynamic generator, 782 Electrohydrodynamic orientation, 785 Electrohydrodynamic power generation, 782 Electrohydrodynamics, 3, 552, 776 Electrohydrodynamic stabilization, 786 Electromagnetic equations, differential, 6, B12, B19, E3, G3 integral, 9, B32, E3, G3 quasi-static, 5, B19, B32, E3, G3 summary of quasi-static, 13, E3, G3 Electromagnetic field equations, summary of, 268, E6, G6 Electromagnetic fields, moving observer and, 254 Electromagnetic theory, 5, B1 summary of, 5, E6, G6 Electromagnetic waves, B13 absorption of, B25 Electromechanical coupling, field description of, 251 Electromechanics, continuum, 330 of elastic media, 651 incompressible fluids and, 737

lumped-parameter, 60 Electron beam, 4, 552, 600, 608 magnetic field confinement of, 601 oscillations of, 600 Electrostatic ac generator, 415 Electrostatic self-excited generator, 388 Electrostatic voltmeter, 94 Electrostriction, incompressibility and, 784 Electrostriction force density, 465 Elements, lumped-parameter electrical, 16 lumped-parameter mechanical, 36 Energy, conservation of fluid, 814 electrical linearity and, 76 electric field system conservation of, 66 internal or thermal gas, 813 internal per unit mass, 815 kinetic per unit mass, 815 magnetic field system conservation of, 63 magnetic stored, 64 potential and kinetic, 214 Energy conversion, cyclic, 79, 110 electromechanical, 79 lumped-parameter systems, 79 Energy density, B23 equal electric and magnetic, B24 Energy dissipated, electromagnetic, B22 Energy flux, B22 Energy function, hybrid, 220 Energy method, 60, 450, E5, G5 Energy relations, summary of, 68, E5, G5 Enthalpy, specific, 820 Equation of motion, elastic media, 668 electromechanical, 84 examples of lumped-parameter, 84, 86 incompressible, irrotational inviscid flow, 738 linearized, 183 lumped mechanical, 49 Equilibrium, of continuum, stability of, 574 dynamic or steady-state, 188 hydromagnetic, 561 kink instability of, 633 potential well stability of, 216 static, 182 Equipotentials, fluid, 752 Eulerian description, 727 Evanescence with convection, 596 Evanescent wave, 556 appearance of, 559 constant flux and, 563 dissipation and, 560 elastic shear, 695 equation for, 557 example of, 556 membrane, 560, 623 physical nature of, 560 signal transmission and, 639 sinusoidal steady-state, 558 thin beam, 684 Evil, 697 Failure in solids, fatigue and creep, 704

Faraday, 1

Faraday disk, 286 Faraday's law, B9 deforming contour of integration and, 262, 300, 315, 565, B32, E3, G3 differential form, 6, B10, E3, G3 example of integral, 262, 276, 286, 297, 315 integral form of, B10, B32 perfectly conducting fluid and, 761 Fatigue, failure in solids by, 704 Feedback, continuous media and, 548 stabilization by use of, 193 Ferroelectrics, B29 piezoelectric materials and, 711 Ferrohydrodynamics, 552, 772 Field circuit of dc machine, 141 Field equations, moving media, generalization of, 252 Fields and moving media, 251 Field transformations, 268, E6, G6; see also Transformations Field winding, ac machine, 120 dc machine, 293 Film, Complex Waves I, xi, 516, 559, 571, 634 Film, Complex Waves II, xi, 573, 606 Filter, electromechanical, 2, 200, 480, 704 First law of thermodynamics, 63 Flow, Hartmann, 884 irrotational fluid, 737 laminar, 725 turbulent, 725 Flowmeter, liquid metal, 363 Fluid, boundary condition for, 725 boundary condition on, inviscid, 752 compressibility of, 725 effect of temperature and pressure on, 724 electric field coupled, 776 electromechanics of, 724 ferromagnetic, 552, 772 ferromagnetic, 552, 772 highly conducting, 760 incompressible, 724, 735 inhomogeneous, 735 internal friction of, 724 inviscid, 724, 725 laminar and turbulent flow of, 725 magnetic field coupling to incompressible, 737 magnetizable, 772 Newtonian, 861 perfectly conducting, 563 solids and, 724 static, 735 viscous, 861 Fluid dynamics, equations of inviscid compressible, 813 equations of inviscid, incompressible, 726 equations of viscous, 871 Fluid flow, accelerating but steady, 753 around a corner, 751 potential, 751 unsteady, 746 variable-area channel, 751

Fluid-mechanical examples with viscosity, 875 Fluid orientation systems, 785 Fluid pendulum, electric-field coupled, 784 magnetic damping of, 750 Fluid pump or accelerator, 776 Fluid stagnation point, 752 Fluid streamlines, 752 Fluid transformer, variable-area channel as, 756 Flux amplification, plasmas and, 354 Flux conservation, lumped-parameter, 211, 220 magnetic fields and, 352 perfectly conducting gas and, 849 Flux density, mechanical amplification of, 354 Flux linkage, 19, E4, G4 example of, 22, 23 Force, charge, B1 derivative of inductance and, 453 electric origin, 67, E5, G5 electromagnetic, 12 field description of, 418 fluid electric, 776 Lorentz, 12, 255, 419 magnetic, B6 magnetization with one degree of freedom, 451 physical significance of electromagnetic, 420 polarized fluids, 463, 572, 784 single ion, 778 surface integral of stress and, 422 Force-coenergy relations, 72, E5, G5 Force density, 7 averaging of electric, 440 averaging of magnetic, 419 divergence of stress tensor and, 422, 427, G9 effect of permeability on, 455, 456 elastic medium, 667 electric, 12, B3, 440, G11 magnetic field systems, 419, 462, G11 electromagnetic fluid, 732 electrostriction, 465, G11 fluid mechanical, 732 fluid pressure, 736 free current, 419, G11 inviscid fluid mechanical, 737 lumped parameter model for, 455 magnetic, 12, 419, B9 magnetization, 448, 450, 462, G11 magnetostriction, 461, 462, G11 polarization, 450, 463, G11 summary of, 448, G11 Forced related to variable capacitance, 75 Force-energy relations, 67, E5, G5 examples of, 70 Force equations, elastic media, 653 Force of electric origin, 60, E5, G5 Fourier series, 340 Fourier transform, two-dimensional, 617 Fourier transforms and series, diffusion

Index

equation and, 340 eigenmodes as, 517 linear continuum systems and, 511, 554, 617 linear lumped systems and, 200 mutual inductance expansions and, 108, 153 Frame of reference, laboratory, 254 Free-body diagram, 49 Free charge density, B28 Free charge forces, avoidance of, 787 Frequency, complex, 181, 554 complex angular, 554 natural, 181, 515 voltage tuning of, 704 Frequency conditions for power conversion, 111, 155 Frequency response of transducer, 204 Friction, coulomb, 42 Frozen fields, perfectly conducting gas and, 849 Fusion machines, instability of, 571 Galilean transformation, 584 Gamma rays, B13 Gas, perfect, 816 Gas constant, 816 universal, 816 Gases, definition of, 724 ionized, 813 Gauss's law, differential form of, B5 example of, B4 integral form of, B3 magnetic field, B12 polarization and, B28 Gauss's theorem, tensor form of, 423, G9 Generators, electric field, 778 electrohydrodynamic applications of, 3 hydroelectric, 152 induction, 134 magnetohydrodynamic applications of, 3 MHD, 744 Van de Graaff, 3, 383, 385 Geometrical compatibility, 53 Geophysics and MHD, 552 Gravitational potential, 733 Gravity, artificial electric, 785 force density due to, 732 waves, 794 Group velocity, 614 power flow and, 638 unstable media and, 617 Guiding structures, evanescence in, 560 Hartmann flow, 884 Hartmann number, 887 Heat transfer, EHD and, 552 Homogeneity, B27 Homogeneous differential equation, solution of, 180 Homopolar machine, 286, 312 armature voltage for, 314 speed coefficient for, 314

summary of equations for, 316 torque for, 316 Hunting transient of synchronous machine, 192 Hydraulic turbine, 151 Hydroelectric generator, 152 Hydromagnetic equilibria, 561, 571 Hysteresis, magnetic, B27 Identities, C1, E1, G1 Impedance, characteristic, 497 Incompressibility, fluid, 735 Incompressible fluids, MHD, 737 Incompressible media, 380 Incremental motions, see Linearization Independence of variables, 69, 97 (see Problem 3.16) Independent variables, change of, 72 Index notation, 421, G7 Inductance, calculation of, 22 electrical linearity and, 20 generalized, 17 quasi-static limit and, B18 Induction, demonstration of motional, 253 law of, B9; see also Faraday's law Induction brake, 134 Induction generator, 134 electric, 400 Induction interaction, 367 Induction law, integral, B32; see also Faraday's law Induction machine, 127 coefficient of coupling for, 135 distributed linear, 368 efficiency of, 134 equivalent circuit for, 131 loading of, 137 lumped-parameter, 368 MHD, 745 power flow in, 133 reactance and resistance dominated, 137 single phase, 138 squirrel-cage, 129 starting of, 137, 139 torque in, 132 torque-slip curve of, 135 variable resistance in, 136 wound rotor, 106 Induction motor, 134 Inductor, 17 Inelastic behavior of solids, 699 Influence coefficients, MHD, 822 variable-area MHD machine, 832 Initial and boundary conditions, 513 Initial conditions, convection and, 587 one-dimensional continuum, 488, 512 Initial value problem, continuum, 488 Instability, absolute, 566 and convective, 604 aeroelastic absolute, 793 convective, 601 dynamic, 192 electrohydrodynamic, 571 engineering limitations from convective, 604

and equilibrium, example of, 185 failure of a static argument to predict, 192 fluid pendulum, 785 fluid turbulence and, 725 graphical determination of, 184 heavy on light fluid, 571 and initial conditions, 184 kink, 627 linear and nonlinear description of, 216 nonconvective, 566 nonlinearity and, 570 omega-k plot for, 569 plasma, 553 in presence of motion, 583 Rayleigh-Taylor, 571 resistive wall, 576, 608 space-time dependence of absolute, 570 static, 182 in stationary media, 554 Integral laws, electromagnetic, 9, B32, E3, G3 Integrated electronics, electromechanics and, 688 Integration contour, deforming, 11, B32 Internal energy, perfect gas, 816 Invariance of equations, 256 Inviscid fluid, boundary condition for, 752 Ion beams, 552 Ion conduction, moving fluid and, 778 Ion drag, efficiency of, 782 Ion-drag phenomena, 776 Ionized gases, acceleration of, 746 Ion source, 776 Isotropic elastic media, 660 Isotropy, B27 Kinetic energy, 214 Kirchhoff's current law, 16 Kirchhoff's laws, 15 electromechanical coupling and, 84 Kirchhoff's voltage law, 16 Klystron, 601 Kronecker delta function, 421, G7 Lagrangian coordinates, 652 surface in, 669 Lagrangian description, 727 Lagrangian to Eulerian descriptions, 483 Lamé constant, 667 numerical values of, 677 Laplace's equation, fluid dynamics and, 737 two-dimensional flow and, 751 Leakage resistance of capacitors, 377 Legendre transformation, 73 Length expander bar, equivalent circuit for, 716 piezoelectric, 712 Levitating force, induction, 369 Levitation, electromechanical, 4, 195, 365, demonstration of magnetic, 370 and instability, 574 of liquids, EHD, 552 MHD, 552

solid and liquid magnetic, 365 Light, velocity of, B14 Linearity, electrical, 20, 30, B27 Linearization, continuum, 483, 510, 556, 652, 842 error from, 224 lumped-parameter, 182 Linear systems, 180 Line integration in variable space, 64, 67 Liquid drops, charge-carrying, 388 Liquid level gauge, 416 Liquid metal brush, 878 numerical example of, 883 Liquid metal MHD, numerical example of 750 Liquid metals, pumping of, 746 Liquid orientation in fields, 785 Liquids, definition of, 724 Liquids and gases, comparison of, 724 Loading factor, MHD machine, 833 Lodestone, B25 Long-wave limit, 283, 574 thin elastic rod and, 683 Lord Kelvin, 389 Lorentz force, 419 Loss-dominated dynamics, continuum, 576 Loss-dominated electromechanics, 229, 249 Loss-dominated systems, 227 Losses, fluid joule, 815 Loudspeaker, model for, 527 Lumped-parameter electromechanics, 60 Lumped-parameter variables, summary of, 35, E4, G4 Mach lines, 624 Mach number, 624, 823 Macroscopic models, electromagnetic, B25 Magnet, permanent, 27 Magnetic axes of rotating machines, 105 Magnetic circuit, example of, 22, 23 Magnetic diffusion, 330, 335 charge relaxation compared to, 401 competition between motion and, 351 cylindrical geometry and, 408 effect of motion on, 354 electrical transient, 338 induction machines and, 746 initial conditions for, 339 limit, of infinite conductivity in, 343 of small conductivity in, 343 liquid metals and, 354 lumped-parameter models for, 331, 334, 336 sinusoidal steady-state, 358 sinusoidal steady-state with motion, 355 steady-state, 337, 347 steady-state in the moving frame, 351 traveling-wave moving media, 364 Magnetic diffusion time, 341, 772 Magnetic field, air-gap, 114 induced and imposed, 212, 286, 332 origin of earths, 336, 552 Magnetic field compression, 354 Magnetic field equations, insulating me-

dium, 773 Magnetic field intensity, 7, B25 Magnetic field system, 6, B19 differential equations for, 6, B20, E6, G6 integral equations for, 10, B32, E3, G3 Magnetic field transformation, example of, 266; see also Transformations Magnetic fluid, demonstration of, 777 Magnetic flux density, 7, B6 Magnetic flux lines, frozen, 763 Magnetic force, field description of, 418, G11 stress tensor for, 422, G11 Magnetic forces and mechanical design, 697 Magnetic induction negligible, B19 Magnetic piston, 354 Magnetic pressure, 369 Magnetic Reynolds numbers, 333, 349, 351, 353, 357, 401, 628, 741, 821 MHD flow, 754 numerical value of, 354 Magnetic saturation in commutator machines, 297 Magnetic surface force, 447 Magnetic tension, 767 Magnetization, B25 effect of free current forces on, 455 Magnetization currents, B25 Magnetization density, 7, B25 Magnetization force, fluids and, 772 one degree of freedom and, 451 Magnetization force density, changes in density and, 461 example of, 460 inhomogeneity and, 460 in moving media, 285 summary of, 448, G11 Magnetoacoustic velocity, 850 Magnetoacoustic wave, 846 electrical losses and, 860 flux and density in, 851 numerical example, in gas, 852 in liquid, 853 Magnetoelasticity, 553 Magnetofluid dynamics, 551 Magnetogasdynamics, 551 Magnetohydrodynamic conduction machine, 740 Magnetohydrodynamic generator, constantarea, 821 variable-area, 828 Magnetohydrodynamics, 551 constant-area channel, 740 viscosity and, 725 Magnetohydrodynamics of viscous fluids, 878 Magnetostriction, 697 one degree of freedom and, 452 Magnetostriction force, incompressible fluid and, 776 Magnetostrictive coupling, 707 Magnetostrictive transducer, terminal representation of, 711 Mass, conservation of fluid, 729

elastic continua, quasi-static limit of, 507 lumped-parameter, 36, 43 total, 46 Mass conservation, 731 Mass density, 45 elastic materials, numerical values of, 486 of solid, 486 numerical values of, 486, G12 Mass per unit area of membrane, 509 Mass per unit length of wire, 511 Matched termination, 497 Material motion, waves and instabilities with, 583 Matter, states of, 724 Maxwell, 1 Maxwell's equations, B12 limiting forms of, B14 Maxwell stress tensor, 420, 441, G7, G11 Mechanical circuits, 36 Mechanical continuum, 479 Mechanical equations, lumped-parameter, 49 Mechanical input power, fluid, 756 variable-area channel, 756 Mechanical lumped-parameter equations, examples of, 49, 51, 53 Mechanics, lumped-parameter, 35 rigid body, 35 transformations and Newtonian, 254 Membrane, elastic continua and, 509, 535, electric field and, 574 equations of motion for, 511, 535, G13 two-dimensional modes of, 622 Membrane dynamics with convection, 584 Mercury, density and conductivity of, 750 properties of, 883 Meteorology, ÉHD and, 552 MFD, 551; see also MHD MGD, 551; see also MHD MHD, 551 compressible fluids and, 813 liquid metal numerical example of, 750 magnetic damping in, 750 transient effects in, 746, 759 transient example of, 750 variable-area channel in, 751 of viscous fluids, 878 MHD conduction machine, 821, 828 equivalent circuit for, 742 pressure drop in, 742 terminal characteristics of, 742 MHD constant-area channel, 740, 820 MHD flows, dynamic effects in, 746 MHD generator, comparison of, 839 compressibility and, 820 constant voltage constrained, 743 distribution of properties in, 827 end effects in, 797 examples of, 840, 841 Mach number in, 823 numerical example of, 826 temperature in, 823 variable-area channel, 828 viscosity and, 725, 884

Newton's laws, 15, 35

elastic media and, 653 Newton's second law, 44, 50

MHD machine, compressible and incompressible, 825 constant velocity, loading factor and aspect ratio, 834 dynamic operation of, 746 equivalent circuit for variable area, 756 loading factor of, 833 operation of brake, pump, generator, 744 quasi-one-dimensional, 828 steady-state operation of, 740 velocity profile of, 891 MHD plane Couette flow, 884 MHD plane Poiseuille flow, 878 MHD pressure driven flow, 884 MHD pump or accelerator, 824 MHD transient phenomena, 759 MHD variable-area channel equations, conservation of energy and, 831, 833 conservation of mass and, 831, 833 conservation of momentum and, 831, 833 local Mach number and, 823, 833 local sound velocity and, 822, 833 mechanical equation of state and, 816, 833 Ohm's law and, 830, 833 thermal equations of state and, 820, 833 MHD variable-area machine, equations for, 833 MHD variable-area pumps, generators and brakes, 751 Microphone, capacitor, 201 fidelity of, 204 Microphones, 200 Microwave magnetics, 553 Microwave power generation, 552 Mobility, 289, B31 ion, 778 Model, engineering, 206 Modulus of elasticity, 485 numerical values of, 486, G12 Molecular weight of gas, 816 Moment of inertia, 36, 48 Momentum, conservation of, see Conserva-tion of momentum Momentum density, fluid, 734 Motor, commutator, 140, 291 induction, 134 reluctance, 156 synchronous, 119 Moving media, electromagnetic fields and, 251 Mutual inductance, calculation of, 22 Natural frequencies, 515 dispersion equation and, 517 Natural modes, dispersion and, 561 kink instability, 635 of membrane, 624, 625 overdamped and underdamped, 583 of unstable wire, 569 Navier-Stokes equation, 872 Negative sequence currents, 144 Networks, compensating, 198 Newtonian fluids, 861

electromechanical coupling and, 84 fluid and, 729, 731 Node, mechanical, 36, 49 Nonlinear systems, 206, 213 Nonuniform magnetic field, motion of conductor through, 367 Normal modes, 511 boundary conditions and, 524 Normal strain and shear stress, 662 Normal stress and normal strain, 661 Normal vector, analytic description of, 269 Oerstad, 1, B25 Ohm's law, 7, B30 for moving media, 284, 298 Omega-k plot, absolutely unstable wire, 567 amplifying wave, 603 convective instability, 603 damped waves, complex k for real omega, 579 elastic guided shear waves, 695 electron oscillations, 601 evanescent wave, 557, 559, 597, 615, 695 moving wire, with destabilizing magnetic force, 603 with resistive wall, complex k for real omega, 611 with resistive wall, complex omega for real k, 610 ordinary wave, with convection, 594 on wires and membranes, 514 ordinary waves, 514, 555 unstable eigenfrequencies and, 569 waves with damping showing eigenfrequencies, 582 wire stabilized by magnetic field, 557 Orientation, electrohydrodynamic, 571 electromechanical, 4 of liquids, dielectrophoretic, 785 EHD, 552 Orthogonality, eigenfunctions and, 341, 519, 520 Oscillations, nonlinear, 226 with convection, 596 Oscillators in motion, 599 Overstability, 192 Particles, charge carriers and, 782 Particular solution of differential equation, 180 Pendulum, hydrodynamic, 746 simple mechanical, 214 Perfect conductor, no slip condition on, 769 Perfect gas law, 816 Perfectly conducting gas, dynamics of, 846 Perfectly conducting media, see Conductor Permanent magnet, in electromechanics, 27 example of, 28 as rotor for machine, 127 Permanent set, solids and, 700

Permeability, 7, B27 deformation and, 459 density dependence of, 454 free space, 7, B7 Permittivity, 9, B30 free space, 7, 9, B2 Perturbations, 183 Phase sequence, 144 Phase velocity, 613 diffusion wave, 358 dispersive wave, 598 membrane wave, 512 numerical elastic compressional wave, 677 numerical elastic shear wave, 677 numerical thin rod, 486, G12 ordinary wave, 487 thin rod, 487 wire wave, 512 Physical acoustics, 553, 651 Piezoelectric coupling, 711 reciprocity in, 712 Piezoelectric devices, example of, 717 Piezoelectricity, 553, 711 Piezoelectric length expander bar, 712 Piezoelectric resonator, equivalent circuit for, 716 Piezoelectric transducer, admittance of, 714 Piezomagnetics, 553 Plane motion, 44 Plasma, confinement of, 552 electromechanics and, 4 evanescent waves in, 561, 638 heating of, 552 lumped-parameter model for, 223 magnetic bottle for, 563 magnetic diffusion and, 408 MHD and, 553 solid state, 553 Plasma dynamics, 553 Plasma frequency, 600 Poiseuille flow, plane, 878 Poisson's ratio, 662 numerical values of, 666 Polarization, effect of motion on, 290 current, B29 density, 7, B28 electric, B27 force, 463, 571, G11 Polarization force, one degree of freedom, 464 Polarization interactions, liquids and, 783 Polarization stress tensor, 463, G11 Pole pairs, 148 Poles in a machine, 146 Polyphase machines, 142 Position vector, 45 Positive sequence currents, 144 Potential, electric, B9 electromagnetic force, 738 gravitational, 733 mechanical, 214 velocity, 737

Potential difference, B10 Potential energy, 214 Potential flow, irrotational electrical forces and, 738 Potential fluid flow, two-dimensional, 751 Potential plot, 214 Potential well, electrical constraints and, 217 electromechanical system and, 217 temporal behavior from, 224 Power, conservation of, 64 Power density input to fluid, 818 Power factor, 126 Power flow, group velocity and, 638 ordinary and evanescent waves and, 638 rotating machines and, 110 Power generation, ionized gases and, 552 microwave, 552, 553 Power input, electrical, 64 fluid electrical, 818 mechanical, 64 mechanical MHD, 743 Power input to fluid, electric forces and, 819 electrical losses and, 818, 819 magnetic forces and, 818 pressure forces and, 818 Power output, electric MHD, 743 Power theorem, wire in magnetic field, 637, 644 Poynting's theorem, B22 Pressure, density and temperature dependence of, 816 hydrostatic, 735 hydrostatic example of, 736 incompressible fluids and significance of, 753 isotropic, 735 magnetic, 369 normal compressive stress and, 735 significance of negative, 753 velocity and, 753 Principal axes, 49 Principal modes, 681 elastic structure, 679 shear wave, 695 Principle of virtual work, see Conservation, of energy Products of inertia, 48 Propagation, 613 Propulsion, electromagnetic, 552 electromechanical, 4 MHD space, 746 Pulling out of step for synchronous machine, 125 Pump, electric field, 776 electrostatic, 778 liquid metal induction, 365 MHD, 744, 746 variation of parameters in MHD, 825 Pumping, EHD, 552 MHD. 552

Quasi-one-dimensional model, charge relaxa-

tion, 392, 394 electron beam, 600 gravity wave, 794 magnetic diffusion, 347 membrane, 509, 648 and fluid, 793 MHD generator, 828 thin bar, 712 thin beam, 683 thin rod, 480, 681 wire or string, 509 in field, 556, 563, 574, 605, 627 Quasi-static approximations, 6, B17 Quasi-static limit, sinusoidal steady-state and, 515, 534 wavelength and, B17 wire and, 534 Quasi-statics, conditions for, B21 correction fields for, B21 elastic media and, 503 electromagnetic, B19 Quasi-static systems, electric, 8 magnetic, 6 Radiation, heat, 815 Rate of strain, 864 Reactance-dominated dynamics, 138, 211, 220, 242, 336, 354, 368, 563, 759 Reciprocity, electromechanical coupling and, 77 piezoelectric coupling and, 713 Reference system, inertial, 44 Regulation, transformer design and, 699 Relative displacement, rotation, strain and, 658 Relativity, Einstein, 254 Galilean, 255 postulate of special, 261 theory of, 44 Relaxation time, free charge, 372 numerical values of, 377 Relay, damped time-delay, 229 Reluctance motor, 156 Resistance-dominated dynamics, 138, 209, 233, 242, 336, 354, 368, 503, 583, 611 MHD, 750 Resistive wall damping, continuum, 583 Resistive wall instability, nature of, 612 Resistive wall wave amplification, 608 Resonance, electromechanically driven continuum and, 533 response of continua and, 515 Resonance frequencies, magnetic field shift of, 563 membrane, 624 natural frequencies and, 515 Resonant gate transistor, 688 Response, sinusoidal steady-state, 181, 200, 514 Rigid body, 44 Rigid-body mechanics, 35 Rotating machines, 103 air-gap magnetic fields in, 114

applications of, 3 balanced two-phase, 113 classification of, 119 commutator type, 140, 255, 292 computation of mutual inductance in, 22 dc, 140, 291 differential equations for, 106 effect of poles on speed of, 149 electric field type, 177 energy conversion conditions for, 110 energy conversion in salient pole, 154 equations for salient pole, 151 hunting transient of synchronous, 192 induction, 127 losses in, 109 magnetic saturation in, 106 mutual inductance in, 108 number of poles in, 146 polyphase, 142 power flow in, 110 salient pole, 103, 150 single-phase, 106 single-phase salient-pole, 79 smooth-air-gap, 103, 104 stresses in rotor of, 697 superconducting rotor in, 92 synchronous, 119 two-phase, smooth-air-gap, 111 winding distribution of, 108 Rotating machines, physical structure, acyclic generator, 287 commutator type, 292 dc motor, 293 development of dc, 295 distribution of currents and, 166, 169 four-pole, salient pole, 164 four-pole, single phase, 147 homopolar, 313 hydroelectric generator, 152 multiple-pole rotor, 146 rotor of induction motor, 107 rotor of salient-pole synchronous, 151 synchronous, salient-pole, 152 salient-pole, two phase, 158 salient-pole, single phase, 150 smooth-air-gap, single phase, 104 stator for induction motor, 106 three-phase stator, 145 turboalternator, 120 two-pole commutator, 294 Rotation, fluid, 865 Rotation vector, 658 Rotor of rotating machines, 104, 107, 112, 120, 146, 147, 150, 151, 152, 158, 164, 166, 169 Rotor teeth, shield effect of, 301 Saliency in different machines, 156 Salient-pole rotating machines, 103, 150 Salient poles and dc machines, 293 Servomotor, 140 Shading coils in machines, 139 Shear flow, 862, 864, 875

magnetic coupling, 878 Shear modulus, 664 numerical values of, 666 Shear rate, 866 Shear strain, 543, 655 normal strain and, 663 shear stress and, 664 Shear stress, 543 Shear waves, elastic slab and, 693 Shearing modes, beam principal, 683 Shock tube, example related to, 276 Shock waves, supersonic flow and, 592 Sinusoidal steady-state, 181, 200, 514 convection and establishing, 592 Sinusoidal steady-state response, elastic continua, 514 Skin depth, 357 numerical values of, 361 Skin effect, 358 effect of motion on, 361 Slip of induction machine, 131 Slip rings, 120 ac machines and, 120 Slots of dc machine, 296 Sodium liquid, density of, 771 Solids, definition of, 724 Sound speed, gases, 844 liquids, 845 Sound velocity, see Velocity Sound waves, see Acoustic waves Source, force, 37 position, 36 velocity, 37 Space charge, fluid and, 780 Space-charge oscillations, 601 Speakers, 200 Specific heat capacity, constant pressure, 817 constant volume, 816 ratio of, 817 Speed coefficient, of commutator machine, torque on dc machine and, 302 Speed control of rotating machines, 149 Speedometer transducer, 170 Speed voltage in commutator machine, 299 Spring, linear ideal, 38 lumped element, 36, 38 quasi-static limit of elastic continua and, 505 torsional, 40 Spring constant, 39 Stability, 182, 566, 583 Stagnation point, fluid, 752 Standing waves, electromagnetic, B16 electromechanical, 516, 559, 596, 771 State, coupling network, 61, 65 thermal, 816 Stator, of rotating machines, 104, 106, 120, 145, 147, 150, 152, 158, 164, 166, 169 smooth-air-gap, 103 Stinger, magnetic, 193 Strain, formal derivation of, 656 geometric significance of, 654 normal, 654 permanent, 700 shear, 543, 654

as a tensor, 659 thin rod, normal, 484 Strain components, 656 Strain-displacement relation, 653 thin-rod, 485 Strain rate, 724, 864 dilatational, 869 Strain-rate tensor, 864 Streaming electron oscillations, 600 Streamline, fluid, 752 Stress, fluid isotropy and, 868 fluid mechanical, 872 hydrostatic, 724 limiting, 700 normal, 432 shear, 432, 543 and traction, 424, G9 Stress components, 425 Stress-strain, nonlinear, 700 Stress-strain rate relations, 868 Stress-strain relation, 660, 668 thin-rod, 485 Stress-tensor, elastic media and, 667 example of magnetic, 428 magnetization, 462, G11 Maxwell, 420 physical interpretation of, 425, G7 polarization, 463, G11 pressure as, 735 properties of, 423, G7 surface force density and, 446, G9 symmetry of, 422 total force and, 444, G9 Stress tensors, summary of, 448, G11 String, convection and, 584 equation of motion for, 511, 535 and membrane, electromechanical coupling to, 522 see also Wire Subsonic steady MHD flow, 823 Subsonic velocity, 587 Substantial derivative, 259, 584, 726; see also Convective derivative Summation convention, 421, G7 Superconductors, flux amplification in, 354 Supersonic steady MHD flow, 823 Supersonic steady-state dynamics, 524 Supersonic velocity, 587 Surface charge density, free, 7 Surface conduction in moving media, 285 Surface current density, free, Surface force, example of, 449 magnetization, 775 Surface force densities, summary of, 448, G11 Surface force density, 445, G11 free surface charge and, 447, G11 free surface currents and, 447, G11 Surface tension, 605 Susceptance, electromechanical driving, 531 Susceptibility, dielectric, 9, B30 electric, 9, B30 magnetic, 7, B27 Suspension, magnetic, 193

Symbols, A1, D1, F1 Symbols for electromagnetic quantities, 7 Synchronous condenser, 127 synchronous machine, 119 equivalent circuit of, 123 hunting transient of, 192 phasor diagram for, 124, 162 polyphase, salient-pole, 157 torque in, 122, 123, 125 torque of salient-pole, two-phase, 160, 162 Synchronous motor, performance of, 126 Synchronous reactance, 123 Synchronous traveling-wave energy conversion, 117 Tachometer, drag-cup, 363 Taylor series, evaluation of displacement with, 483 multivariable, 187 single variable, 183 Teeth of dc machine, 296 Temperature, electrical conductivity and, 380 Tension, of membrane, 509 of wire, 511 Tensor, first and second order, 437 one-dimensional divergence of, 482 surface integration of, 428, 441, 444, G9 transformation law, 437, G10 transformation of, 434, G9 Tensor strain, 659 Tensor transformation, example of, 437 Terminal pairs, mechanical, 36 Terminal variables, summary of, 35, E4, G4 Terminal voltage, definition of, 18 Theorems, C2, E2, G2 Thermonuclear devices, electromechanics and, 4 Thermonuclear fusion, 552 Theta-pinch machine, 408 Thin beam, 683 boundary conditions for, 687 cantilevered, 688 deflections of, 691, 692 eigenvalues of, 692 electromechanical elements and, 688, 691, 701, 704 equation for, 687 resonance frequencies of, 692 static loading of, 701 Thin rod, 681 boundary conditions for, 494 conditions for, 683 equations of motion for, 485, G13 force equation for, 484 longitudinal motion of, 480 transverse motions of, 682 Three-phase currents, 143 Time constant, charge relaxation, 372 magnetic diffusion, 341 Time delay, acoustic and electromagnetic, 499 Time-delay relay, electrically damped, 249 Time derivative, moving coordinates and, 258 Time rate of change, moving grain and, 727 Torque, dc machine, 302 electrical, 66

Lorentz force density and, 301 pull-out, 124 Torque-angle, 123 Torque-angle characteristic of synchronous machine, 125 Torque-angle curve, salient-pole synchronous machine, 163 Torque-slip curve for induction machine, 135 Torque-speed curve, single phase induction machine, 139 Torsional vibrations of thin rod, 543 Traction, 424, 432 pressure and, 735 stress and, 432, G9 Traction drives, 310 Transducer, applications of, 2 continuum, 704 example of equations for, 84, 86 fidelity of, 203 incremental motion, 180, 193, 200 Magnetostrictive, 708 Transfer function capacitor microphone, 204 electromechanical filter, 706 Transformations, electric field system, 264 Galilean coordinate, 254, 256 integral laws and, 11, 276, 300, 315, B32 Lorentz, 254 Lorentz force and, 262 magnetic field system, 260 primed to imprimed frame, 439 summary of field, 268, E6, G6 vector and tensor, 434, G9 Transformer, electromechanical effects in, 697 step-down, 698 tested to failure, 698 Transformer efficiency, mechanical design and, 699 Transformer talk, 697 Transient response, convective instability, 621 elastic continua, 517 MHD system, 751 one-dimensional continua, 511 superposition of eigenmodes in, 518 supersonic media, 593 Transient waves, convection and, 587 Transmission line, electromagnetic, B16 parallel plate, B15 thin rod and, 488 Transmission without distortion in elastic structures, 696 Traveling wave, 487 convection and, 586 magnetic diffusion in terms of, 357 single-phase excitation of, 118 standing wave and, 116 two-dimensional, 622 two-dimensional elastic, 694 two-phase current excitation of, 116 Traveling-wave induction interaction, 368 Traveling-wave MHD interaction, 746 Traveling-wave solutions, 554 Traveling-wave tube, 602 Turboalternator, 120 Turbulence in fluids, 725 Turbulent flow, 43

Ultrasonic amplification, 602 Ultrasonics in integrated electronics, 688 Units of electromagnetic quantities, 7 Van de Graaff generator, example of, 383, 385 gaseous, 778 Variable, dependent, 180 independent, differential equation, 180 thermodynamic independent, 64 Variable capacitance continuum coupling, 704 V curve for synchronous machine, 125 Vector, transformation of, 434, 659 Vector transformation, example of, 435 Velocity, absolute, 44 acoustic elastic wave, 673, 677 acoustic fluid wave, 844, 846 Alfvén wave, 763, 772 charge-average, B5 charge relaxation used to measure, 396 charge relaxation wave, 395 compressional elastic wave, 673, 677 dilatational elastic wave, 673, 677 elastic distortion wave, 675, 677 fast and slow wave, 586 light wave, B14 magnetic diffusion wave, 358 magnetic flux wave, 114 magnetoacoustic wave, 850, 852 measurement of material, 356, 362 membrane wave, 512 phase, 488 shear elastic wave, 675, 677 thin rod wave, 486, 487, 682 wavefront, 618 with dispersion, 598 wire or string wave, 512 Velocity potential, 737 Viscosity, 862 coefficient of, 863 examples of, 875 fluid, 724 mathematical description of, 862 second coefficient of, 871 Viscous flow, pressure driven, 877 Viscous fluids, 861 Viscour losses, turbulent flow, 725 Voltage, definition of, B10 speed, 20, 21 terminal, 18 transformer, 20, 21 Voltage equation, Kirchhoff, 16 Ward-Leonard system, 307 Water waves, 794 Wave amplification, 601 Wave equation, 487 Wavenumber, 357, 513

complex, 554, 607 Wave propagation, 487 characteristics and, 487, 586, 618 group velocity and, 616 phase velocity and, 613 Wave reflection at a boundary, 493 Waves, acoustic elastic, 673 acoustic in fluid, 544, 841, 842, 845 Alfvén, 759 compressional elastic, 673 convection and, 586 cutoff, see Cutoff waves damping and, 576 diffusion, 355, 576 dilatational, 672 dispersionless, 555 dispersion of, 488 of distortion, 675 elastic shear, 678 electromagnetic, B13, 488 electromechanical in fluids, 759 evanescent, see Evanescent waves fast and slow, 586 fast and slow circularly polarized, 631 fluid convection and, 860 fluid shear, 760 fluid sound, 813 incident and reflected at a boundary, 494 light, B13 longitudinal elastic, 673 magnetoacoustic, 841, 846 motion and, 583 plasma, 553, 600, 638 radio, B13 rotational, 671 shear elastic, 675 stationary media and, 554 surface gravity, 794 thin rod and infinite media, 673 Wave transients, solution for, 490 Wind tunnel, magnetic stinger in, 193 Windings, balanced two-phase, 113 dc machine, 292 lap, 296 wave, 296 Wire, continuum elastic, 509, 535 convection and dynamics of, 584 dynamics of, 554 equations of motion for, 511, G13 magnetic field and, 556, 566, 627 two-dimensional motions of, 627 Yield strength, elastic, 700 Young's modulus, 485, G12 Zero-gravity experiments, KC-135 trajec-

tory and, 787